



The Era of Tiny IoT Platforms

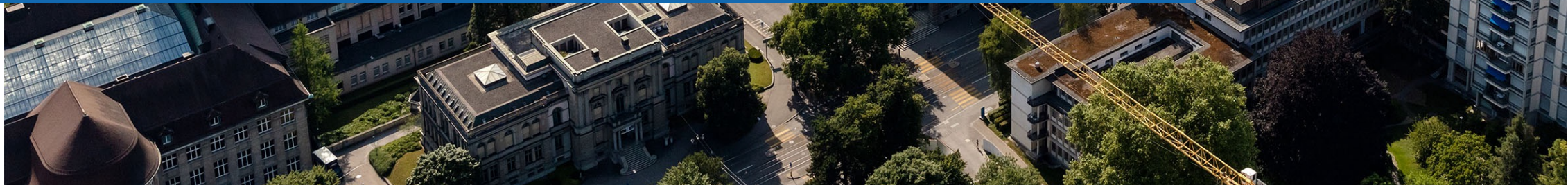
Integrated Circuits for Intelligent Biomedical/Neuromorphic Sensors

Kwantae Kim

Established Researcher at ETH Zurich, Switzerland



Tuesday, April 23, 2024



Outline

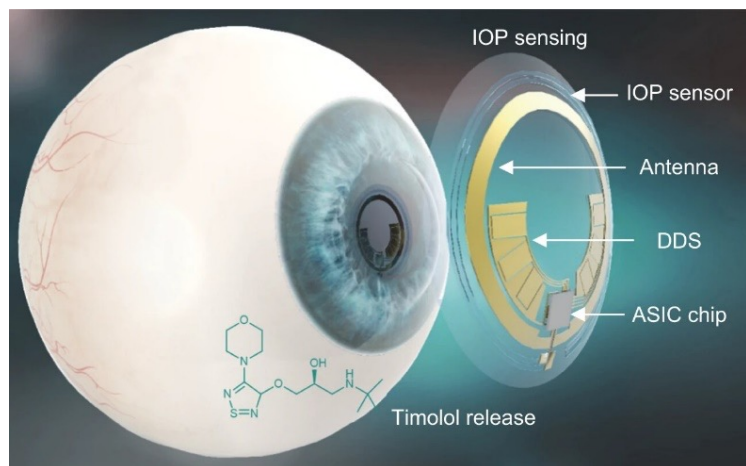
- **Visions of IoT Sensors**
- **Biomedical Sensor**
- **Neuromorphic Sensor**
- **Outlook**

Outline

- **Visions of IoT Sensors**
- Biomedical Sensor
- Neuromorphic Sensor
- Outlook

Era of Internet of Things (IoT)

Smart Healthcare



Smart Farming



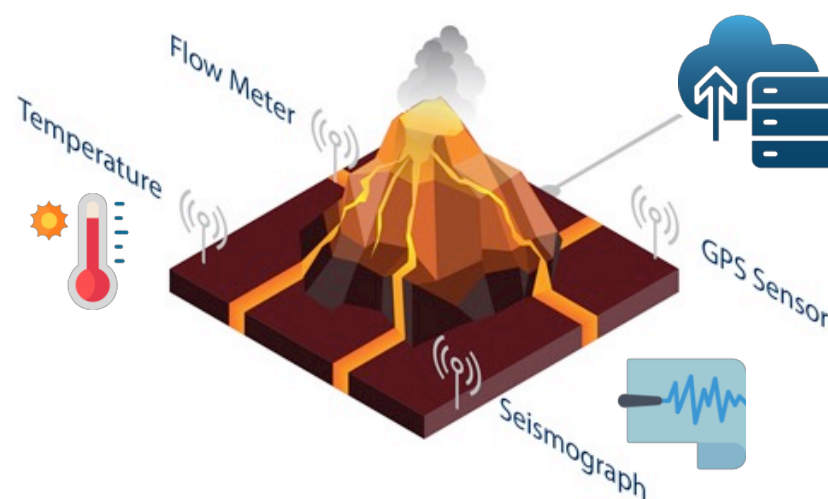
Smart Pet Care



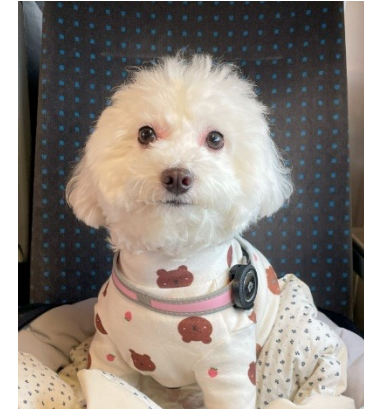
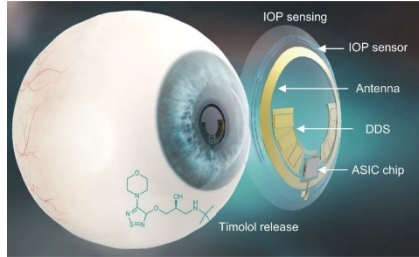
Smart Factory



Smart Environment

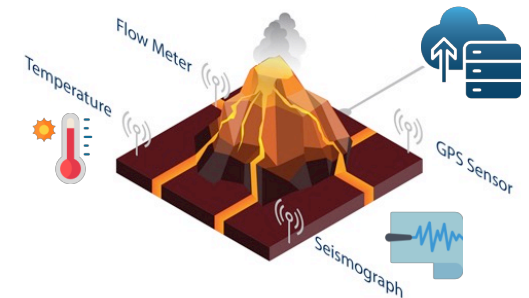


Era of Tiny IoT Platforms



In the next 10 years,

1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless
IoT platforms will enrich our daily lives!



Example 1: Bio-Signal Sensing

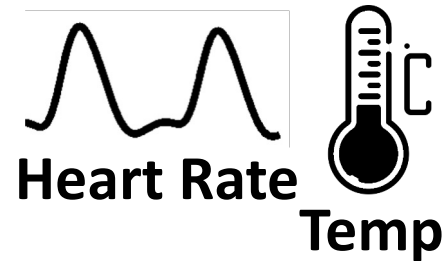
1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless



OURA



PCB of Oura Ring 2

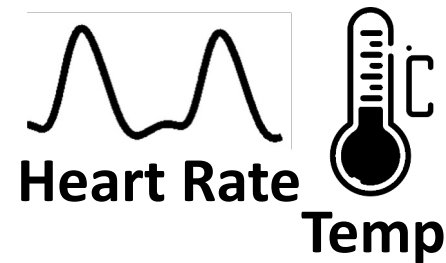


Example 1: Bio-Signal Sensing

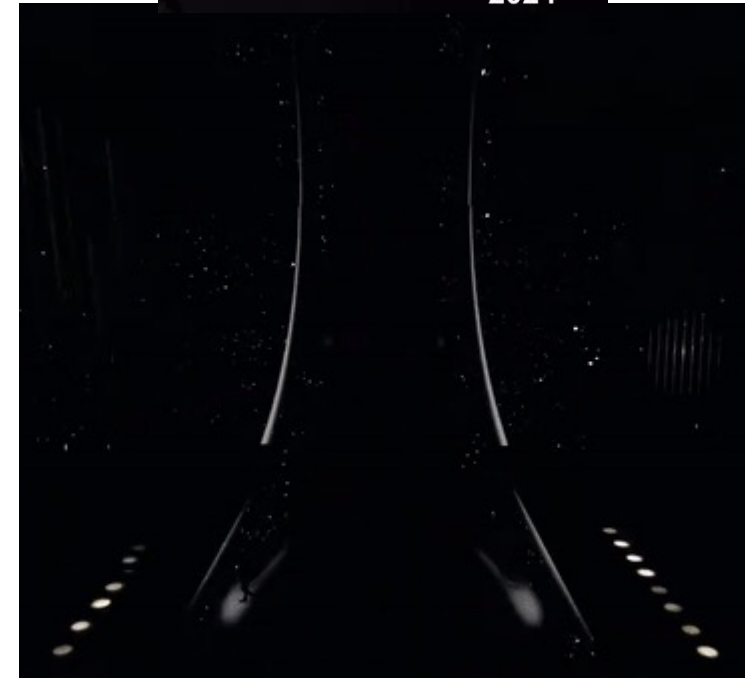
1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless



OURA

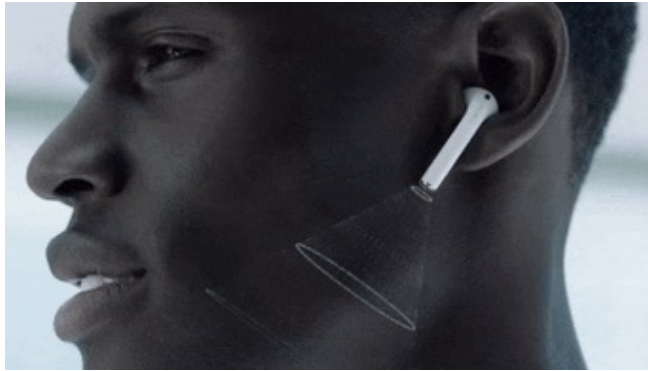


SAMSUNG Galaxy Ring 2024



Example 2: Audio Sensing

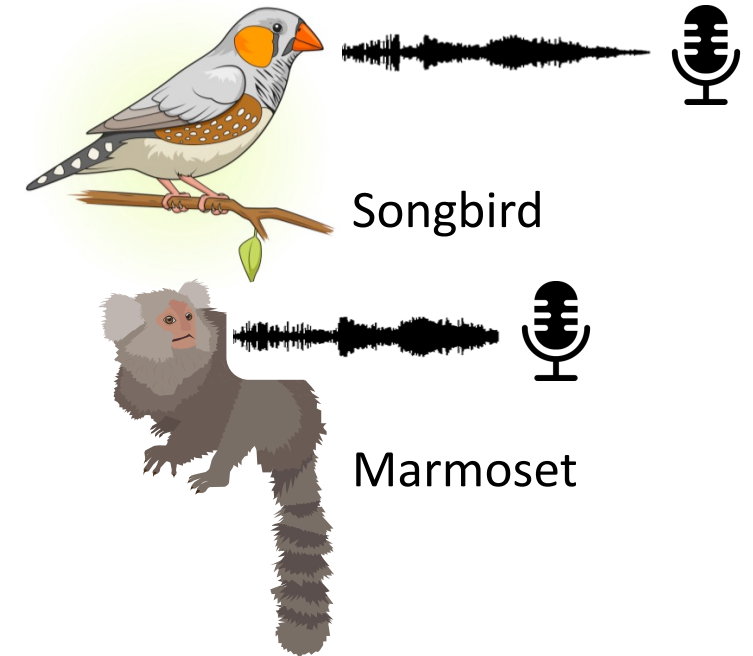
1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless



Earbuds



Smart Home

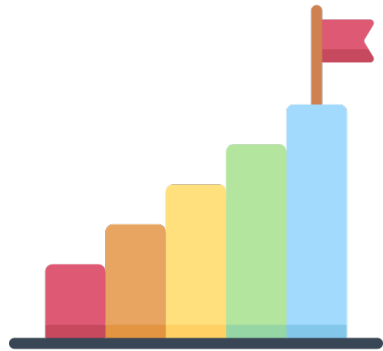


Songbird

Marmoset

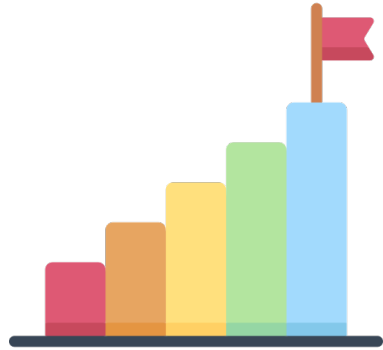
Vocal Studies of Animals

Ultimate Research Goal

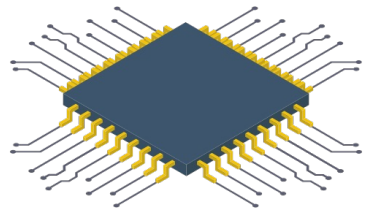
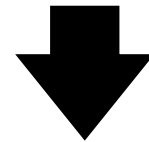


Democratizing the IoT platform with
1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless
features to enrich our daily lives

Ultimate Research Goal



Democratizing the IoT platform with
1) Tiny + 2) Sensory + 3) Intelligent + 4) Wireless
features to enrich our daily lives



We need **Highly Integrated System-on-Chip (SoC)**
under **Limited Energy Sources** and **Limited Silicon Area**

Outline

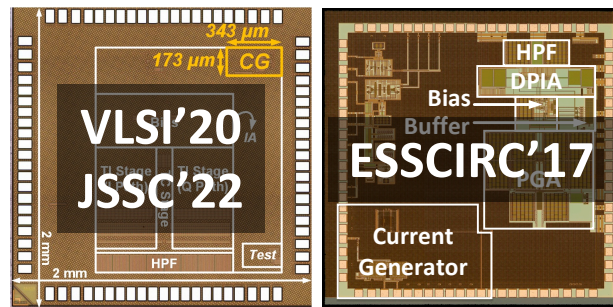
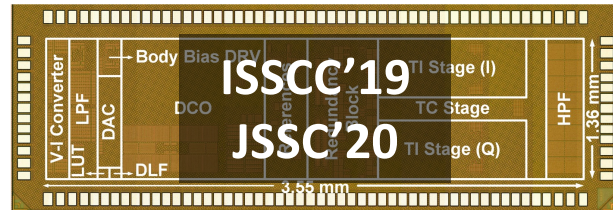
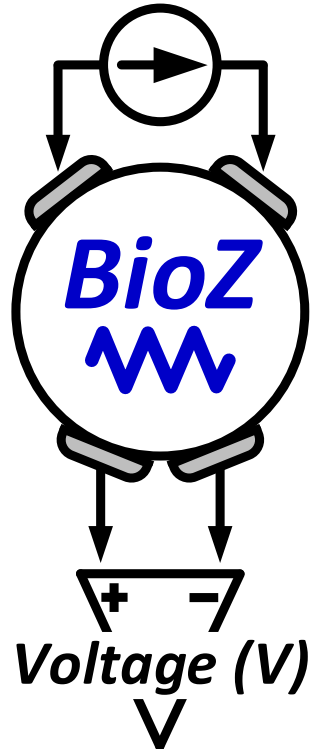
- Visions of IoT Sensors
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- **Neuromorphic Sensor**
- Outlook

Biomedical Sensor IC

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Ohm's Law
 $V = IR$ or $R = V/I$

Current (I)

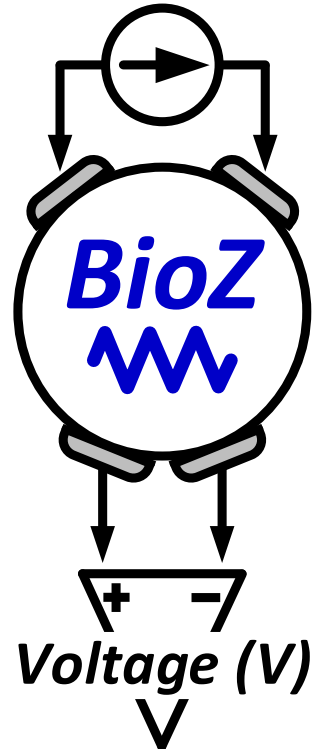


Biomedical Sensor IC

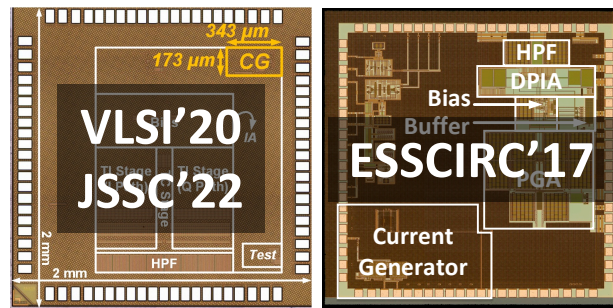
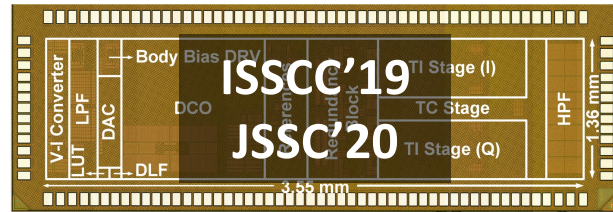
$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Ohm's Law
 $V = IR$ or $R = V/I$

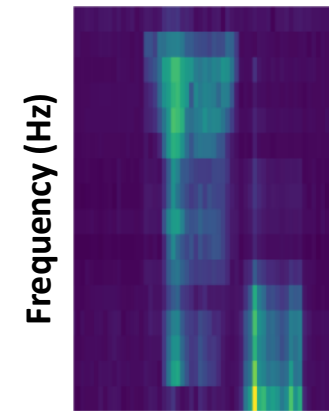
Current (I)



Voltage (V)

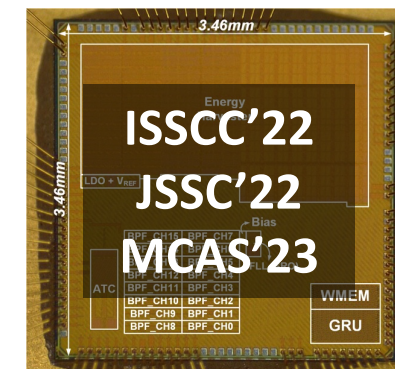


Neuromorphic Sensor IC



Score

"Yes"	0.95
"Silence"	0.02
"Unknown"	0.01
"Down"	0.01
"Go"	0.00
⋮	

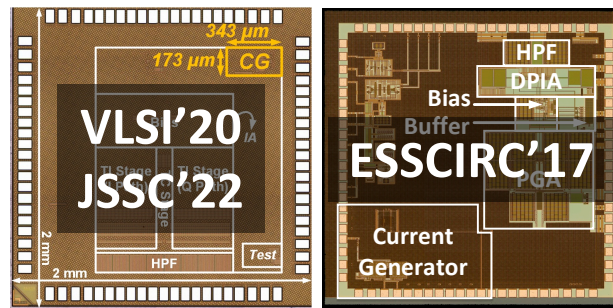
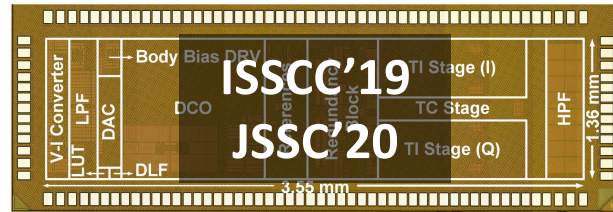
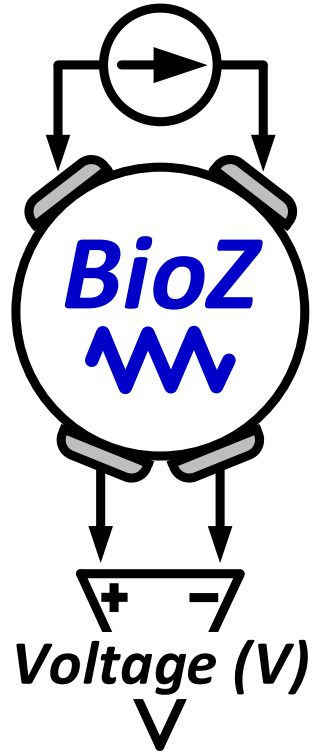


Biomedical Sensor IC

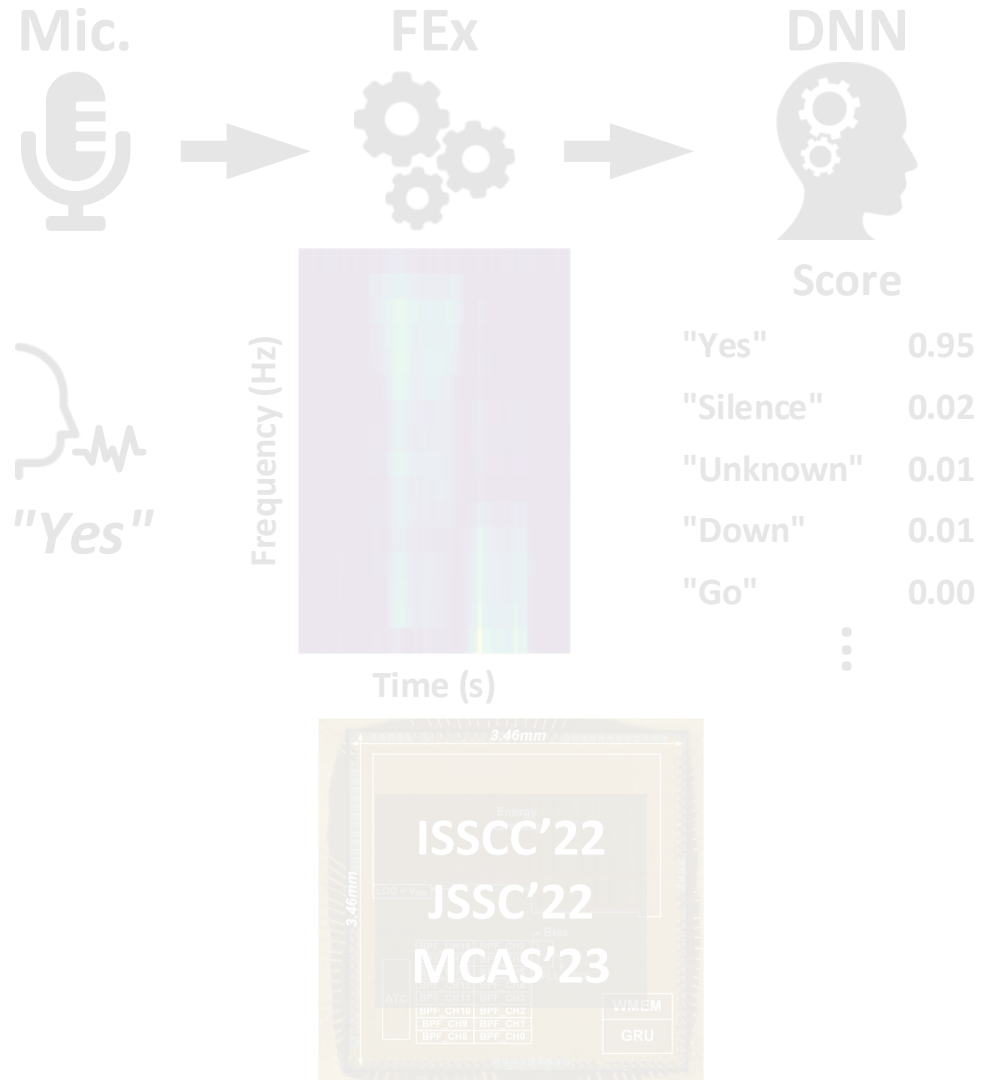
$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Ohm's Law
 $V = IR$ or $R = V/I$

Current (I)



Neuromorphic Sensor IC



Bioimpedance (BioZ) Sensing

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



Voltage (V)

V

Very Tiny Amplitude
$100\text{m}\Omega_{\text{RMS}}$ Resolution¹

Even Finer Resolution
is Needed!



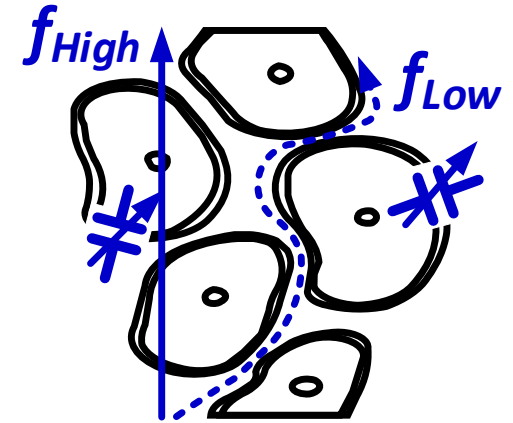
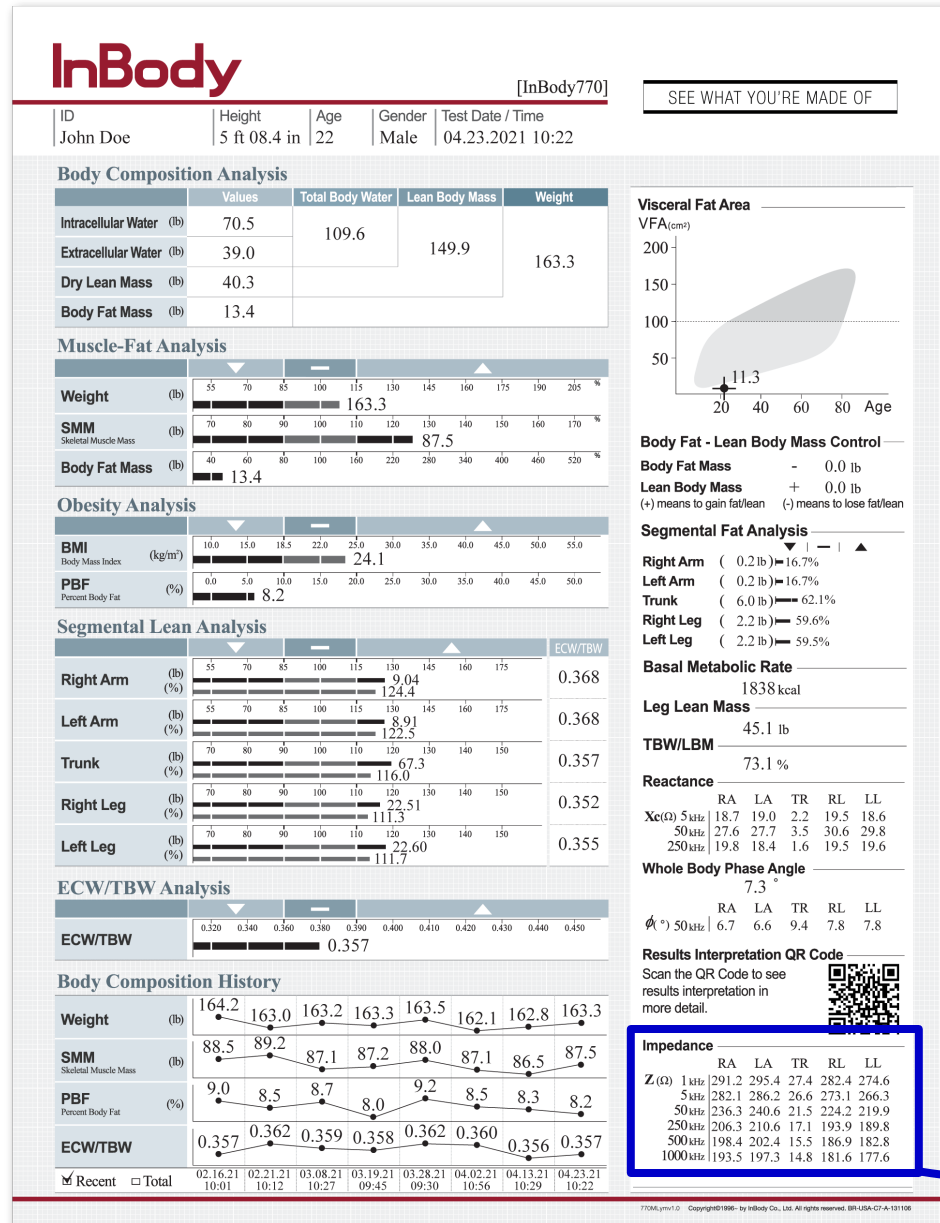
Bioimpedance (BioZ) Sensing

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



Voltage (V)



Impedance

	RA	LA	TR	RL	LL
Z (Ω) 1 kHz	291.2	295.4	27.4	282.4	274.6
5 kHz	282.1	286.2	26.6	273.1	266.3
50 kHz	236.3	240.6	21.5	224.2	219.9
250 kHz	206.3	210.6	17.1	193.9	189.8
500 kHz	198.4	202.4	15.5	186.9	182.8
1000 kHz	193.5	197.3	14.8	181.6	177.6

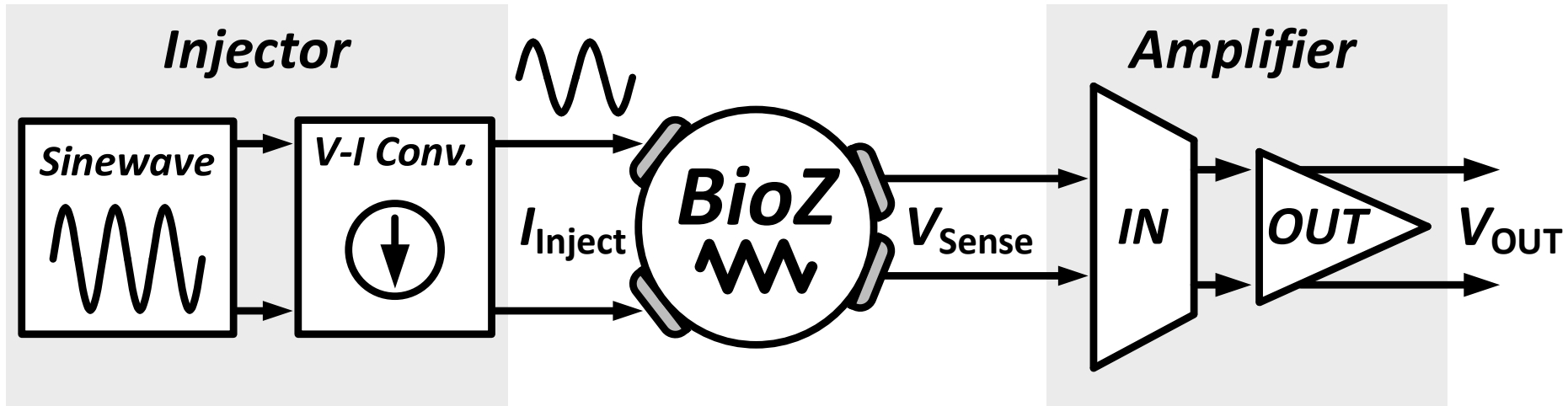
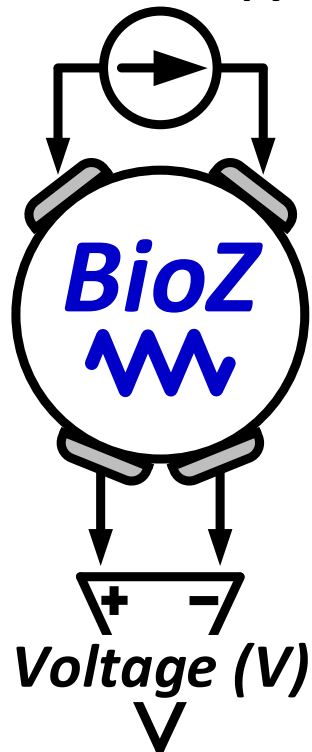
Impedance

Z (Ω)	RA	LA	TR	RL	LL
1 kHz	291.2	295.4	27.4	282.4	274.6
5 kHz	282.1	286.2	26.6	273.1	266.3
50 kHz	236.3	240.6	21.5	224.2	219.9
250 kHz	206.3	210.6	17.1	193.9	189.8
500 kHz	198.4	202.4	15.5	186.9	182.8
1000 kHz	193.5	197.3	14.8	181.6	177.6

Bioimpedance (BioZ) Sensing

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



Current Injector Needs:

Pure Sinewave

(Total Harmonic Distortion (THD) < 1%)

Low Power (<10 μ W)

Low Area (Smaller is Better)

Voltage Amplifier Needs:

Low Noise (<100nV/ $\sqrt{\text{Hz}}$)

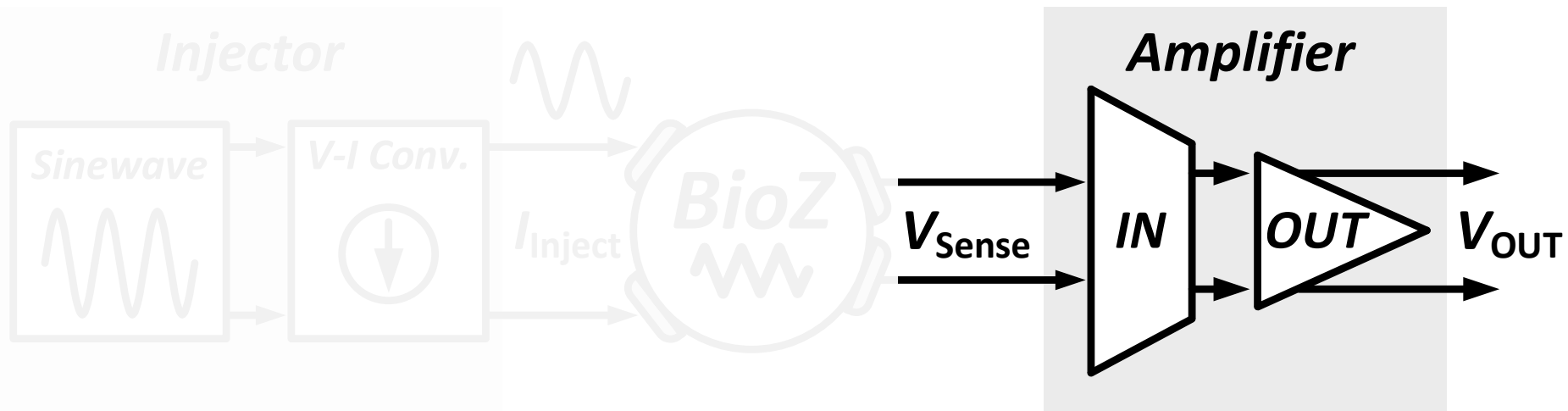
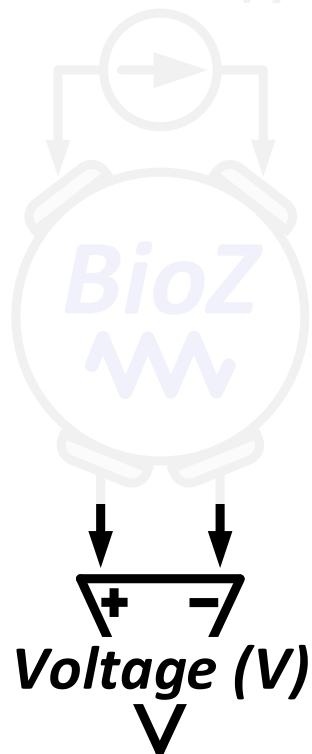
Wide Bandwidth (>100kHz)

Low Power (<10 μ W)

Bioimpedance (BioZ) Sensing

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



Current Injector Needs:

Pure Sinewave

(Total Harmonic Distortion (THD) < 1%)

Low Power (<10μW)

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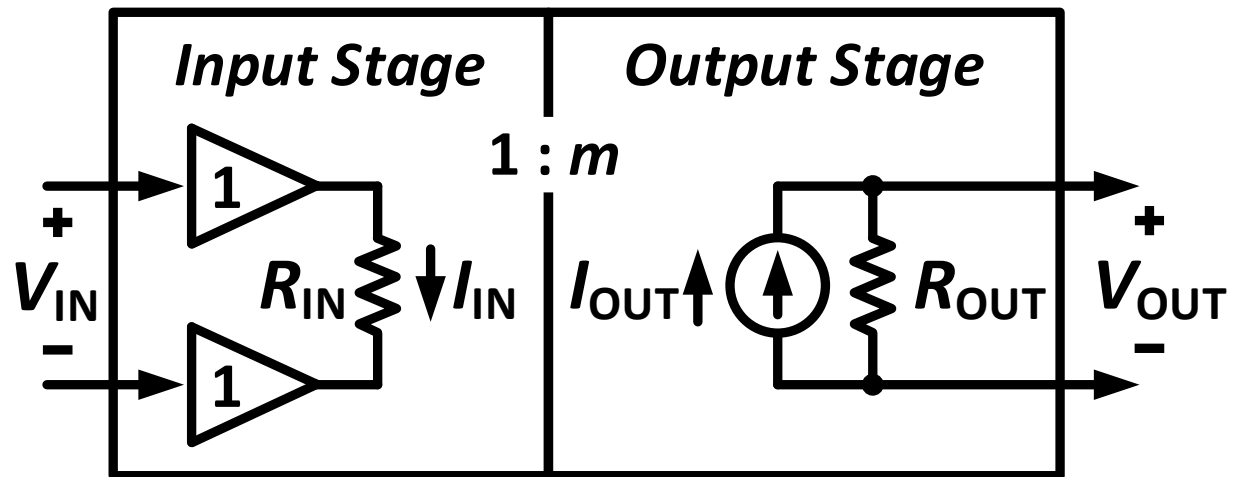
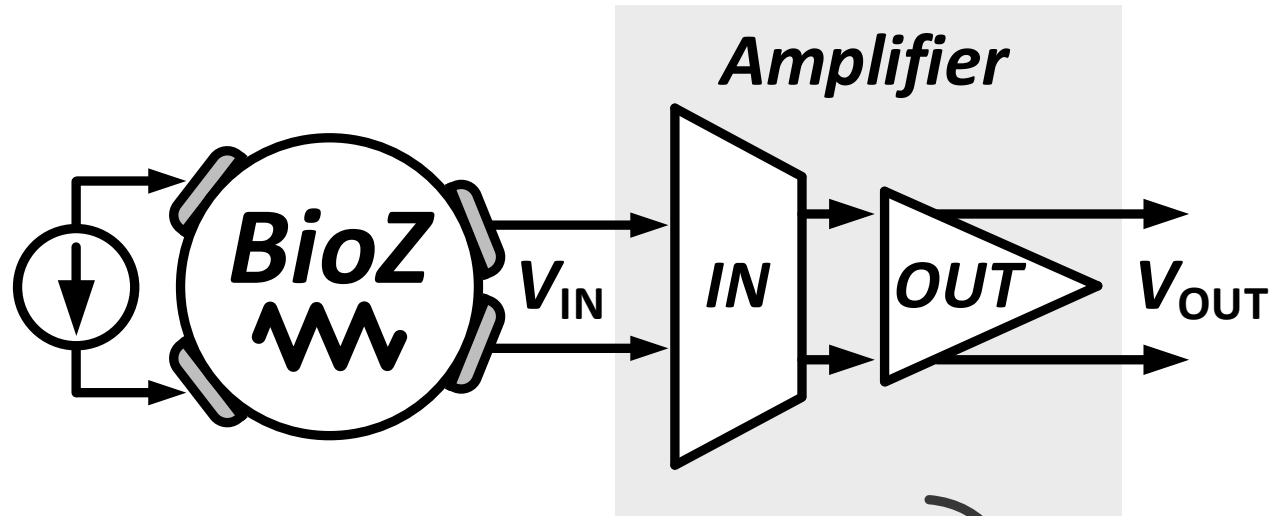
Voltage Amplifier Needs:

Low Noise (<100nV/√Hz)

Wide Bandwidth (>100kHz)

Low Power (<10μW)

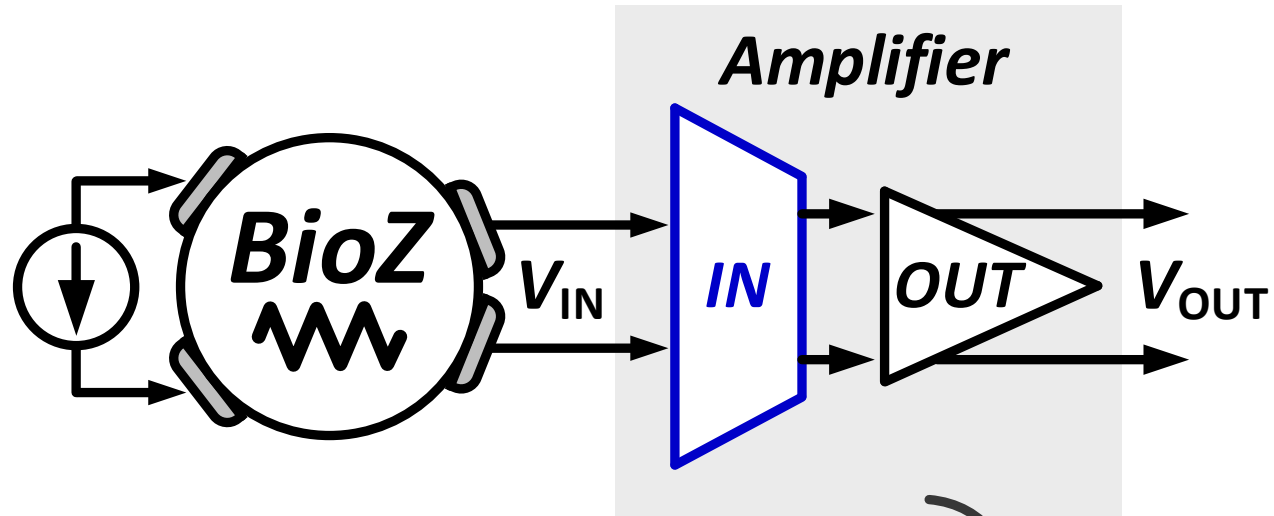
Voltage Readout Amplifier



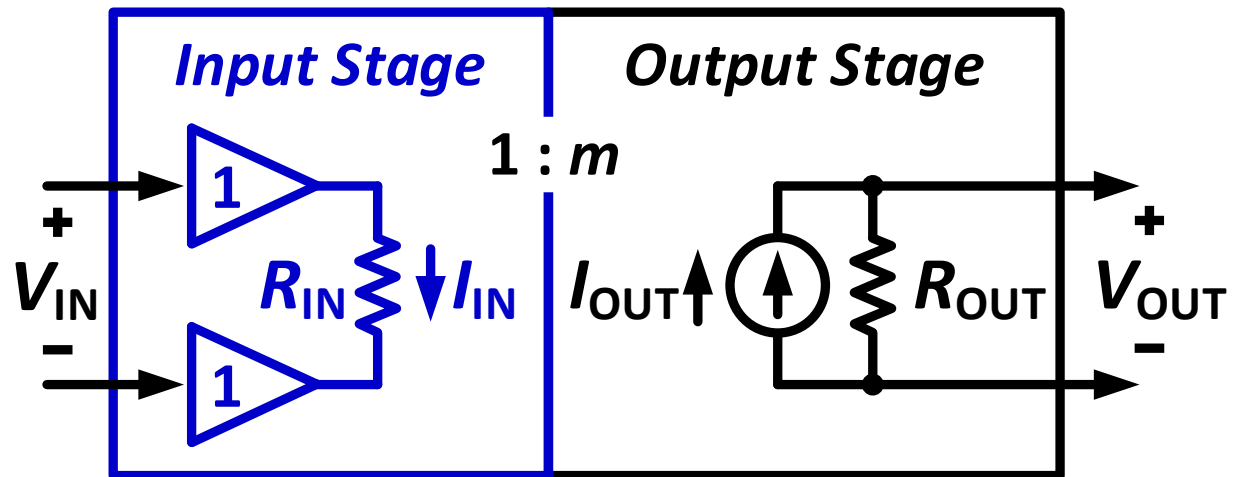
Amplifier Gain

$$A_v = m \frac{R_{OUT}}{R_{IN}}$$

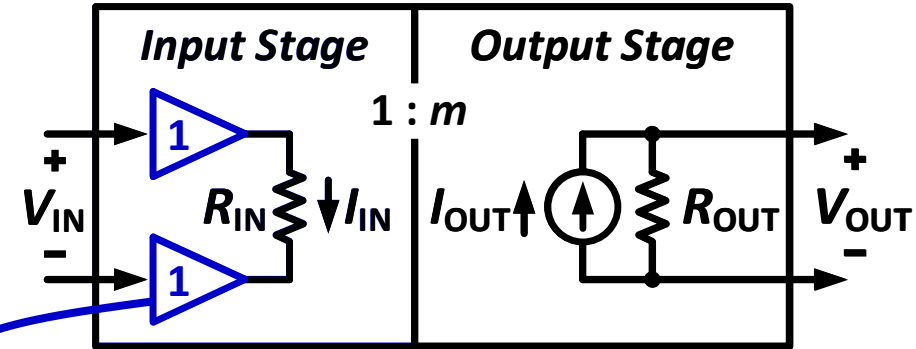
Voltage Readout Amplifier



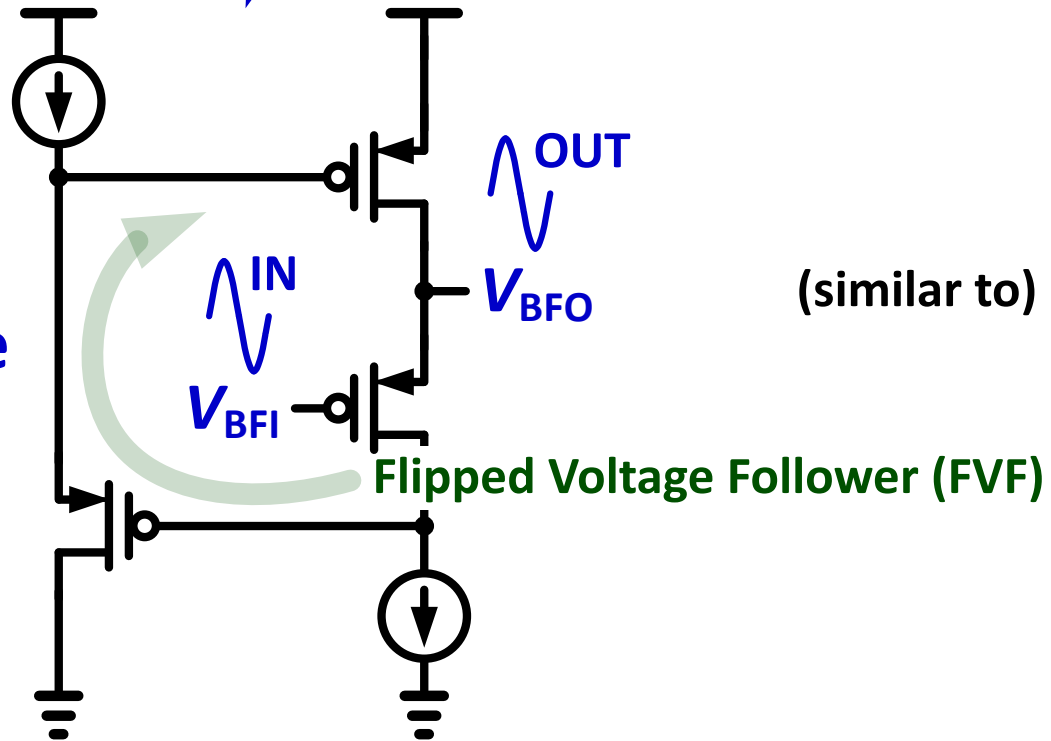
**Input Stage Determines
Noise & BW
Performance!**



Voltage Buffer

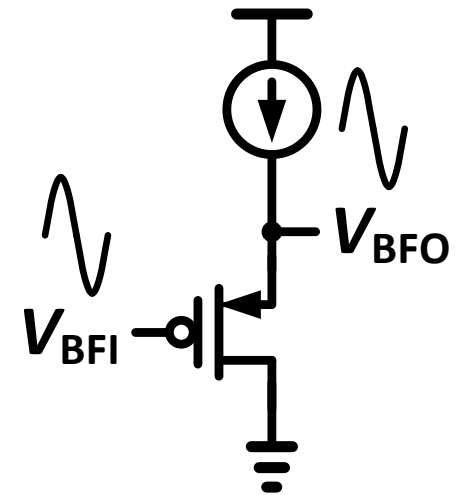


Buffer Stage

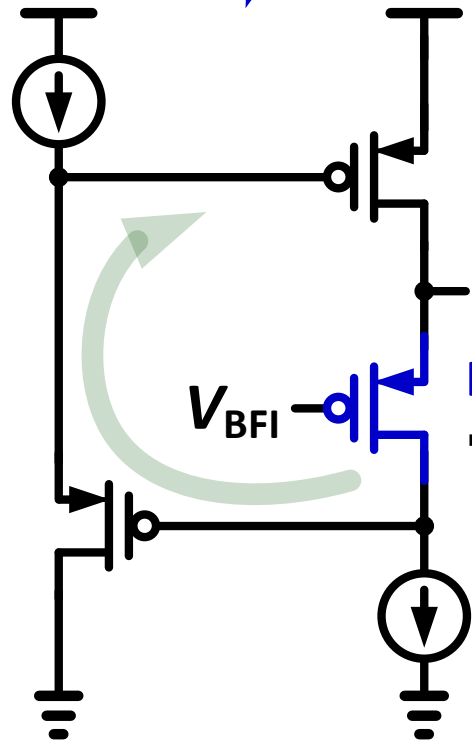
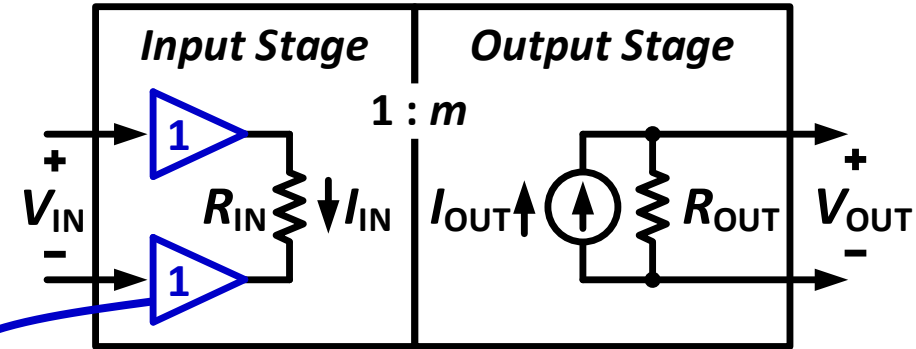


Flipped Voltage Follower (FVF)

Source Follower

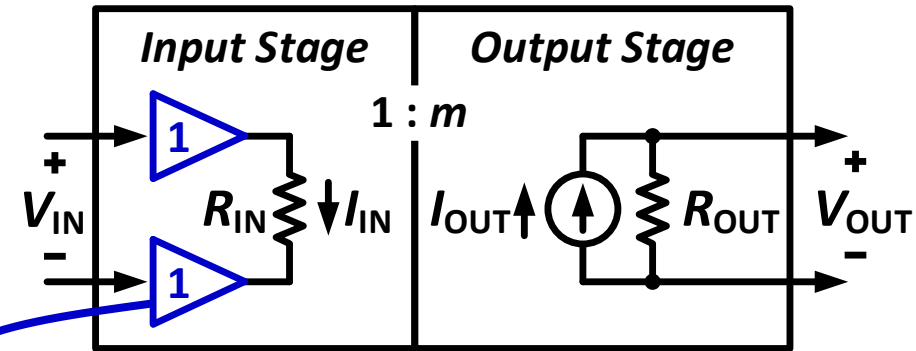


Voltage Buffer: *Low-Noise* Design



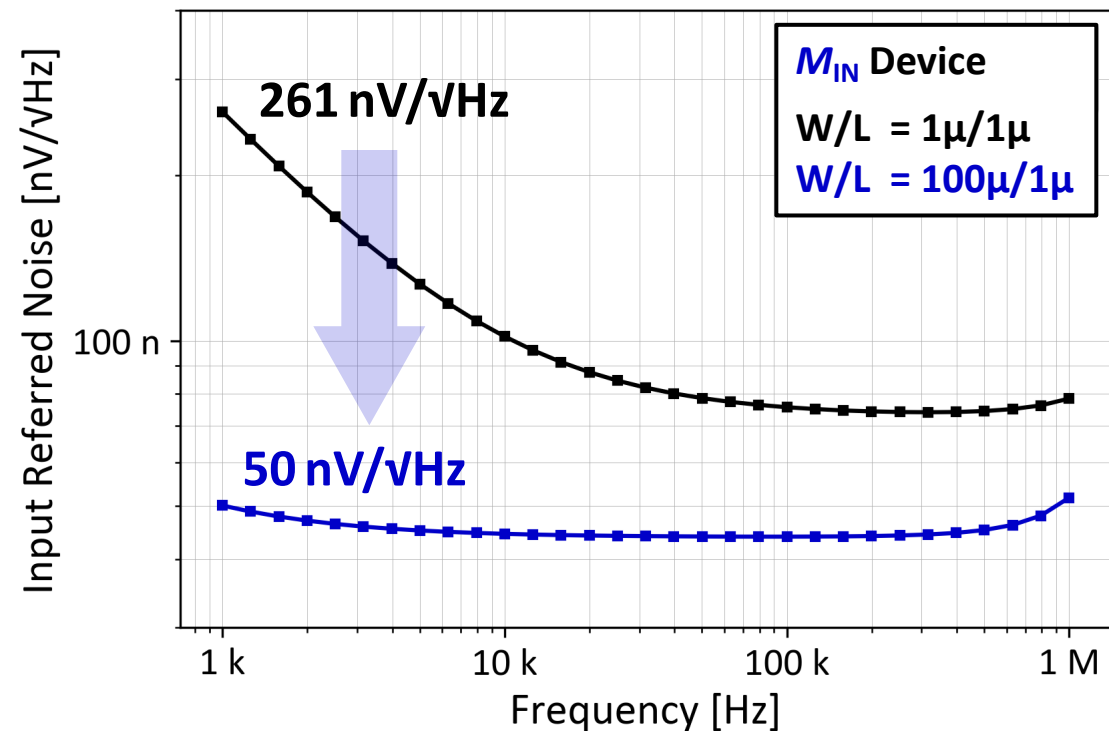
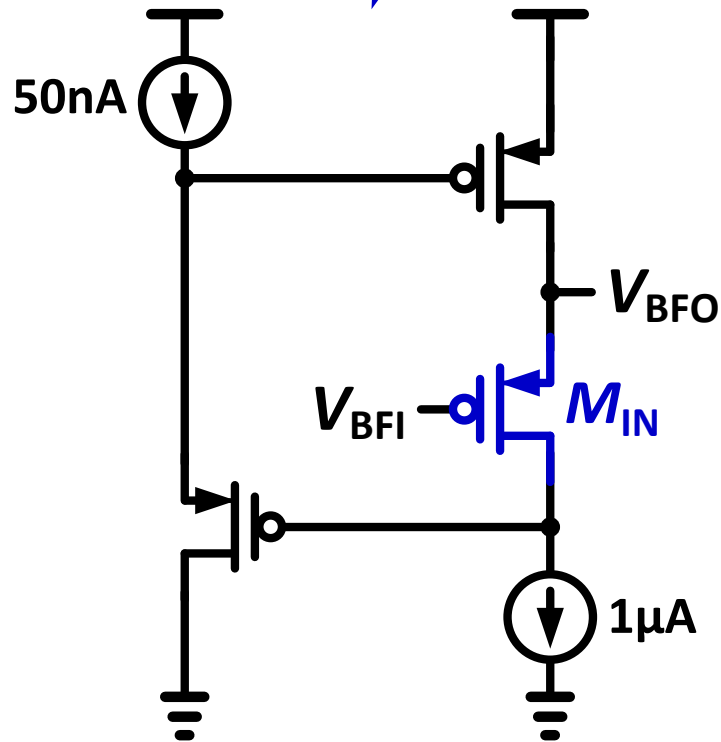
Input Transistor is the Main Signal Driver
 → The Main **Noise** Contributor!

Voltage Buffer: *Low-Noise* Design

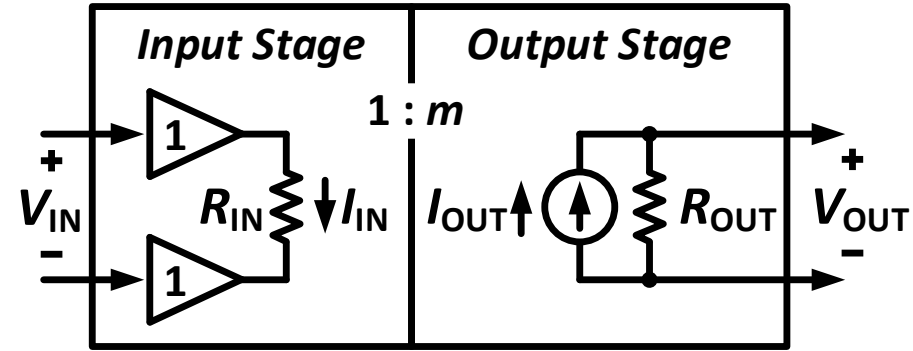


$$V_n^2(f) = \frac{8 kT}{3 g_m} + \frac{K}{C_{ox} W L} \frac{1}{f} \quad [V^2/\text{Hz}]$$

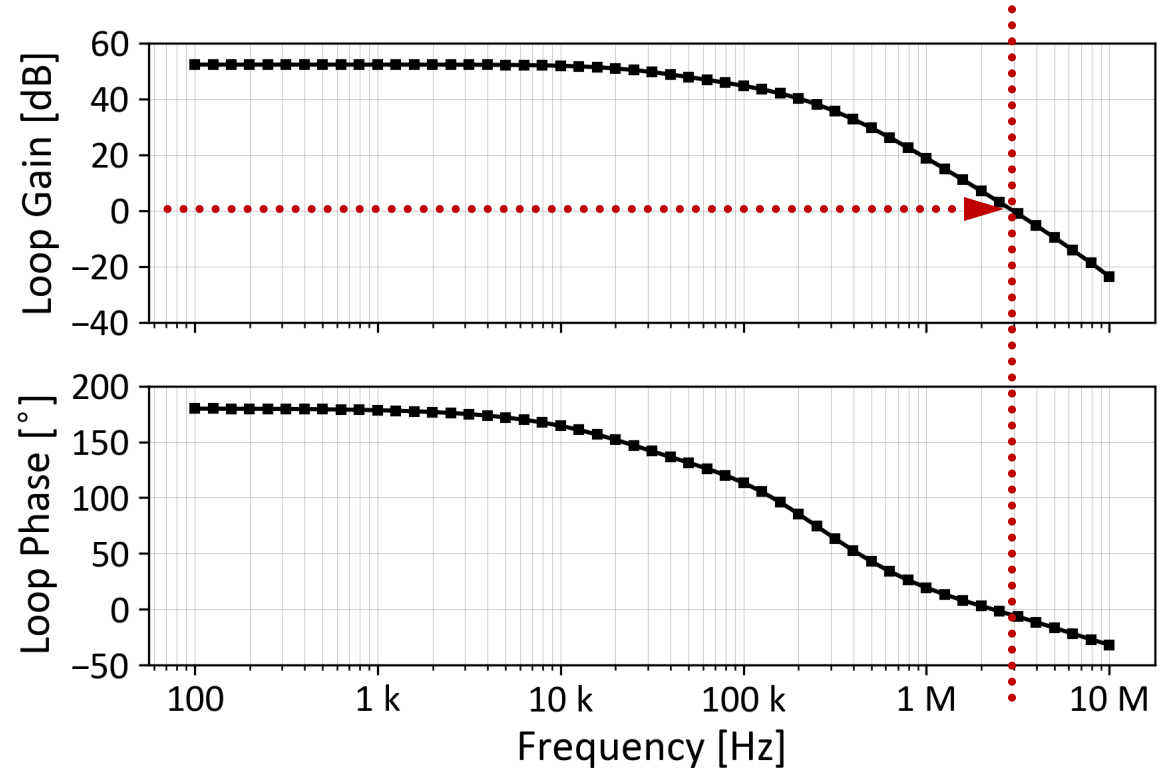
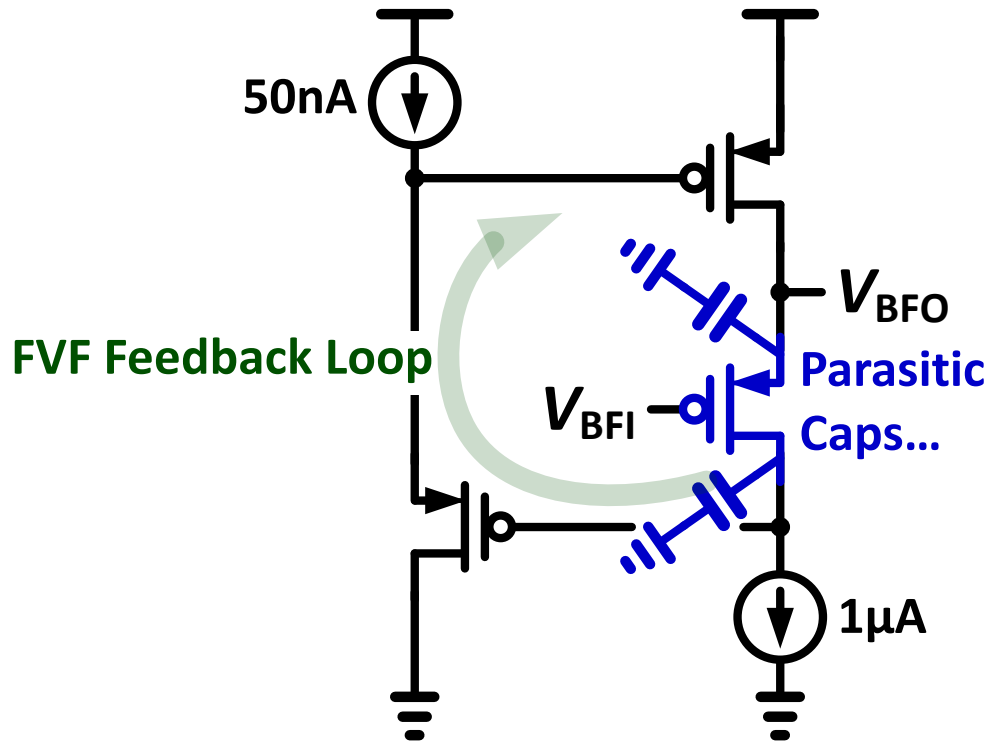
Thermal Flicker



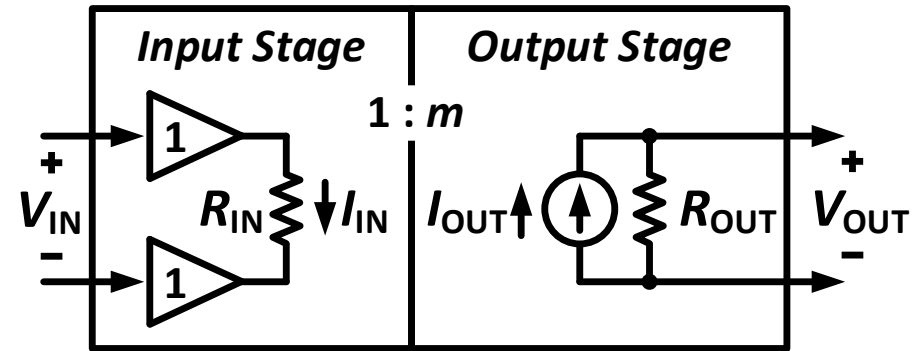
Voltage Buffer: *Feedback Stability*



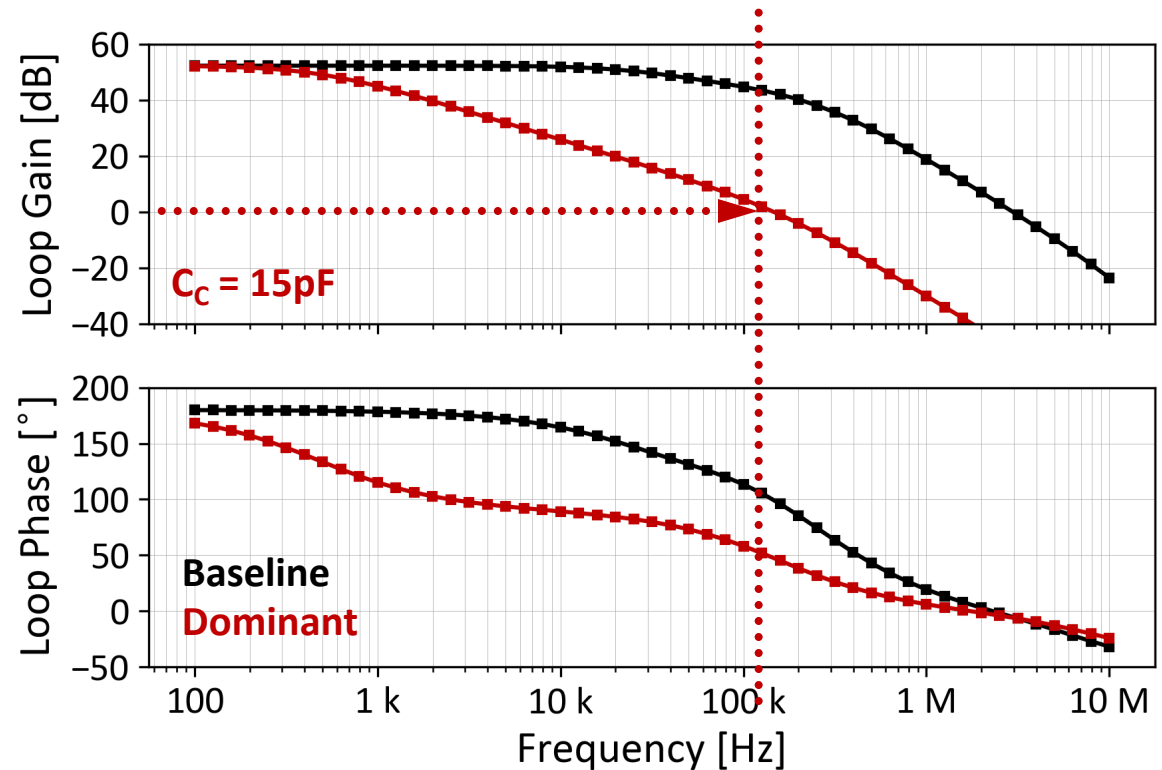
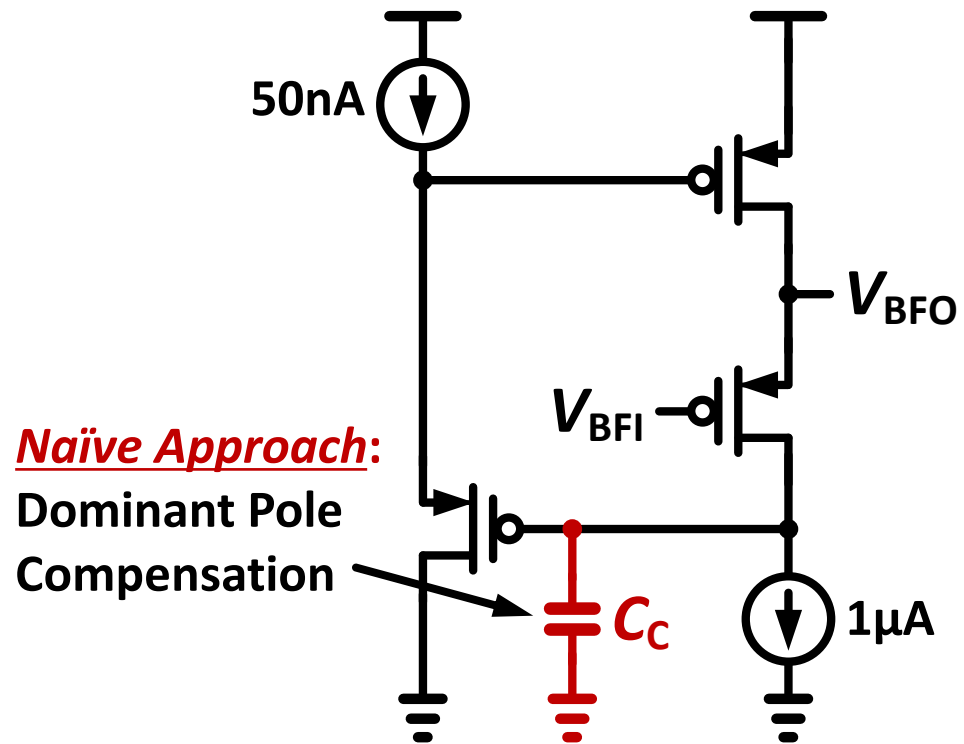
Phase Margin = -5.28°
 → Unstable!



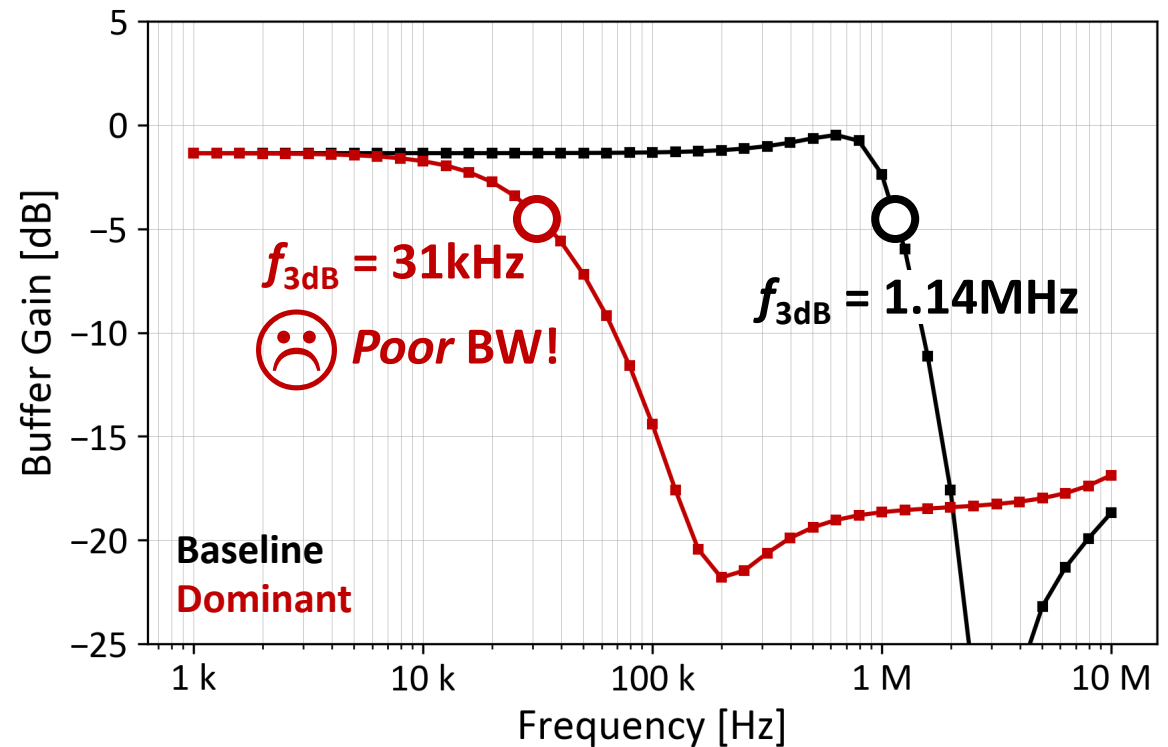
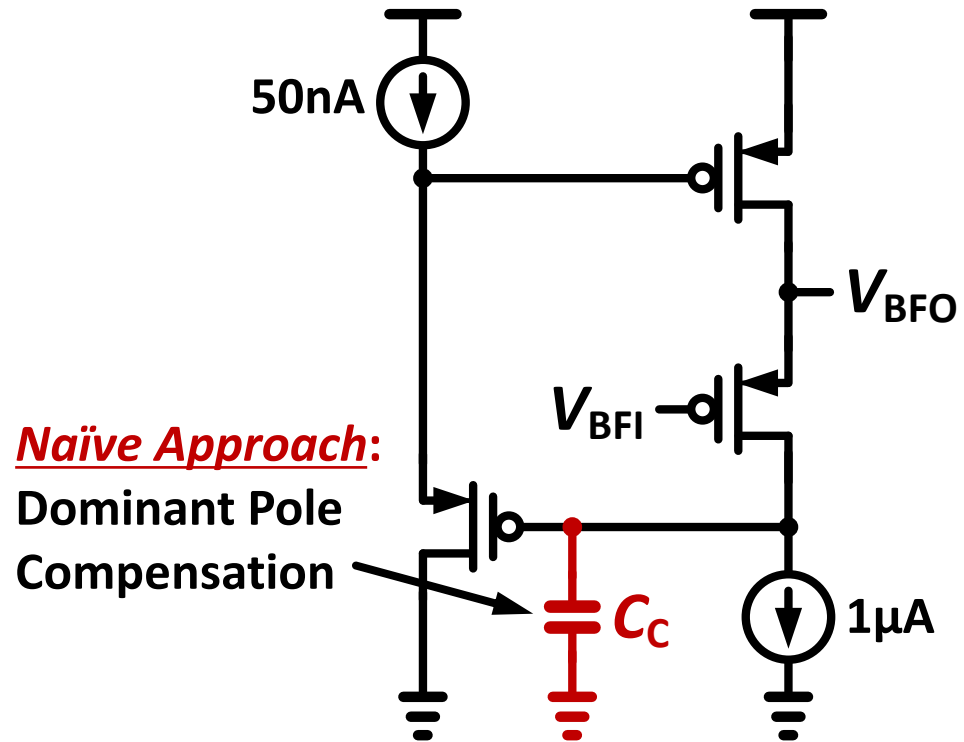
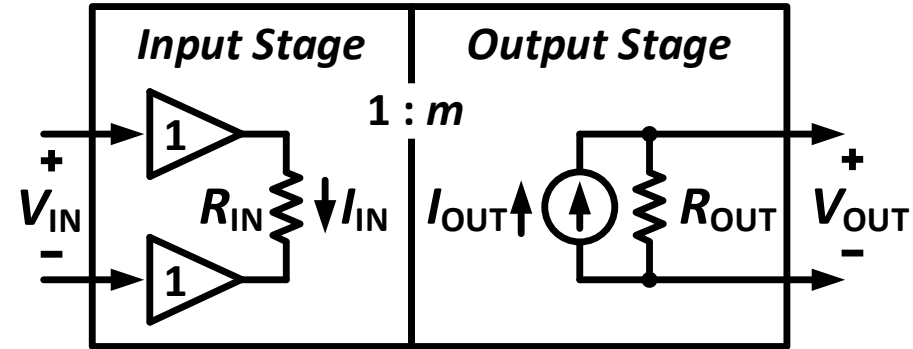
Voltage Buffer: *Feedback Stability*



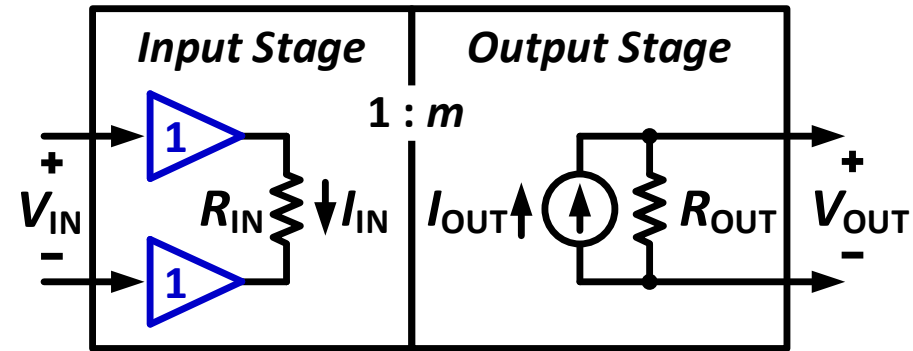
Phase Margin = 47°
 → This is okay, but ...



Voltage Buffer: *Bandwidth Problem*



Voltage Buffer



Design Dilemma:

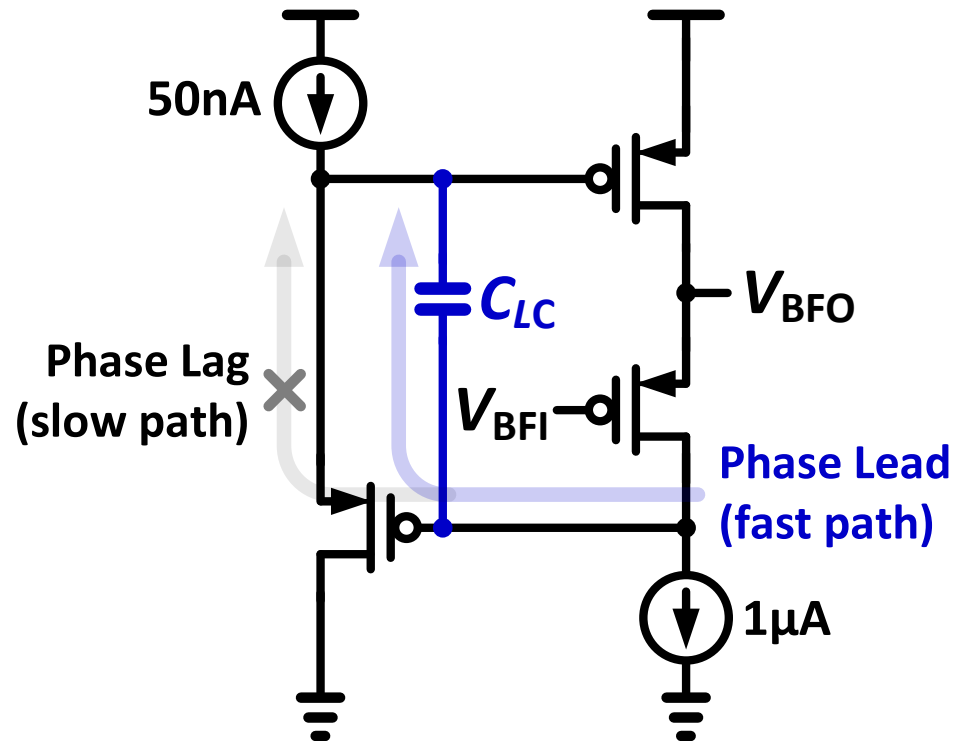
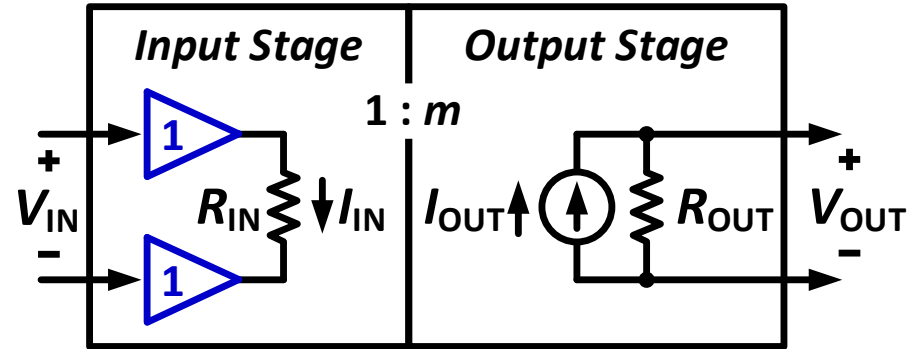
✓ **Low Noise** ($<100\text{nV}/\sqrt{\text{Hz}}$)

✓ **Low Power** ($<10\mu\text{W}$)

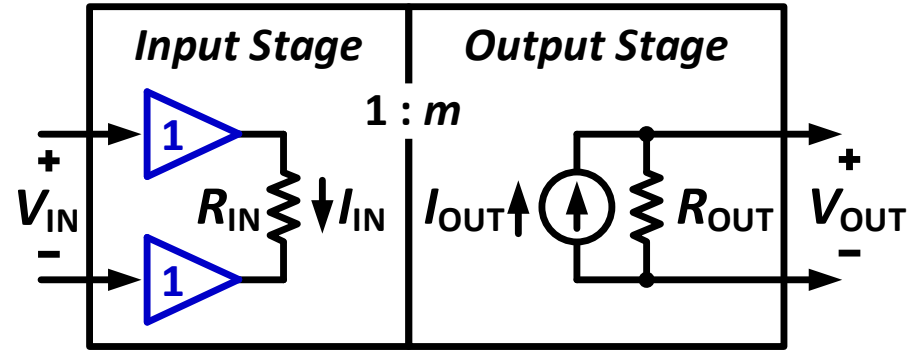
✓ **Stability** ($\text{PM} > 50^\circ$)

✗ **Bandwidth** ($>100\text{kHz}$)

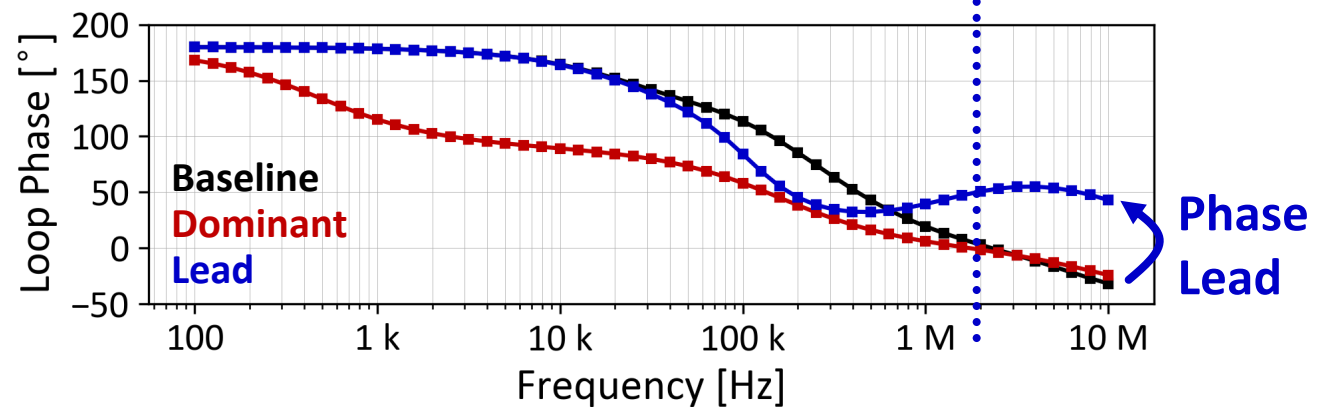
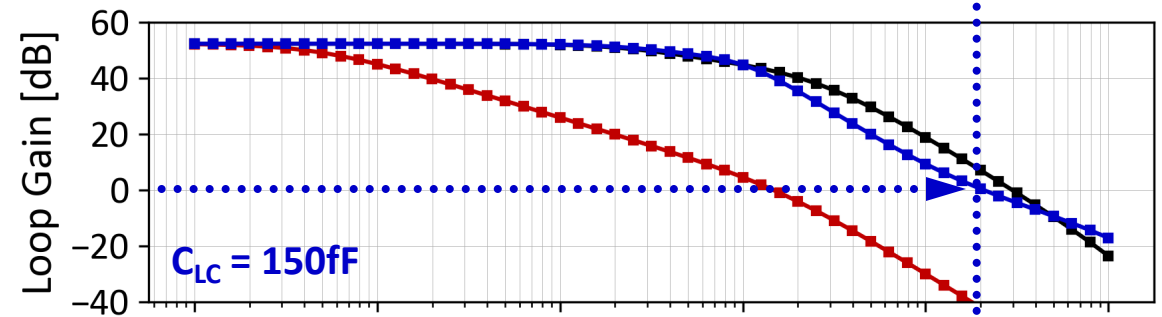
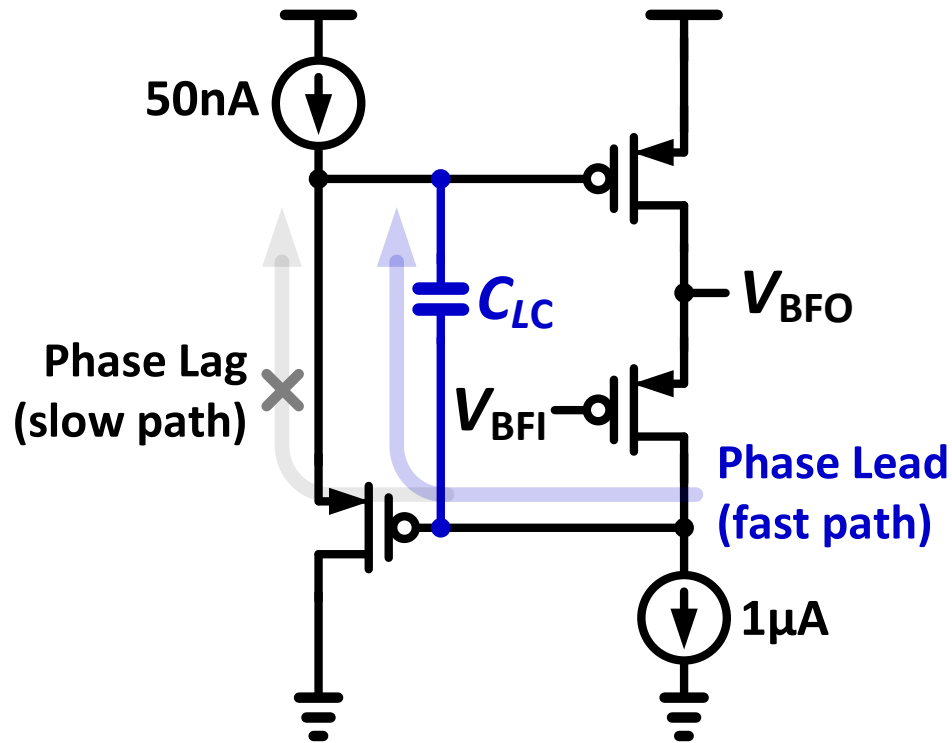
Voltage Buffer: *Lead Compensation (LC)*¹



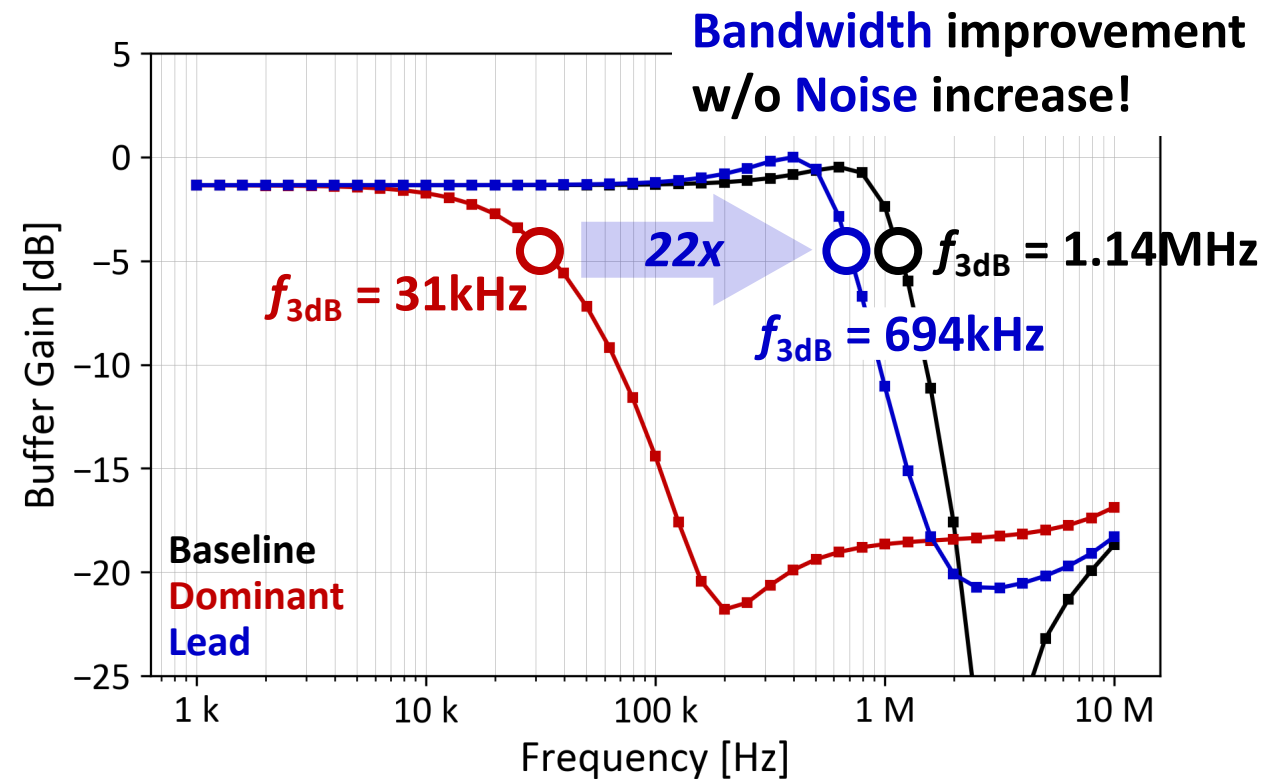
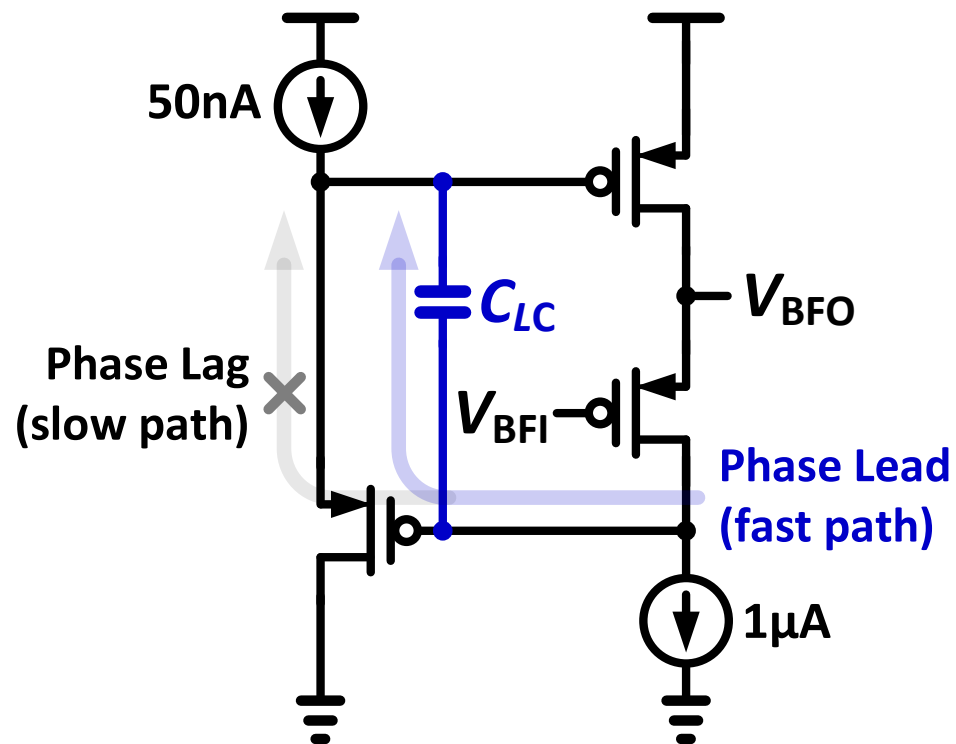
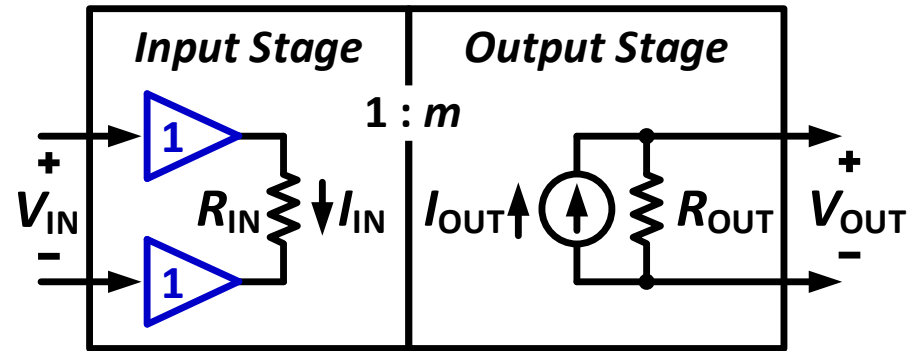
Voltage Buffer: *Lead Compensation (LC)*¹



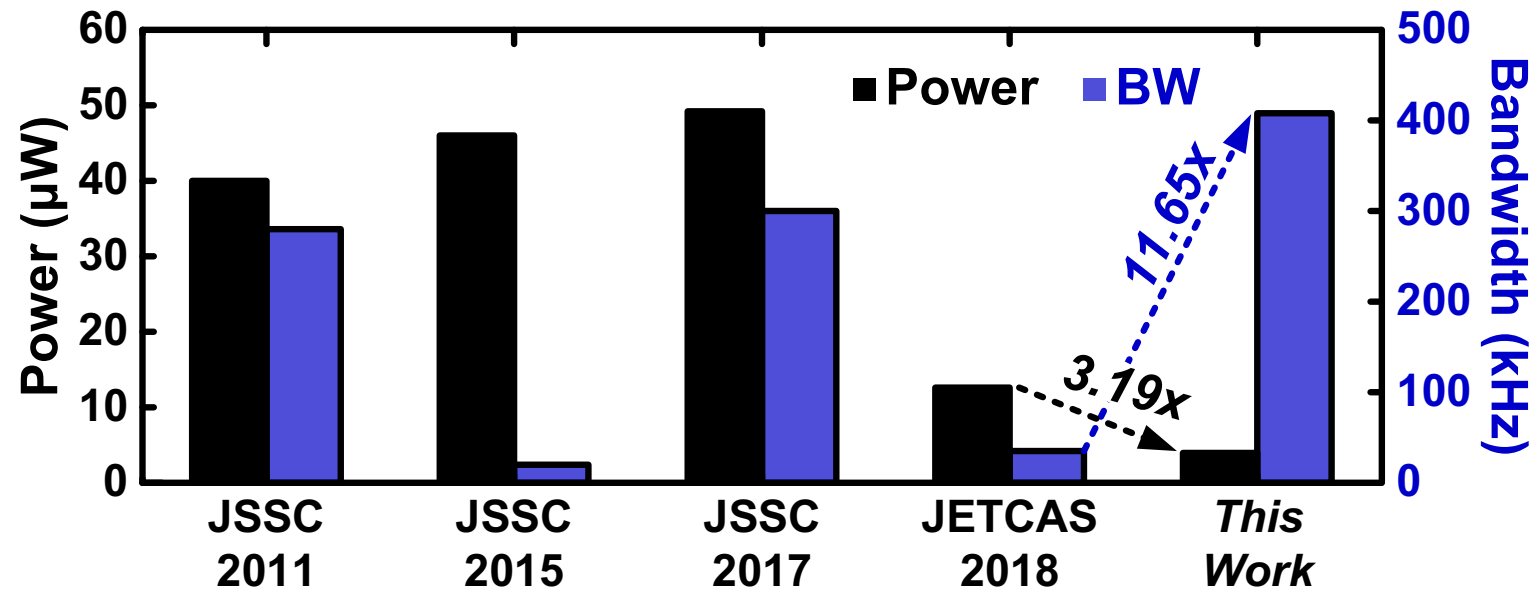
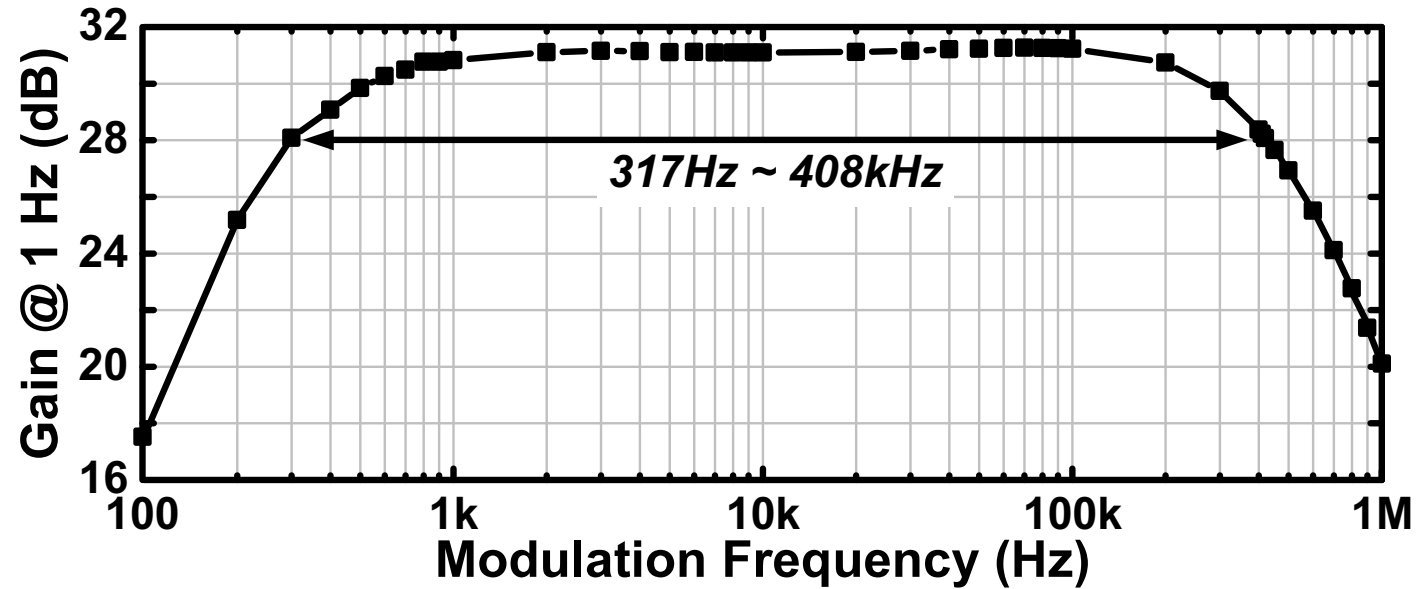
😊 Phase Margin = 51°
 → Stable!



Voltage Buffer: *Lead Compensation (LC)*¹



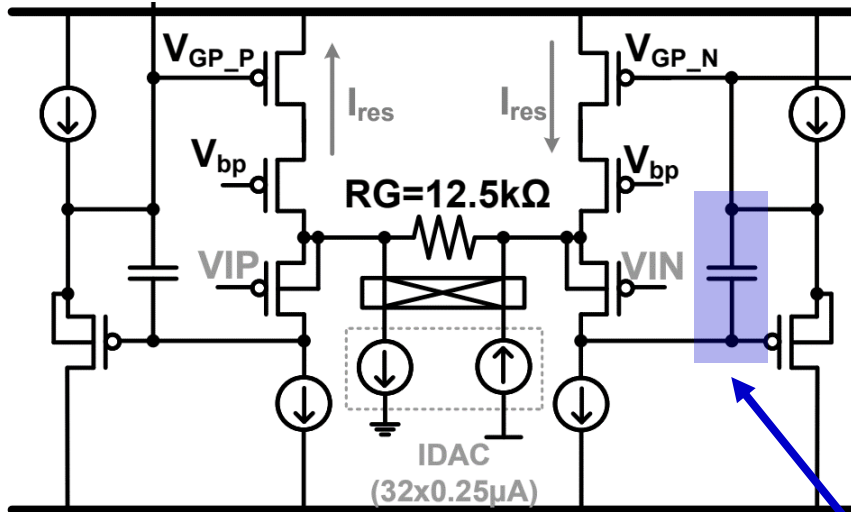
Power-BW-Efficient Amplifier¹



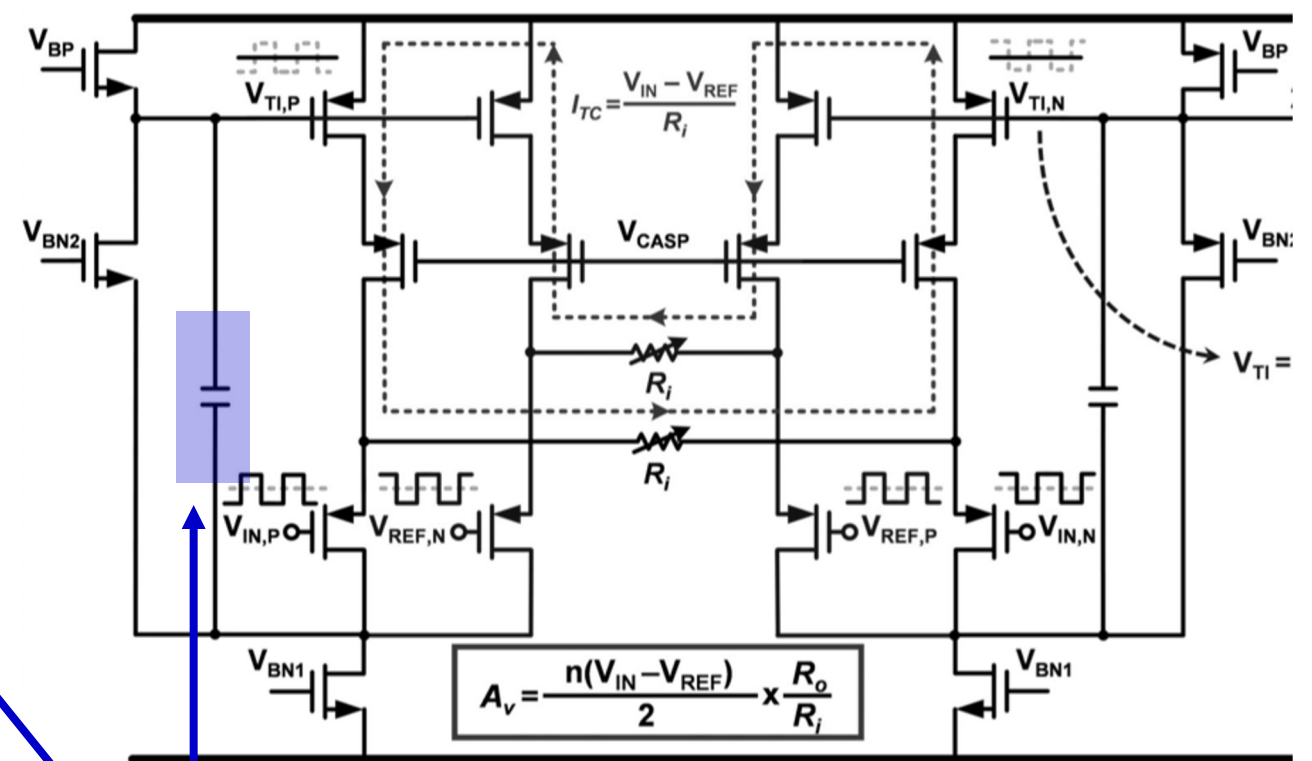
¹K. Kim, ISSCC 2019 + JSSC 2020

Adoption in Other Works

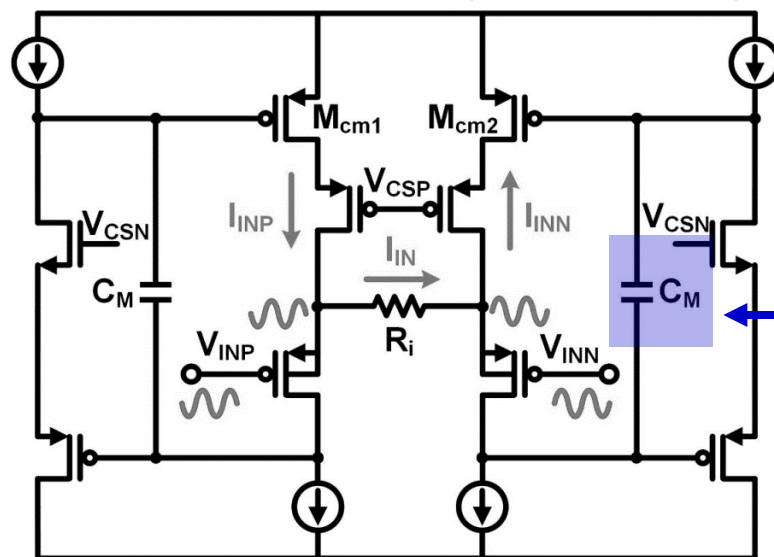
T. Zhang, ISSCC 2021 (A*STAR)



S.-I. Cheon, ISSCC 2024 (KAIST)



Q. Pan, ISSCC 2022 (Fudan Univ.)



Lead Compensated buffer design
has been **Widely Adopted!**

Open-Sourced Amplifier Design¹

A 3.11 μW 40 $\text{nV}/\sqrt{\text{Hz}}$ Instrumentation Amplifier for Bio-Impedance Sensors Exploiting Positive-Feedback-Assisted Gain Boosting

Kwantae Kim and Shih-Chii Liu

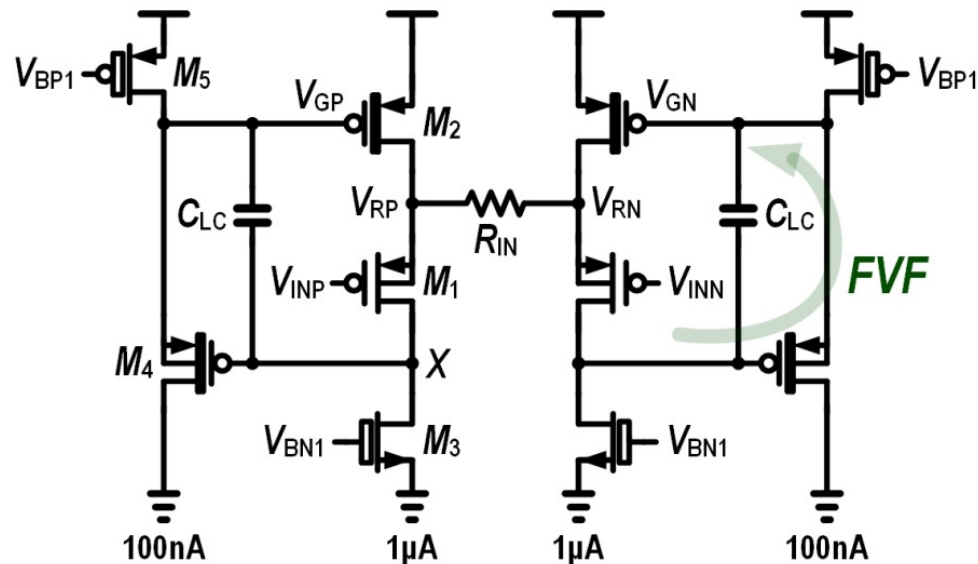
Institute of Neuroinformatics, University of Zürich and ETH Zürich, 8057 Zürich, Switzerland

Email: {kwantae, shih}@ini.uzh.ch

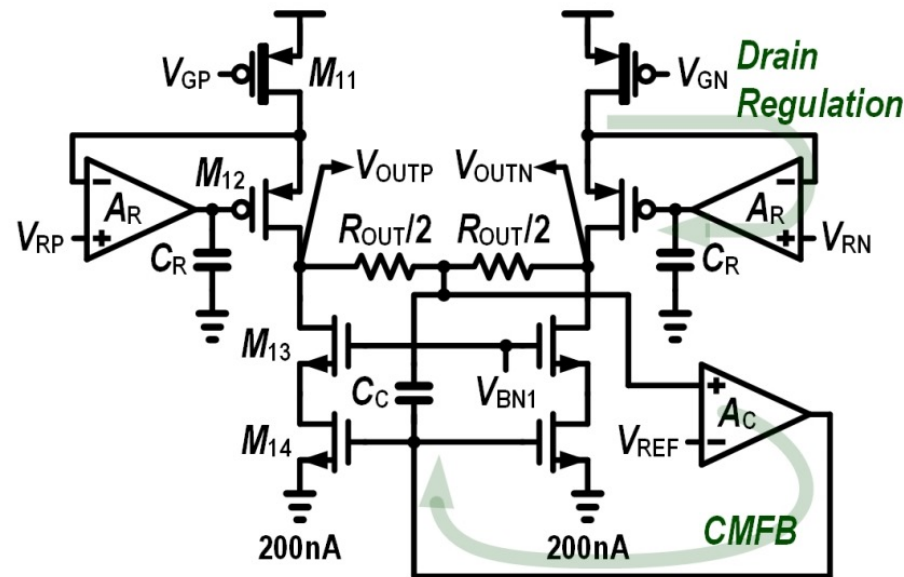
TABLE I
DESIGN PARAMETERS OF THE BASELINE IA

TC Stage	W/L (μm)	A_R Amplifier	W/L (μm)
M_1	500/0.25	M_{R1}	0.25/0.25
M_2	$5 \times (0.5/4)$	M_{R2}	0.25/0.25
M_3	5/16	M_{R3}	0.25/0.25
M_4	1/0.25	M_{R4}	0.5/16
M_5	0.25/4	C_R	5 fF
C_{LC}	100 fF		
TI Stage	W/L (μm)	A_C Amplifier	W/L (μm)
M_{11}	0.5/4	M_{C1}	0.25/0.25
M_{12}	1/0.25	M_{C2}	0.25/0.25
M_{13}	0.5/0.5	M_{C3}	0.25/0.25
M_{14}	0.25/16	M_{C4}	0.25/4
C_C	100 fF	V_{REF}	0.5 V
R_{OUT}	$500 \times R_{IN}$		

< Transconductance (TC) >



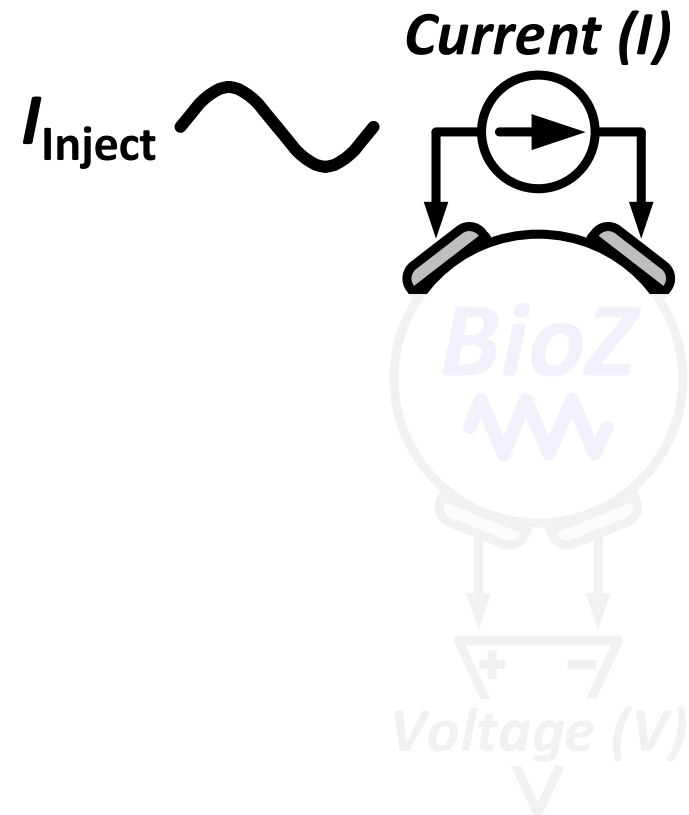
< Transimpedance (TI) >



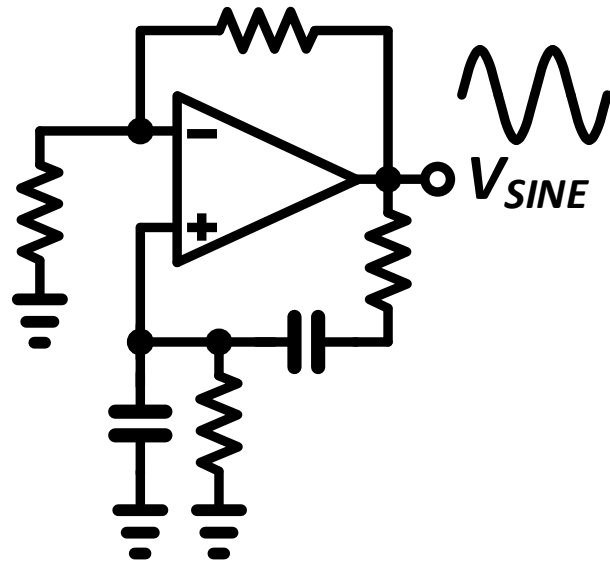
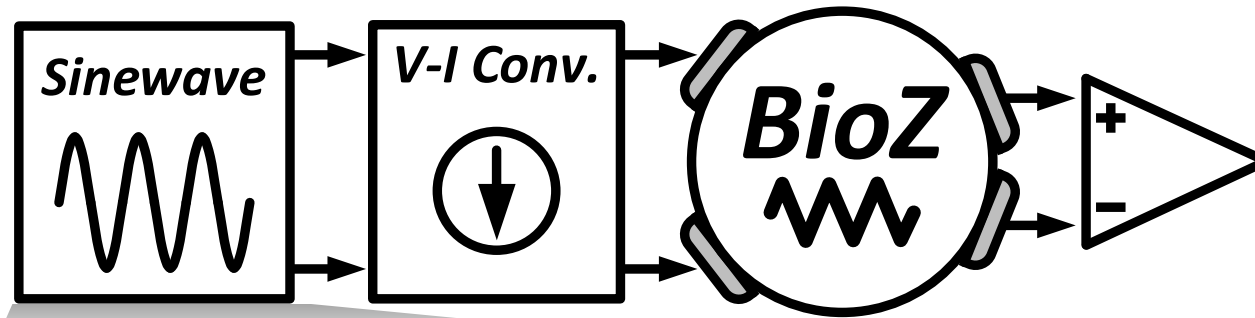
Open-Sourced
Schematic Parameters
(TSMC 65nm)

Bioimpedance (BioZ) Sensing

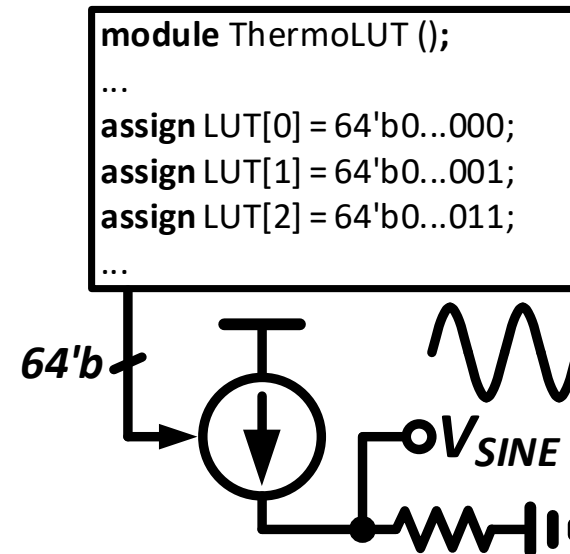
$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$



Sinewave Generation



Analog RC-OSC¹

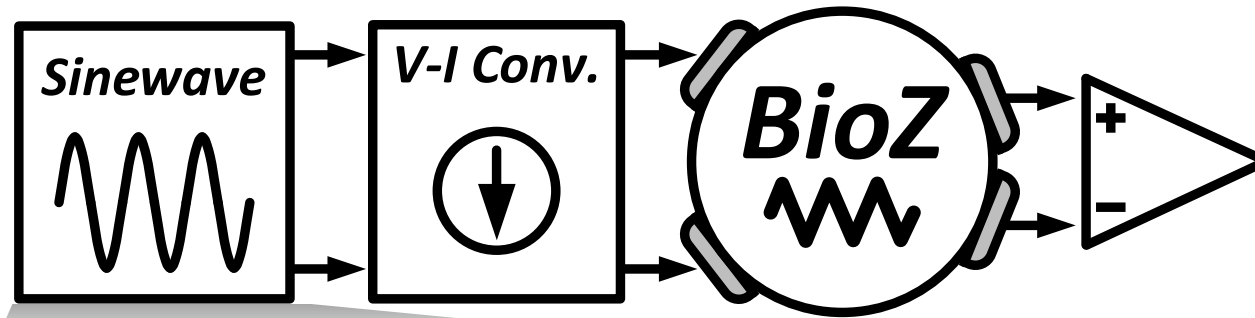


Digital LUT
+ Analog DAC²

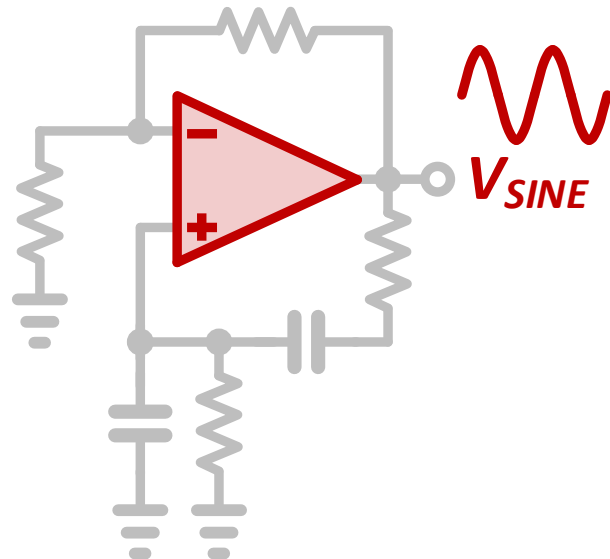
¹S. Hong, ISSCC 2014

²M. Kim, ISSCC 2017

Sinewave Generation



☹ $>100\mu\text{W}$

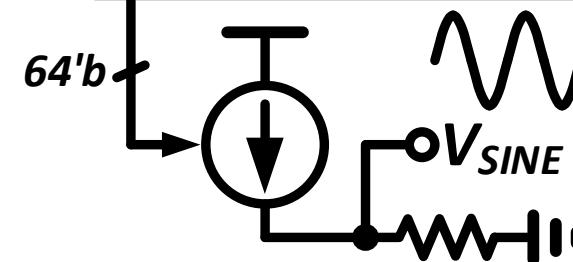


Analog RC-OSC¹

```

module ThermoLUT ();
...
assign LUT[0] = 64'b0...000;
assign LUT[1] = 64'b0...001;
assign LUT[2] = 64'b0...011;
...

```

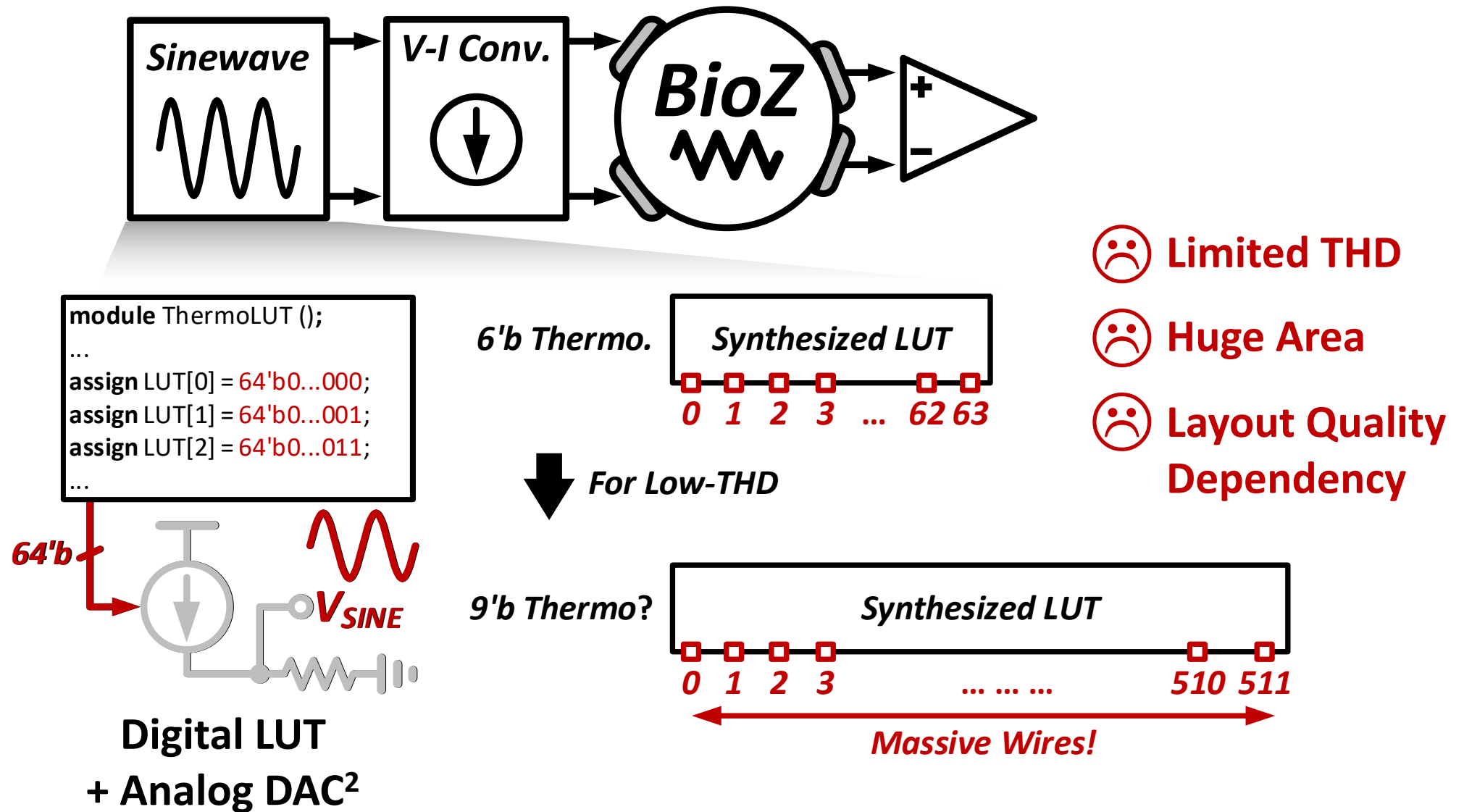


Digital LUT
+ Analog DAC²

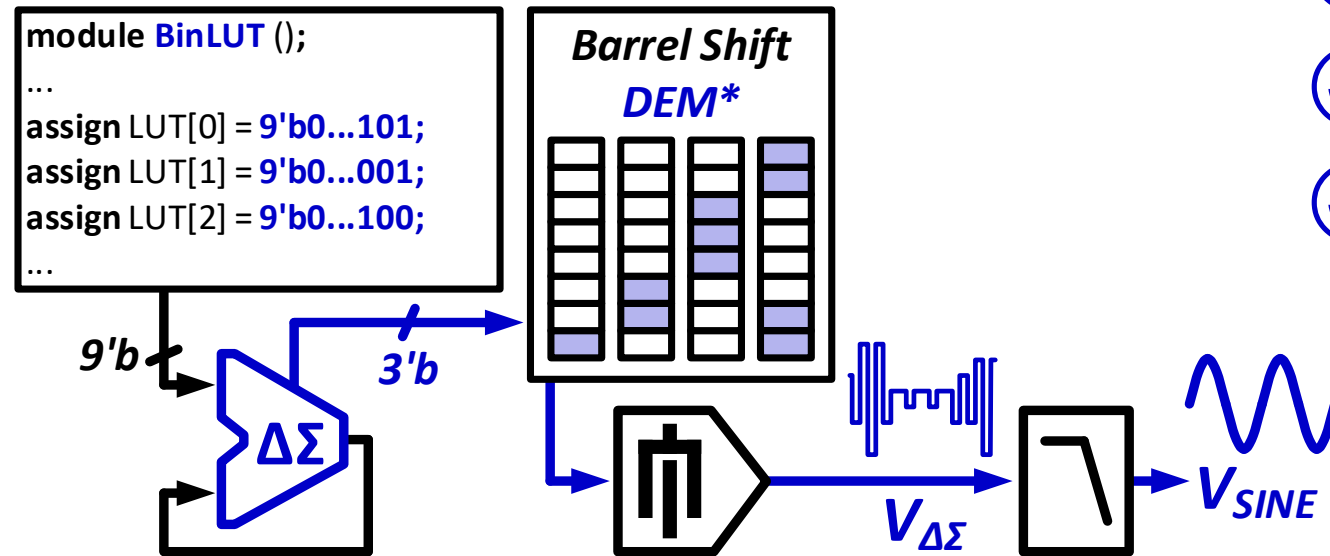
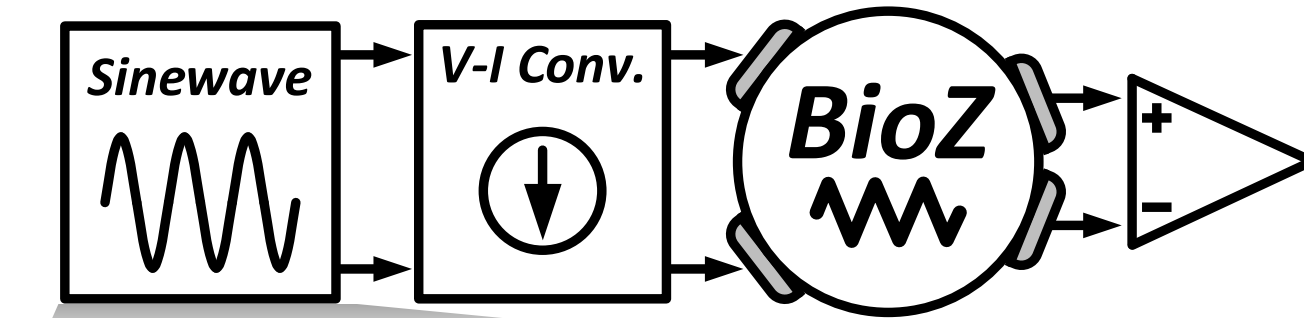
¹S. Hong, ISSCC 2014

²M. Kim, ISSCC 2017

Sinewave Generation



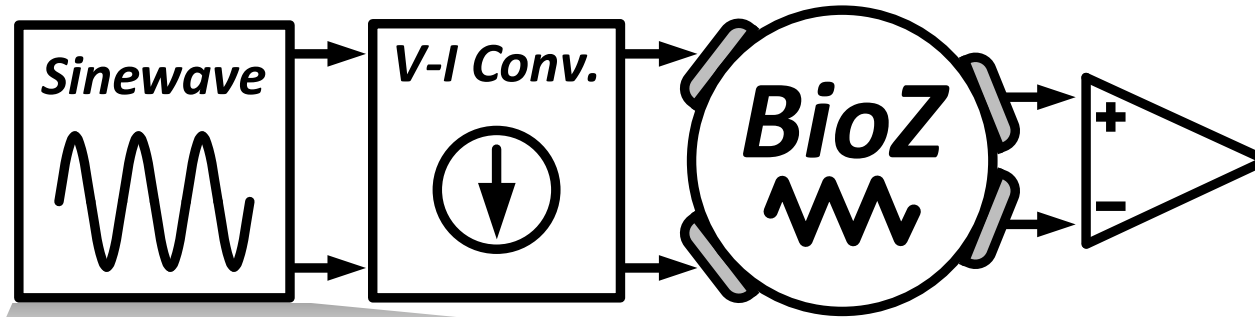
Digital $\Delta\Sigma^1$ Sinewave Generation



- 😊 Low THD
- 😊 Compact Area
- 😊 Layout Quality Independent (DEM)

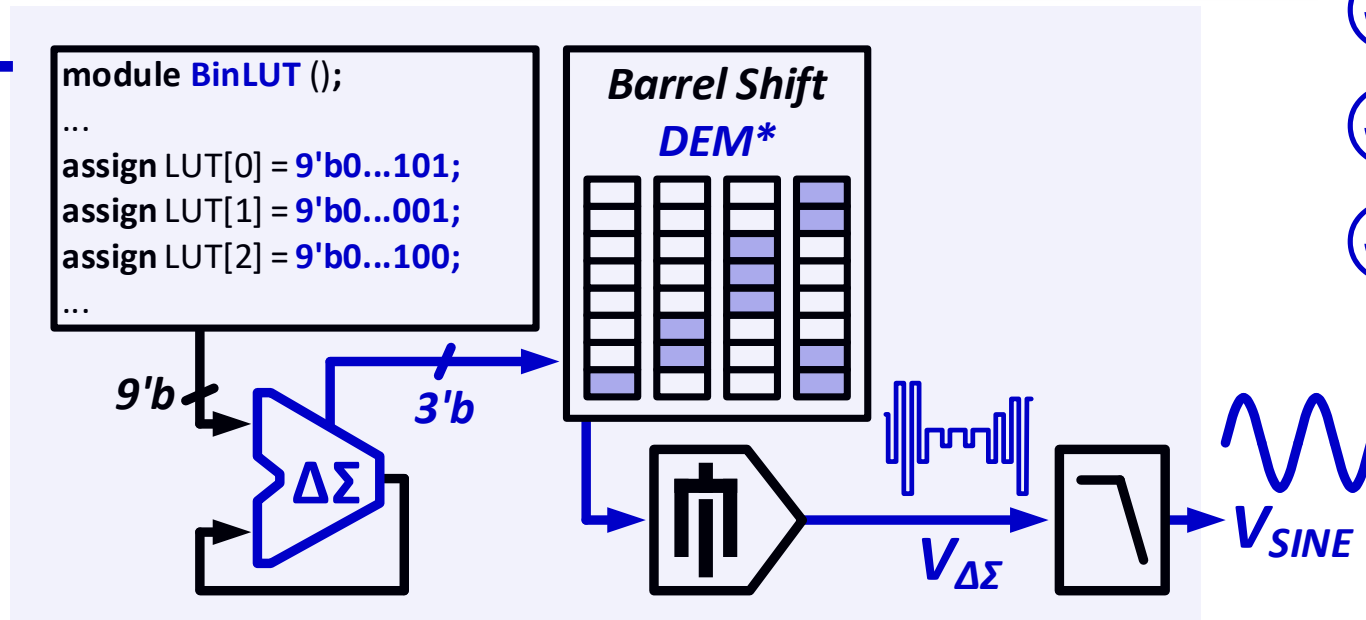
Digital LUT + Digital $\Delta\Sigma$ + Analog DAC + Analog LPF

Digital $\Delta\Sigma^1$ Sinewave Generation



0.5V
(Near-Threshold)

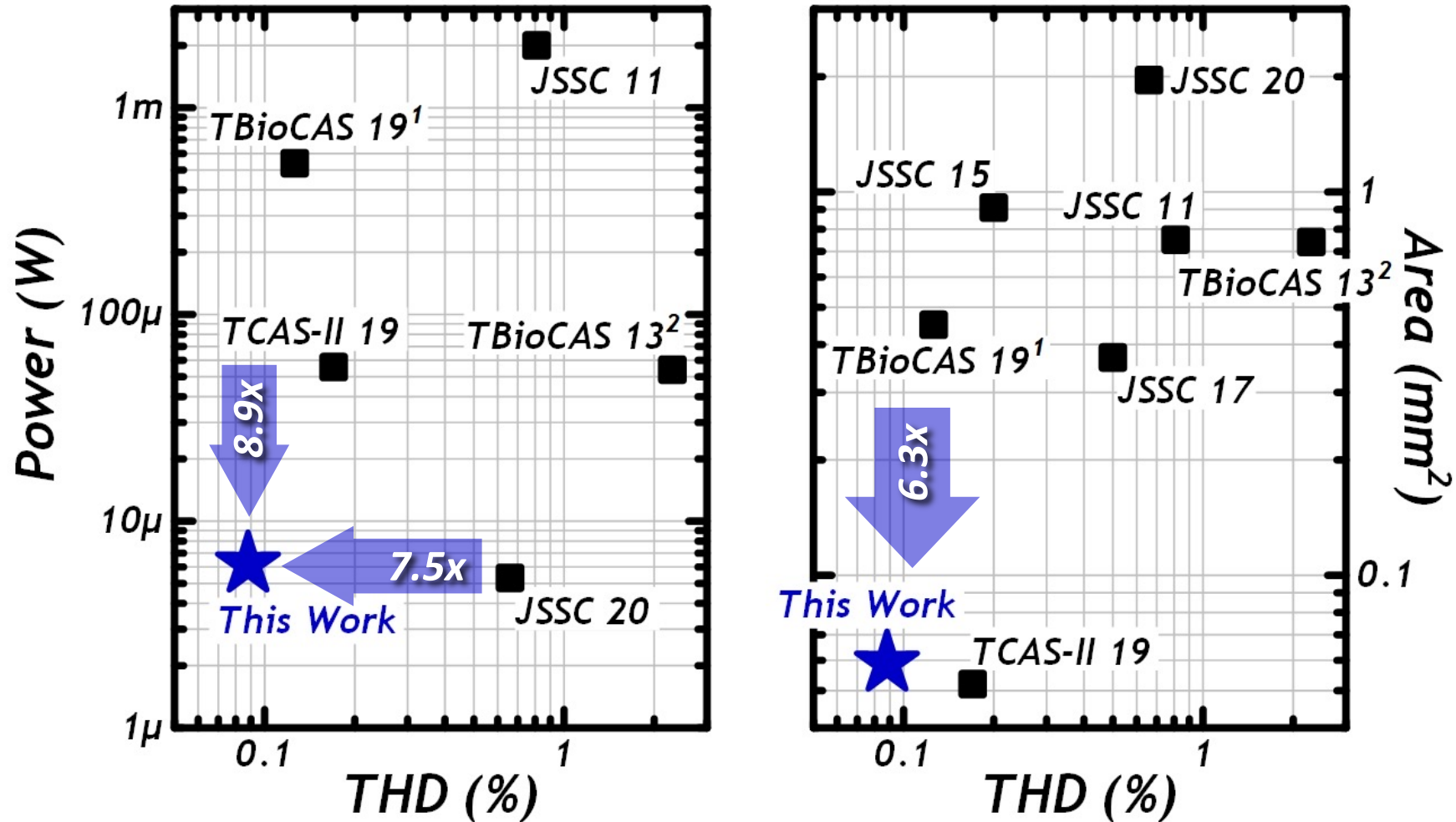
😊 **Low Power**



- 😊 **Low THD**
- 😊 **Compact Area**
- 😊 **Layout Quality Independent (DEM)**

Digital LUT + Digital $\Delta\Sigma$ + Analog DAC + Analog LPF

Comparison Graph



The First **Sub-10 μ W, Sub-0.1% THD**
Sinusoidal Current Generator for BioZ Sensor ICs

Open-Sourced Sinewave Generator^{1,2}

박사 학위 논문
Ph.D. Dissertation

정확한 위상 측정이 가능한 저전력
생체임피던스 센서 회로

Low Power Bio-Impedance Sensor IC
with Precise Phase Measurement

2021

김관태 (金官泰 Kim, Kwantae)

한국과학기술원

Korea Advanced Institute of Science and Technology

the 2nd-order LPF which follows the capacitive-DAC as depicted in Figure 31. In addition, a doubled sampling ratio from 64× to 128× helps to relax the filter design requirements. Finally, when it is 3rd-order ΔΣ modulated, the output spectrum looks like rightmost graph of Figure 34. With the 20 kHz of output frequency, we can expect the overall noise sources, including spurs and quantization noise, can be suppressed down below the -60 dBc, after low pass filtering. Although similar concept of this approach is provided in [35], it required a large-bit-width of 18 bit in PRNG to dither its poor spur performance of the oscillator, degrading its power efficiency. Here, the most of spur reduction is obtained only by the 9 bit LUT that is sufficient to provide > 70dB of in-band spur. Furthermore, the use of error-feedback-based ΔΣ modulator halves the required number of accumulators in each of the modulator stages compared to the output-feedback structure [35], greatly simplifying the implementation of the ΔΣ modulator.

The integer-numbered sinewave is generated using MATLAB environment. Detailed MATLAB code is provided as below.

Table II. MATLAB Code Used for the Generation of Pseudo-Sine Waveform

```

%%% Start
% Created by Kwantae Kim, 2019
clc;
clear;
close all;

% Global Parameters
clock=2.56*10^6;
OSR=128;

%%% PseudoSine Parameters
Am = OSR*2; % Define Amplitude for the Integer Components of LUT
% This is not a Pk-Pk value
f = clock/OSR; % Frequency
AmHigh = Am-1; % Max. Amplitude
AmLow = -Am+1; % Min. Amplitude

%%% Array Generation (Sine, PseudoSine)
t=1/(f*OSR) : 1/(f*OSR) : (1/f); % x-axis for Time (1 Period)
Sine=AmHigh*sin(2*pi*f*t); % Sine Generation

PseudoSineValue = zeros(1,OSR); % Array for PseudoSine Value
PseudoSineValue(1) = AmLow;
for i = 2:1:Am*2
    PseudoSineValue(i) = PseudoSineValue(i-1)+1;
end
    
```

MATLAB (Software)

4.2.2 Implementation of the Digital Circuit

Functional verifications of the register transfer level (RTL) design for the digital part of proposed CG IC is done with Cadence Virtuoso (AMS simulator), and the logic synthesis of RTL design is processed with Synopsys Design Compiler.

Verilog (Hardware)

Table III describes an instantiation-tree of the RTL design.

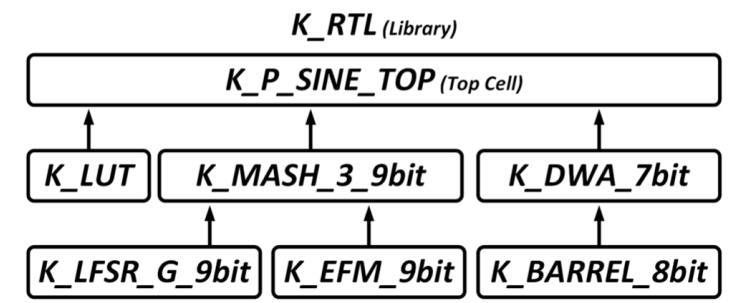


Figure 35. Instantiation-tree of the RTL design.

Table III. Register Transfer Level Design of the Digital Part

```

//Verilog HDL for "K_RTL", "K_P_SINE_TOP" "functional"

module K_P_SINE_TOP (
    input wire clk,
    input wire resetn,
    output wire [7:0] sine_outp,
    output wire [7:0] sine_outn,
    output wire sine_dac_rst_p,
    output wire sine_dac_rst_n
);

wire [8:0] lut_out;
K_LUT lut(
    .clk (clk),
    .resetn (resetn),
    .out (lut_out)
);

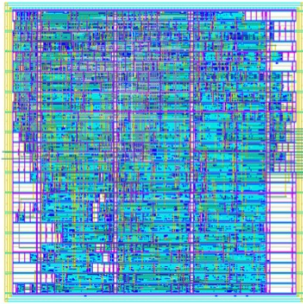
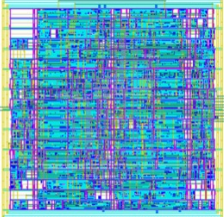
wire [6:0] dsm_out;
    
```

¹K. Kim, VLSI 2020 + JSSC 2022

²K. Kim, Ph.D. Dissertation

Adoption in Other Work

TABLE II
COMPARISON AND SUMMARY OF SSG
PRESENTED IN [9] AND THIS BRIEF

	SSG based on [9]	This work
		
SFDR	64.7 dBc	64.6 dBc
THD	0.089%	0.074%
Output Power ^a	3 bits	2 bits
Cell area ^b	44.7 μW	24.8 μW
Floorplan area ^{b,c}	10148 μm^2	5334 μm^2
	13447 μm^2	7252 μm^2

^a $V_{DD} = 1.8 \text{ V}$, $f_{CK} = 4 \text{ MHz}$,

^{a,b}Using the same $0.18 \mu\text{m}$ CMOS standard cells,

^cLayout that meets DRC requirements and is fully routed.

1764

IEEE TRANSACTIONS ON CIRCUITS AND SYSTEMS—II: EXPRESS BRIEFS, VOL. 70, NO. 5, MAY 2023

A Sinusoidal Signal Generator Using a Delta-Sigma Modulated Look-Up Table and Analysis of Dither

Jaehyeong Park^{ID}, Graduate Student Member, IEEE, and Matthew L. Johnston^{ID}, Senior Member, IEEE

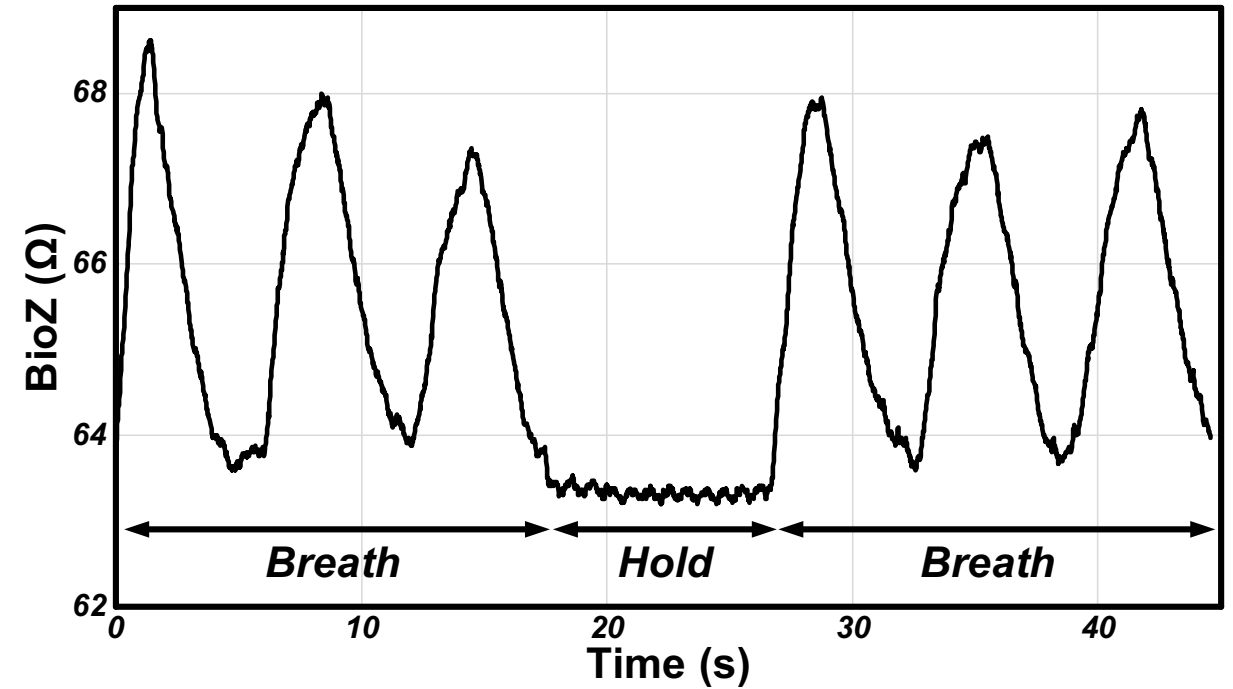
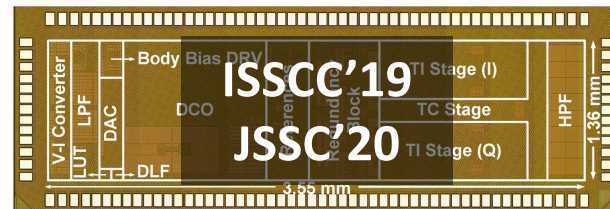
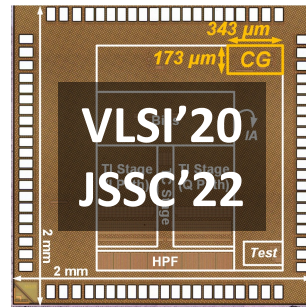
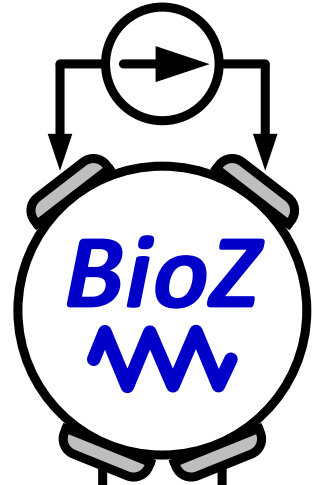
- [9] K. Kim, S. Kim, and H.-J. Yoo, "Design of sub-10- μW sub-0.1% THD sinusoidal current generator IC for bio-impedance sensing," *IEEE J. Solid-State Circuits*, vol. 57, no. 2, pp. 586–595, Feb. 2022.
- [14] K. Kim, "Low power bio-impedance sensor IC with precise phase measurement," Ph.D. dissertation, School Electr. Eng., Korea Adv. Inst. Sci. Technol., Daejeon, South Korea, 2021.

The design is **Replicated & Improved**
by Oregon State University

In-Vivo BioZ Measurement

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



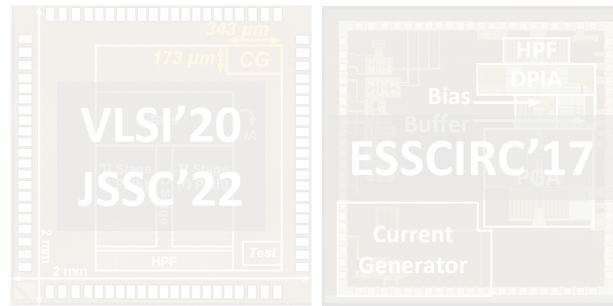
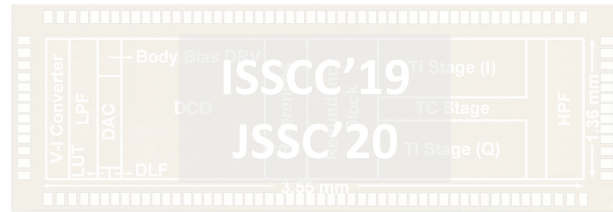
Biomedical Sensor IC

$$\text{BioZ} = \frac{V_{\text{Sense}}}{I_{\text{Inject}}}$$

Current (I)



Voltage (V)



Neuromorphic Sensor IC

Mic.



FEx

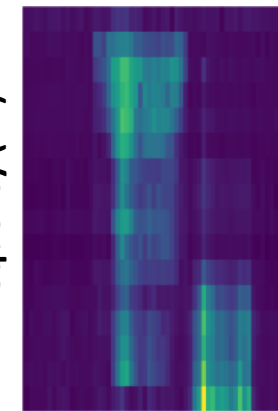


DNN



"Yes"

Frequency (Hz)

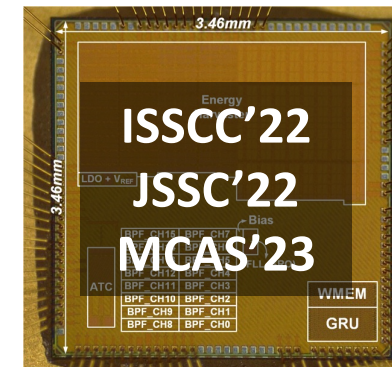


Time (s)

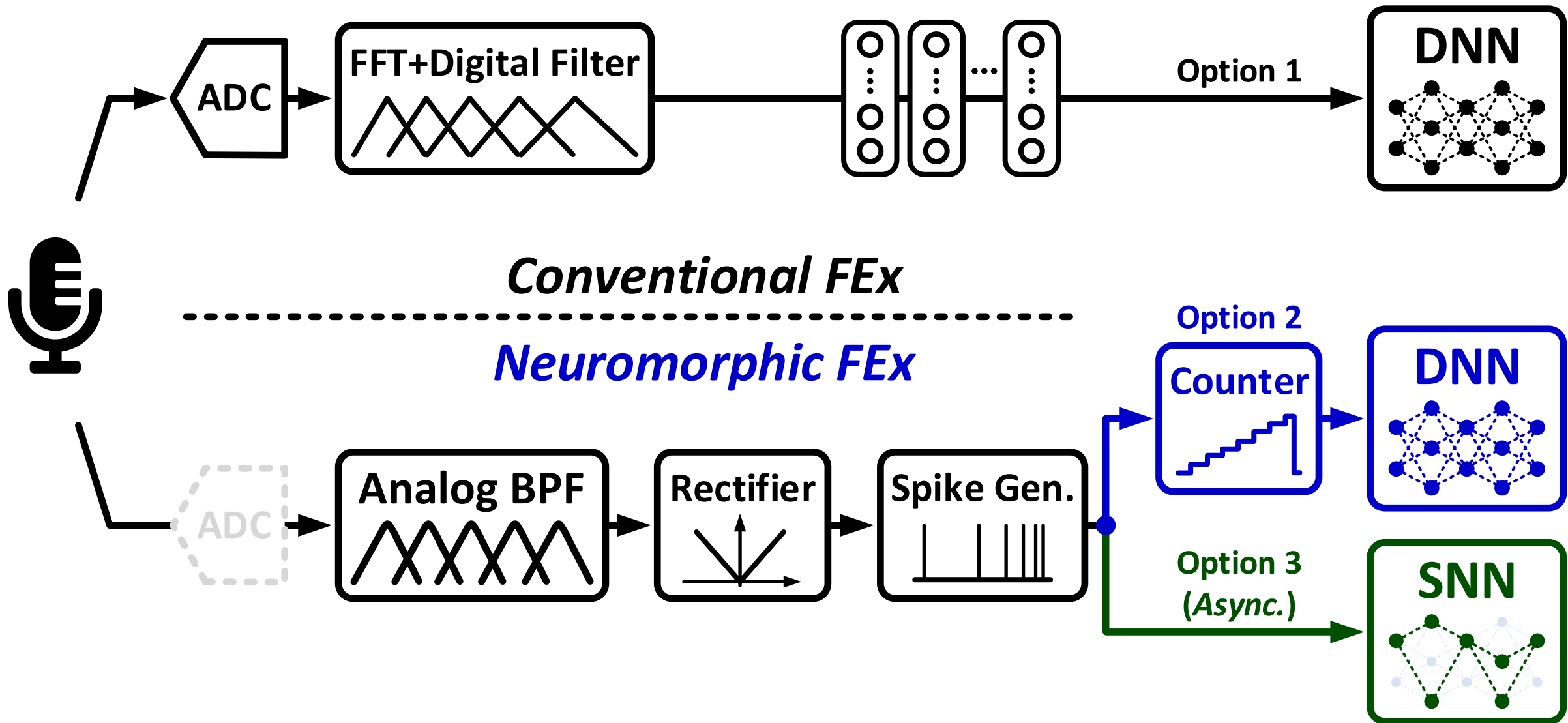
Score

"Yes"	0.95
"Silence"	0.02
"Unknown"	0.01
"Down"	0.01
"Go"	0.00

⋮

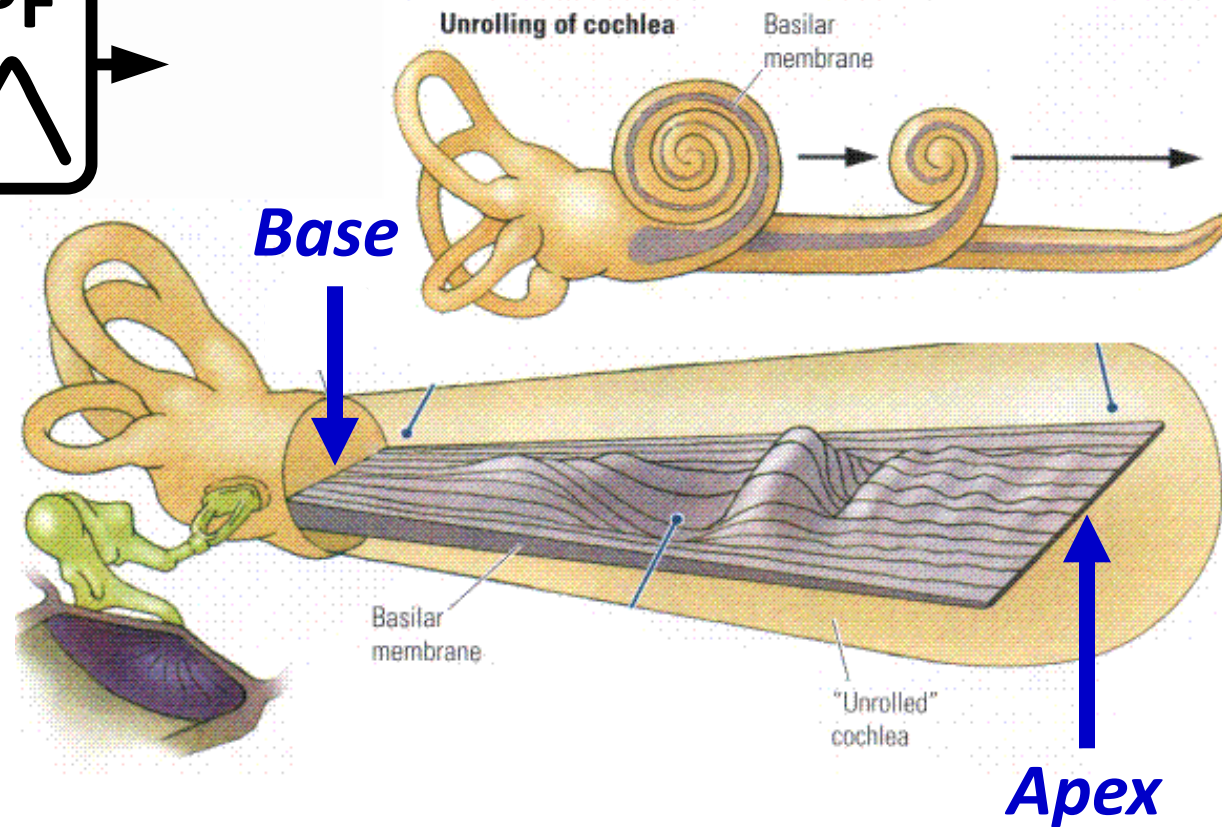


Neuromorphic Audio Processing Architecture

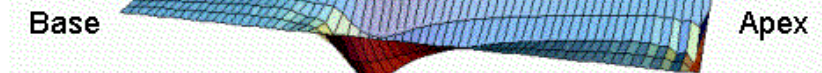


Neuromorphic Audio Processing Architecture

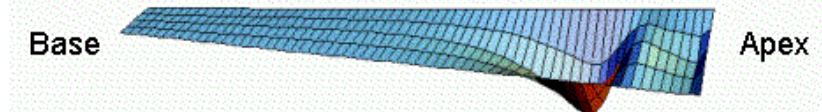
Cochlea



6kHz Input



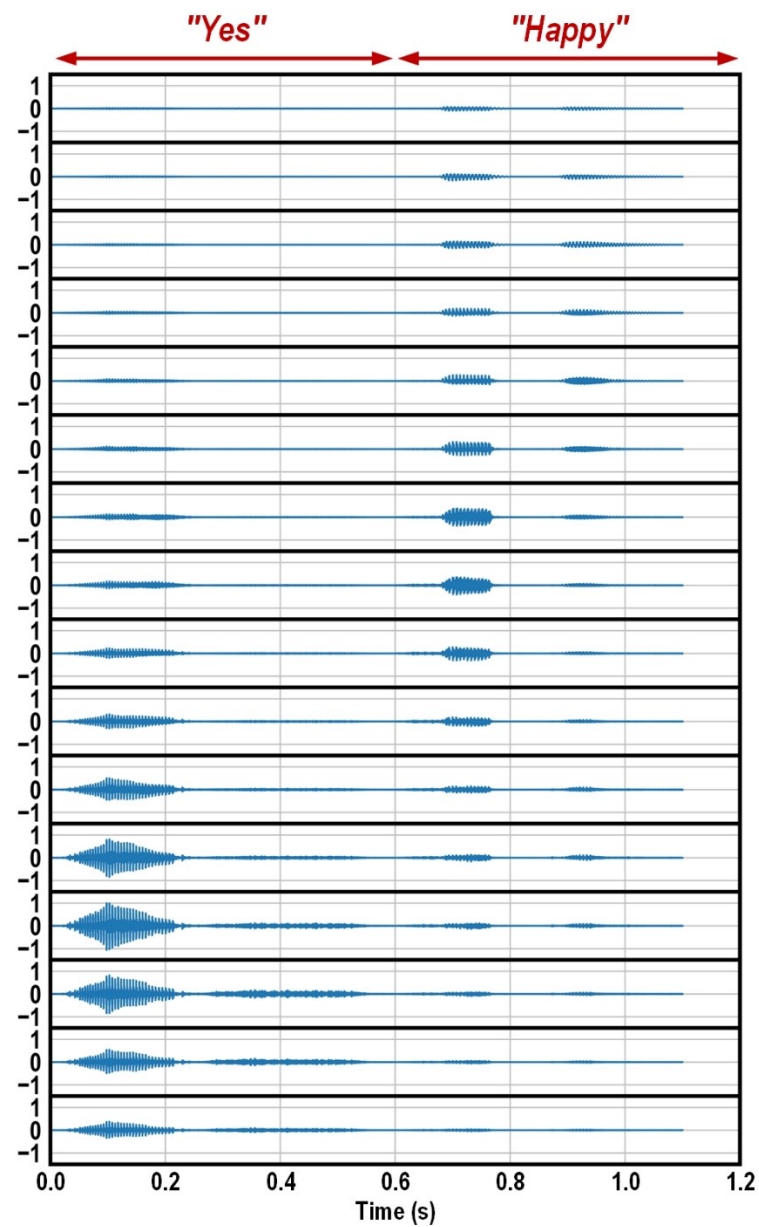
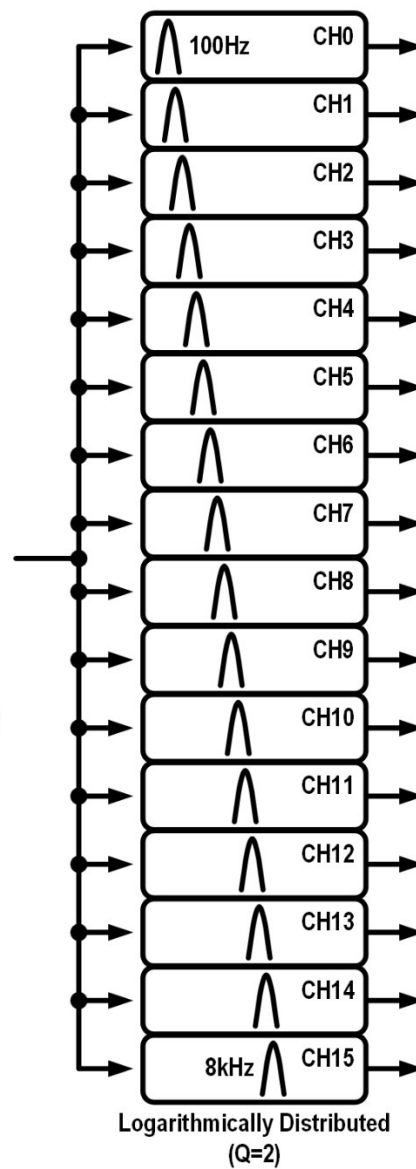
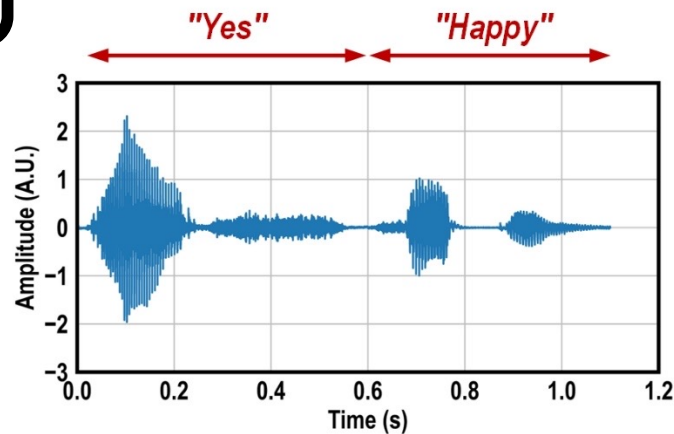
2kHz Input



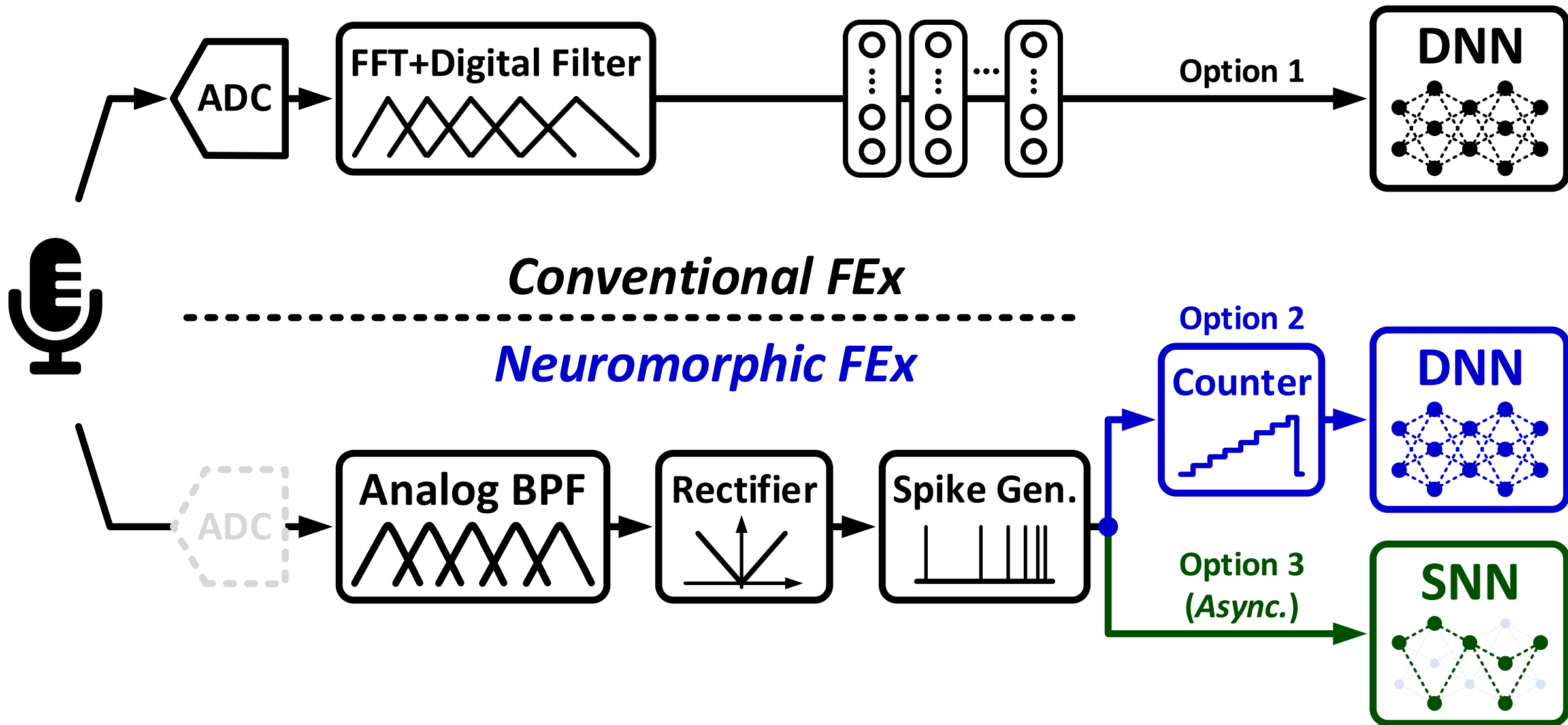
Neuromorphic Audio Processing Architecture

Cochlea

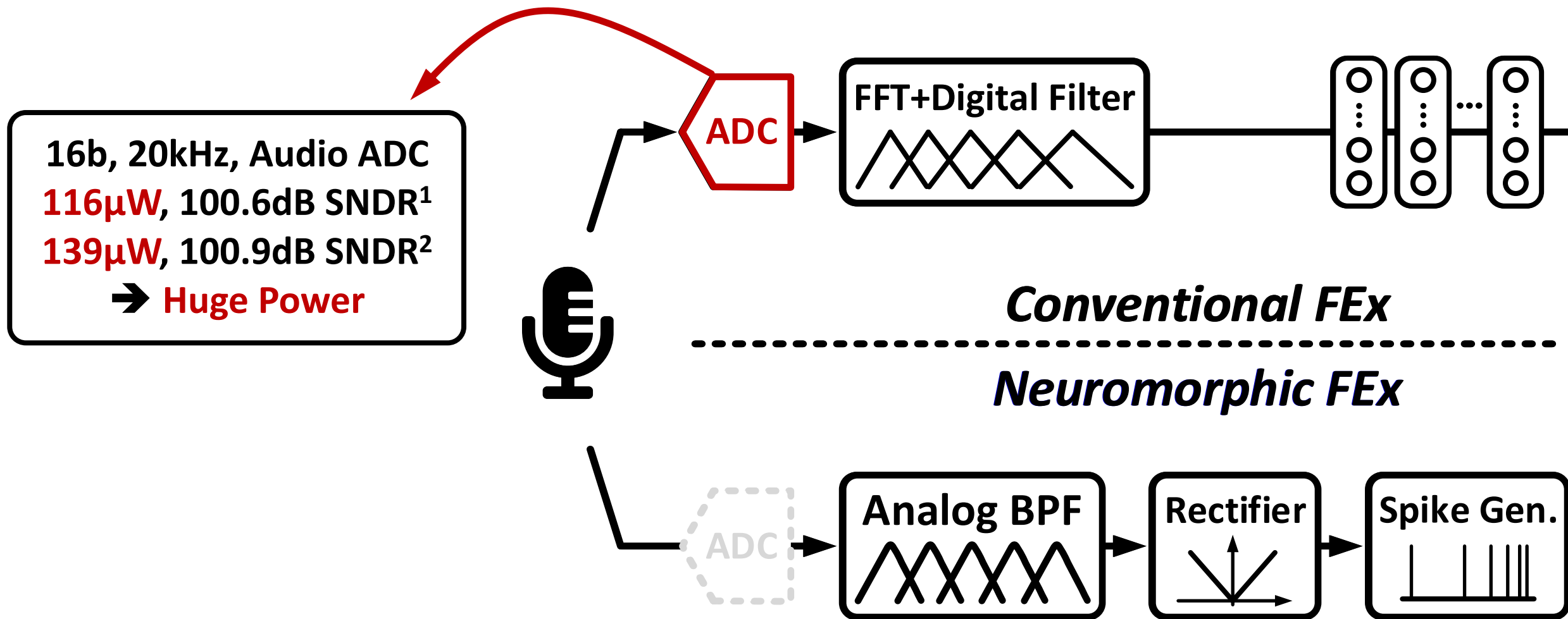
Analog BPF



Neuromorphic Audio Processing Architecture

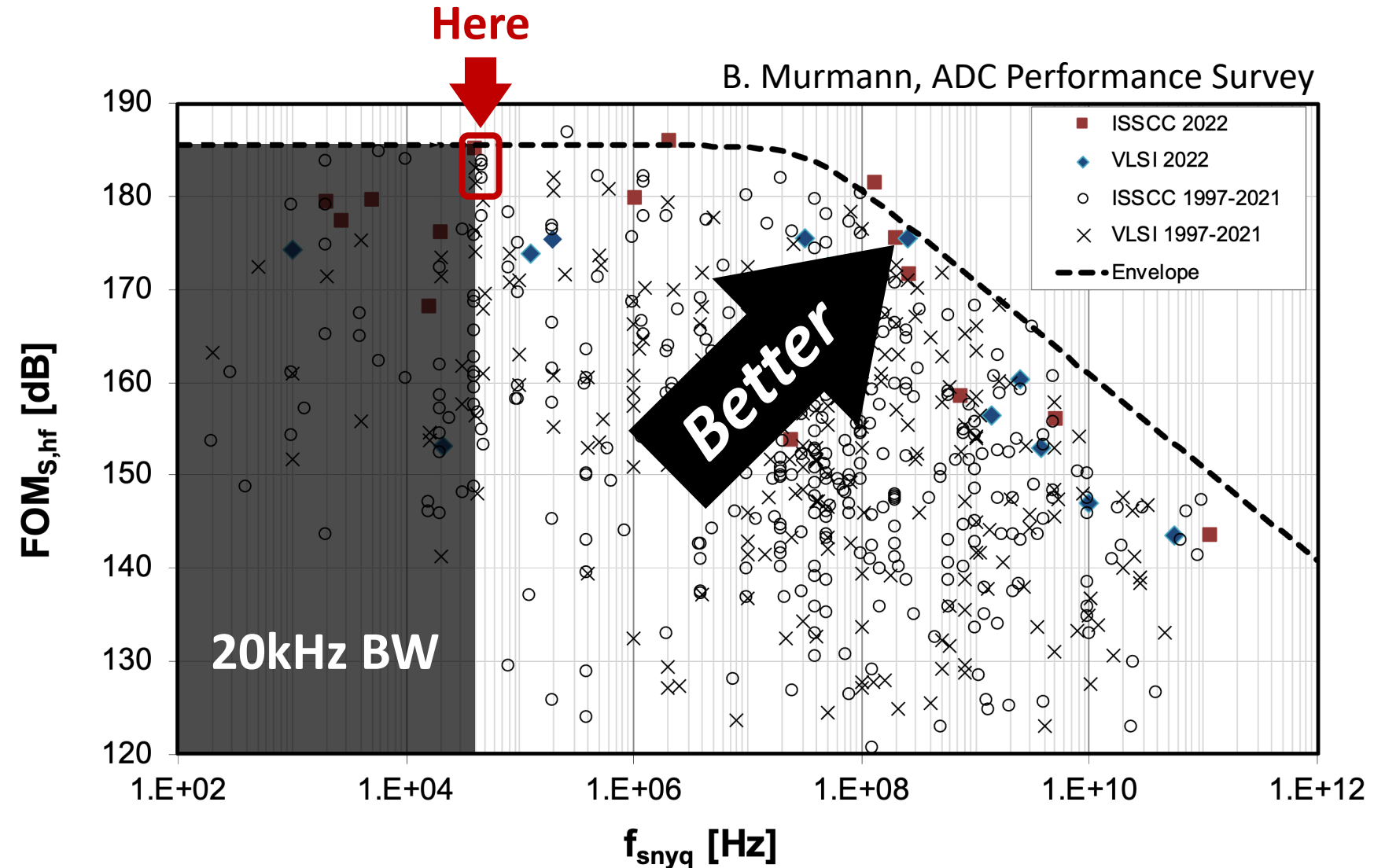


Why Neuromorphic Approach? – 1. ADC



Why Neuromorphic Approach? – 1. ADC

16b, 20kHz, Audio ADC
116 μ W, 100.6dB SNDR¹
139 μ W, 100.9dB SNDR²
→ Huge Power



¹C. Lo, ISSCC 2021

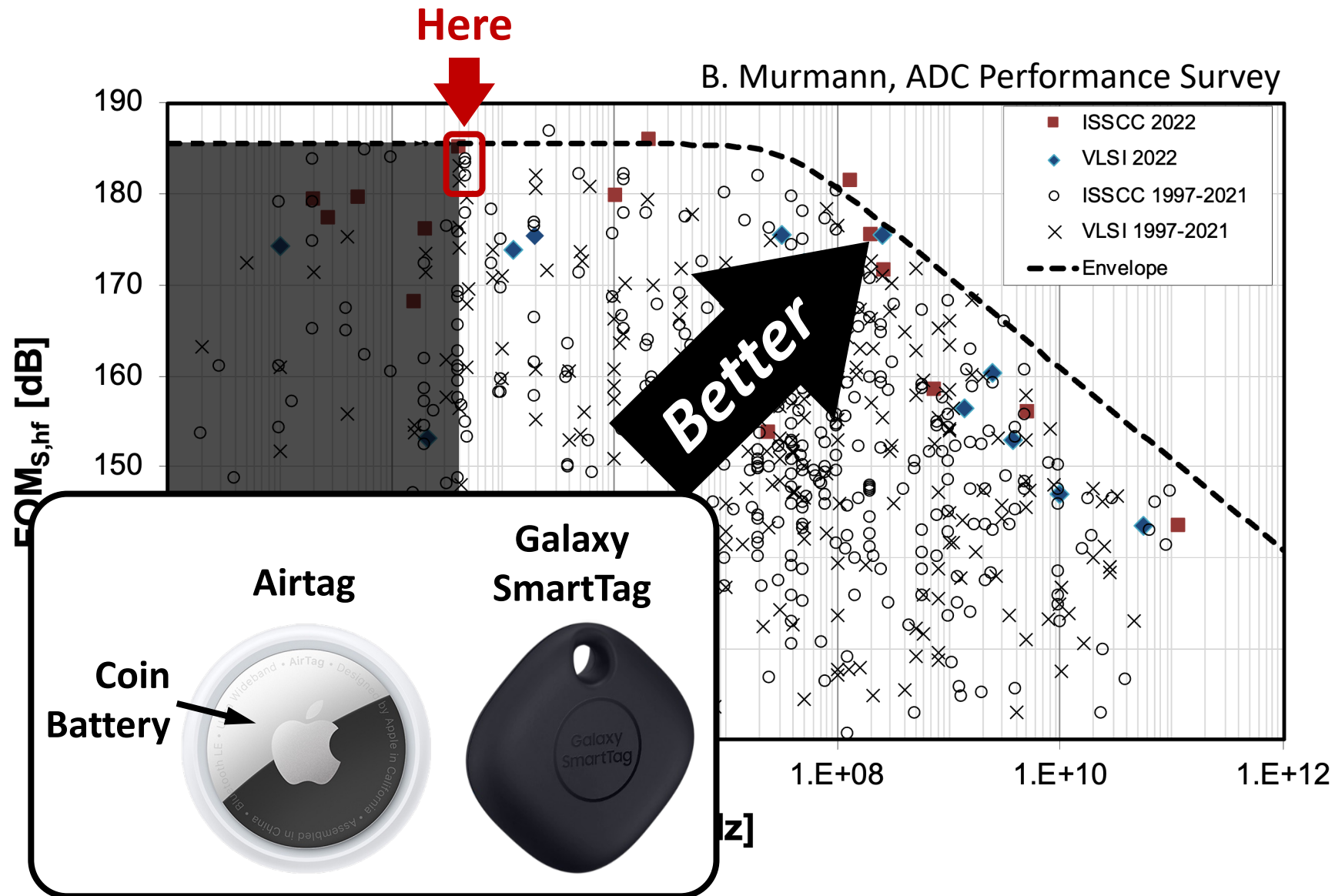
²S. Mondal, ISSCC 2021

Why Neuromorphic Approach? – 1. ADC

16b, 20kHz, Audio ADC
116 μ W, 100.6dB SNDR¹
139 μ W, 100.9dB SNDR²
 ➔ **Huge Power**



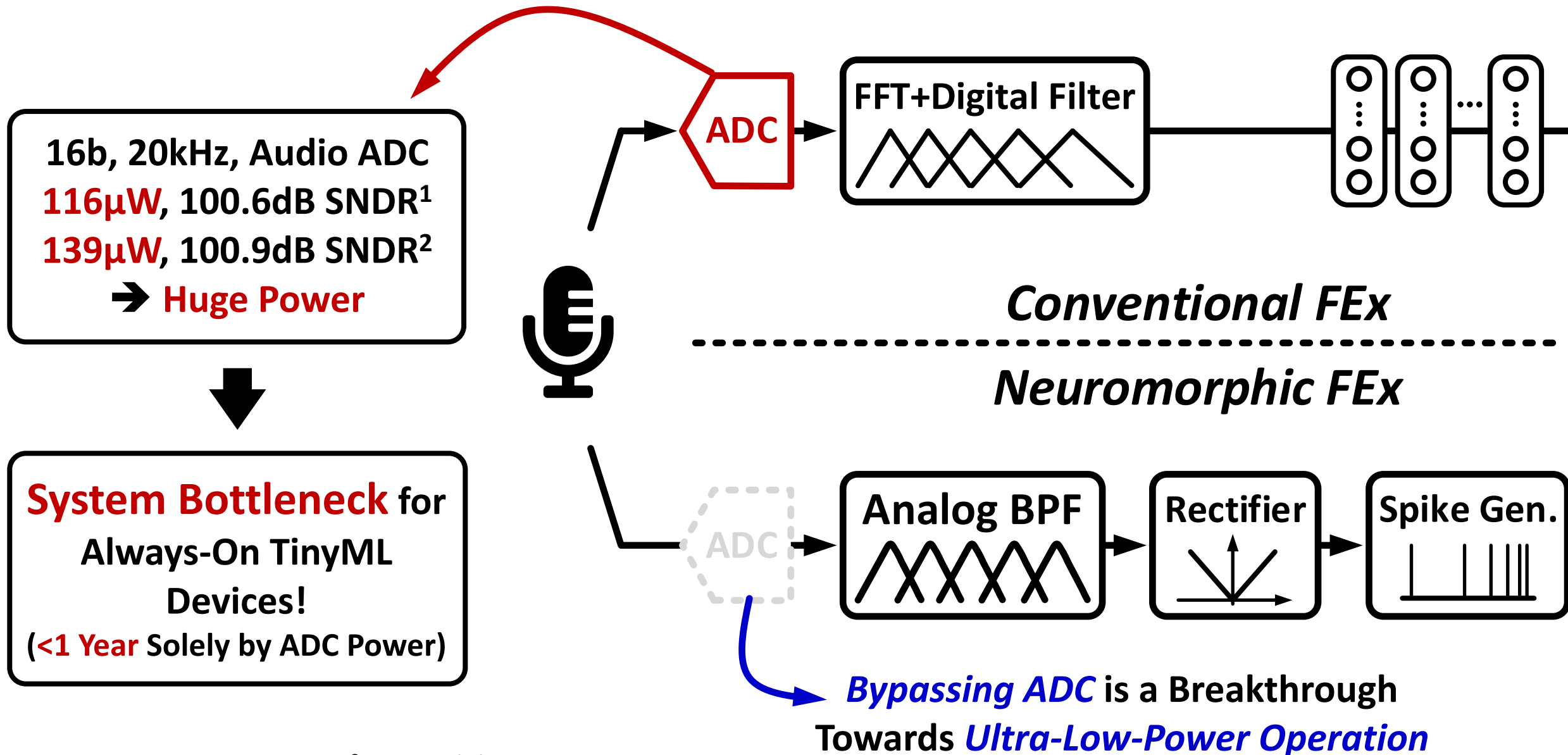
System Bottleneck for
 Always-On TinyML
 Devices!
 (<1 Year Solely by ADC Power)



¹C. Lo, ISSCC 2021

²S. Mondal, ISSCC 2021

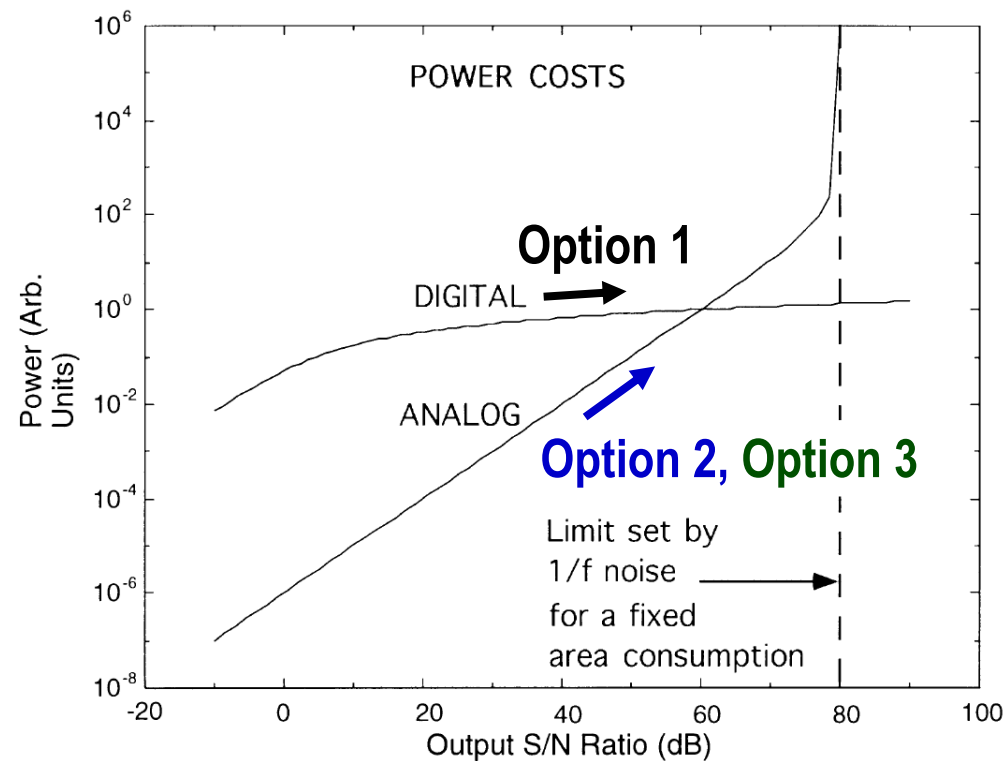
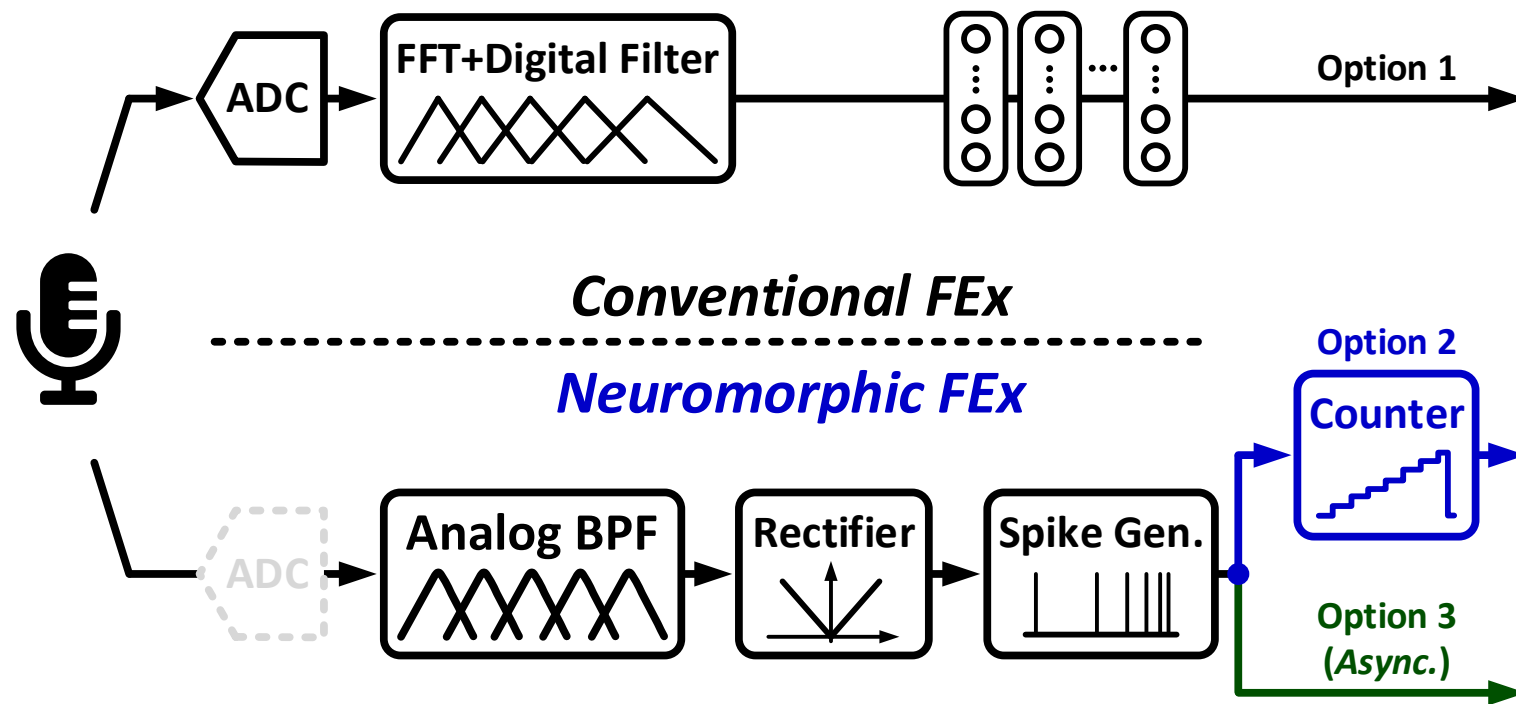
Why Neuromorphic Approach? – 1. ADC



¹C. Lo, ISSCC 2021

²S. Mondal, ISSCC 2021

Why Neuromorphic Approach? – 2. Analog vs Digital

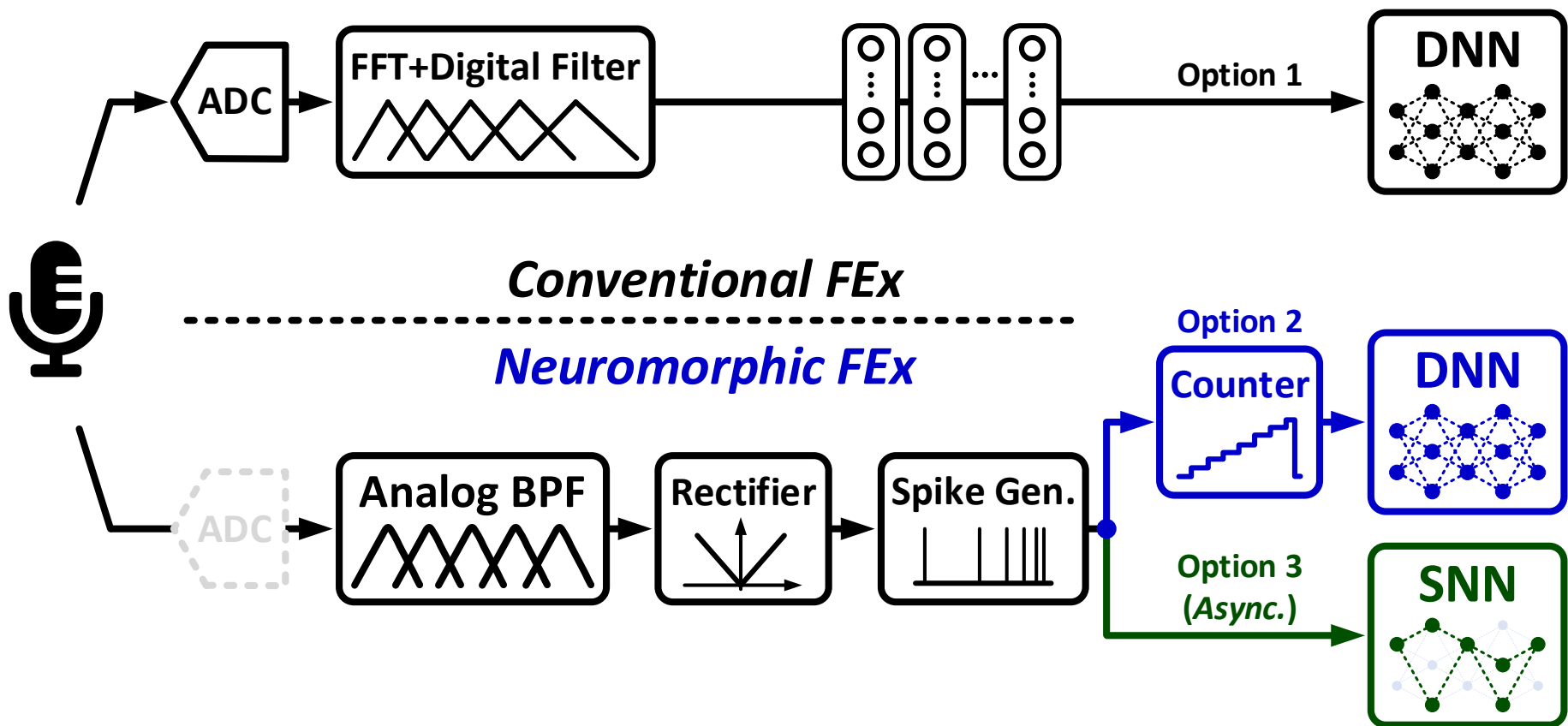


**Analog Signal Processing is
More Efficient up to ~8-Bit Precision^{1,2}**

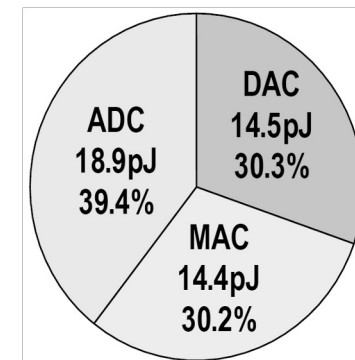
¹R. Sarpeshkar, Neural Computation 1998

²B. Murmann, TVLSI 2021

Why Neuromorphic Approach? – 3. Neural Network



In-Memory Computing
 $E_{ADC} + E_{DAC} \sim 58\%^1 - 70\%^2$!



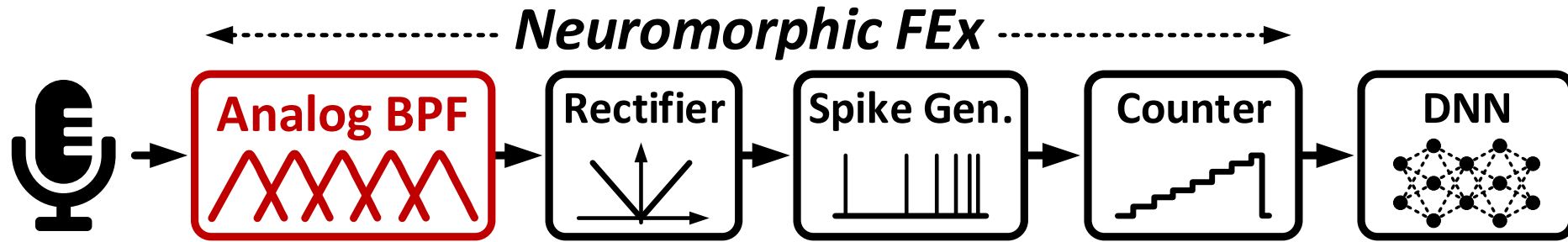
Fully *Asynchronous*
 (or Fully Analog)
 → No ADC & DAC

Spike-Based Interfaces can be
 Seamlessly Integrated with *Asynchronous SNNs*

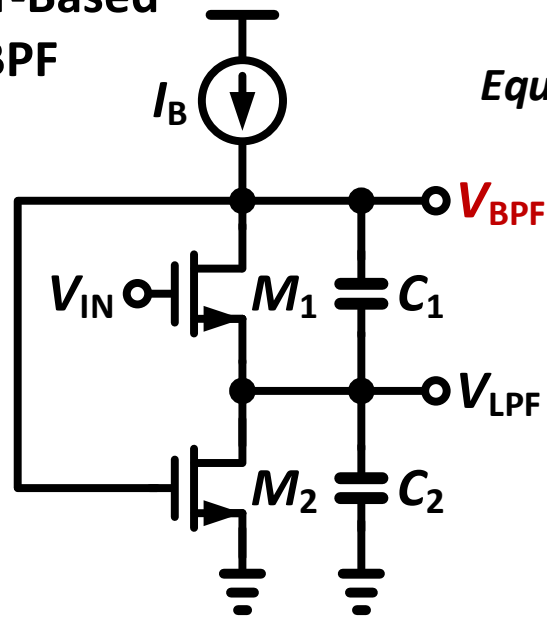
¹A. Biswas, ISSCC 2018

²J.-O. Seo, ISSCC 2022

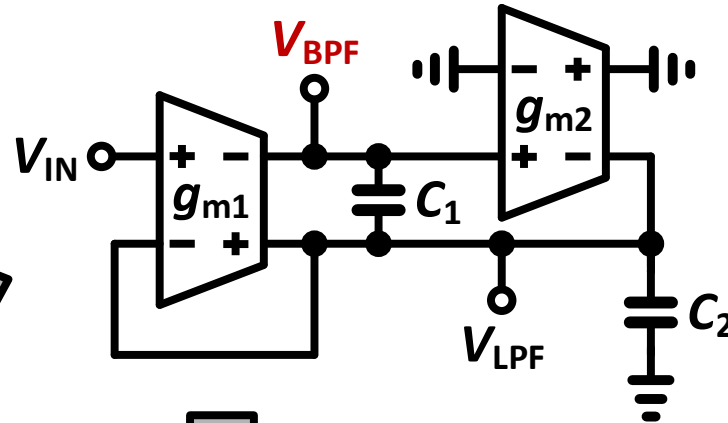
Analog FEx (Voltage Domain)



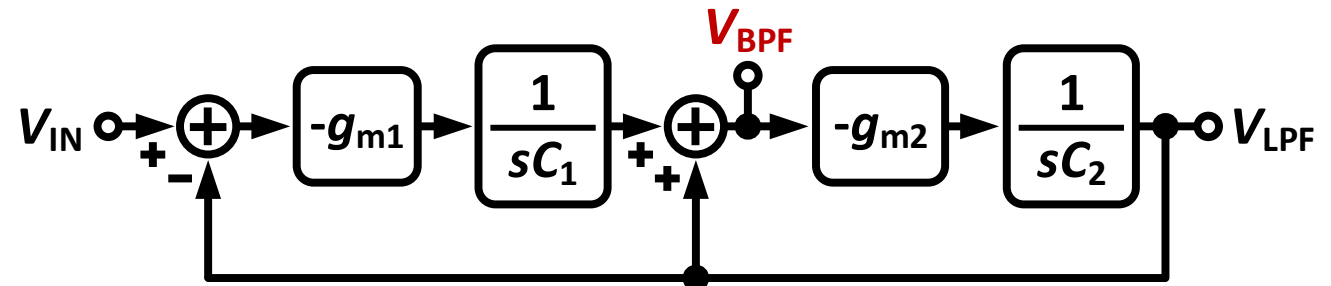
Source-Follower-Based
2nd-Order BPF



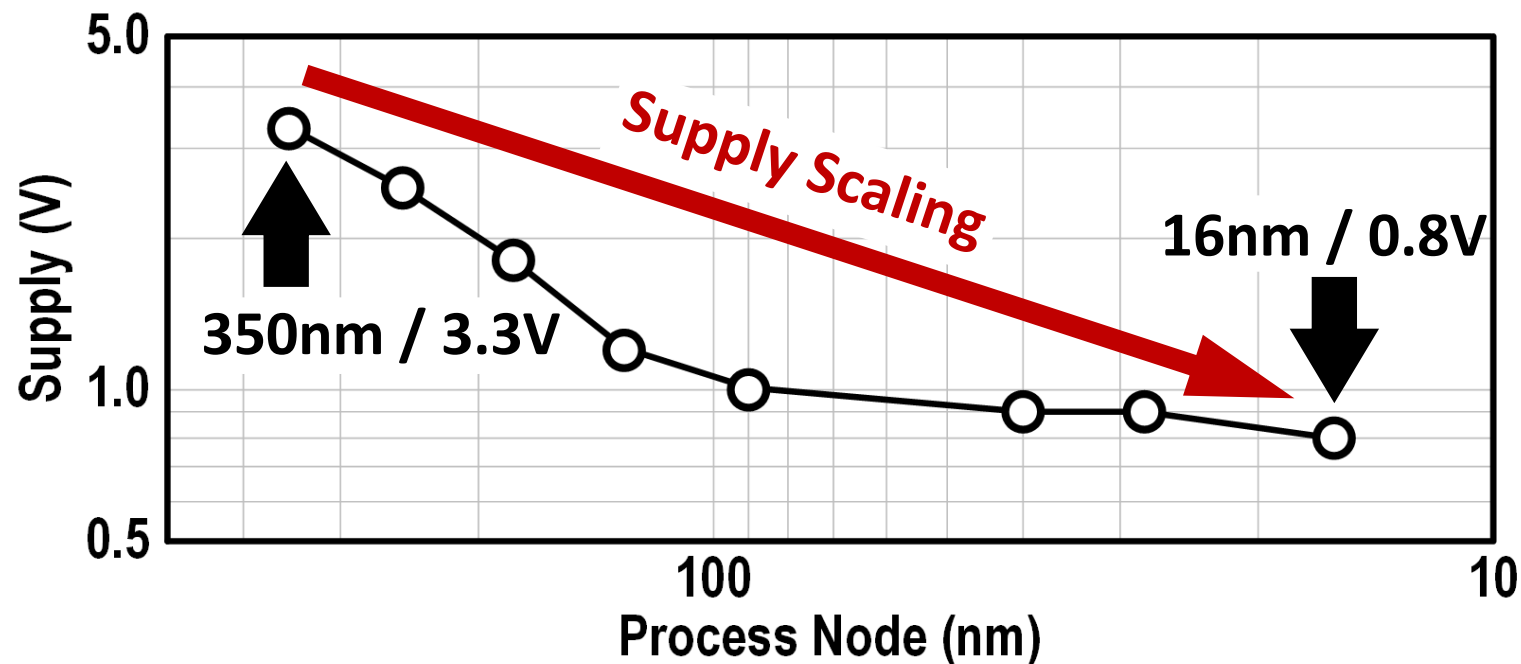
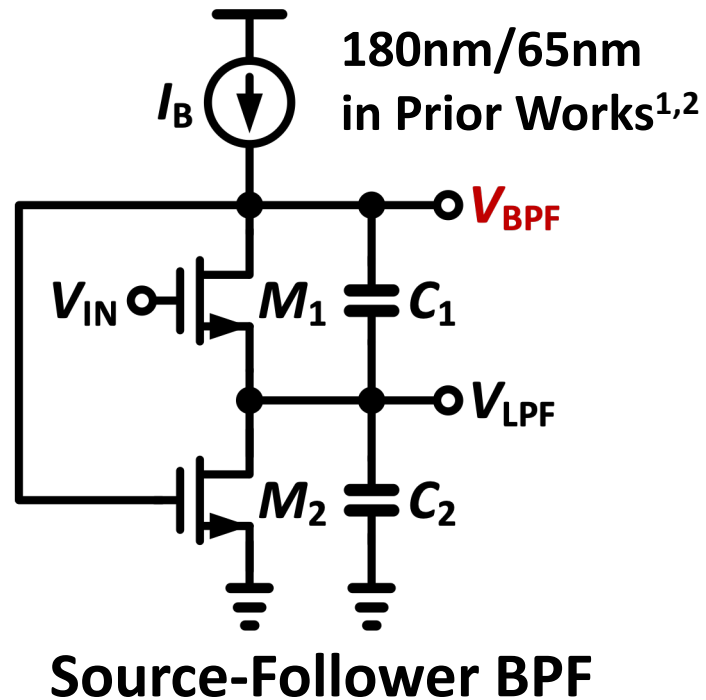
$g_m C$
Equivalent



Small-Signal



Analog FEx (Voltage Domain)



✘ Signal Swing (Headroom)

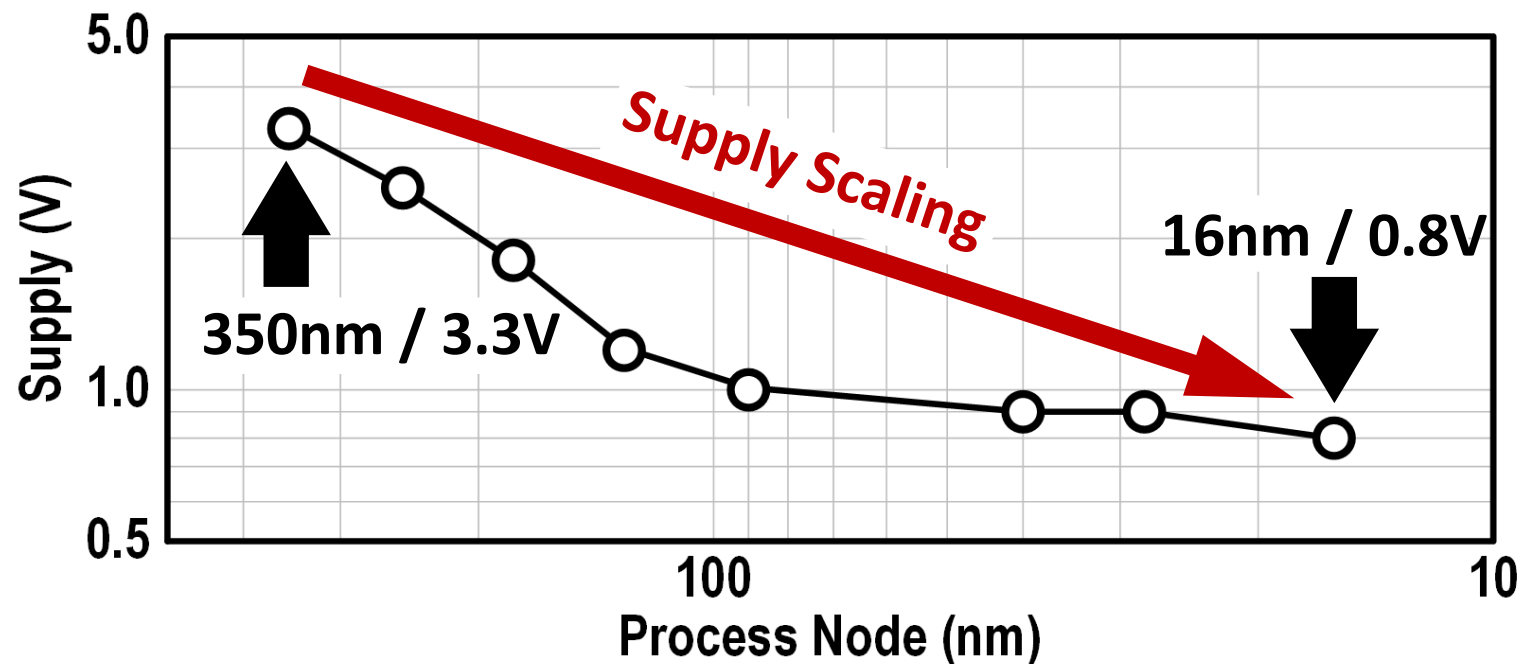
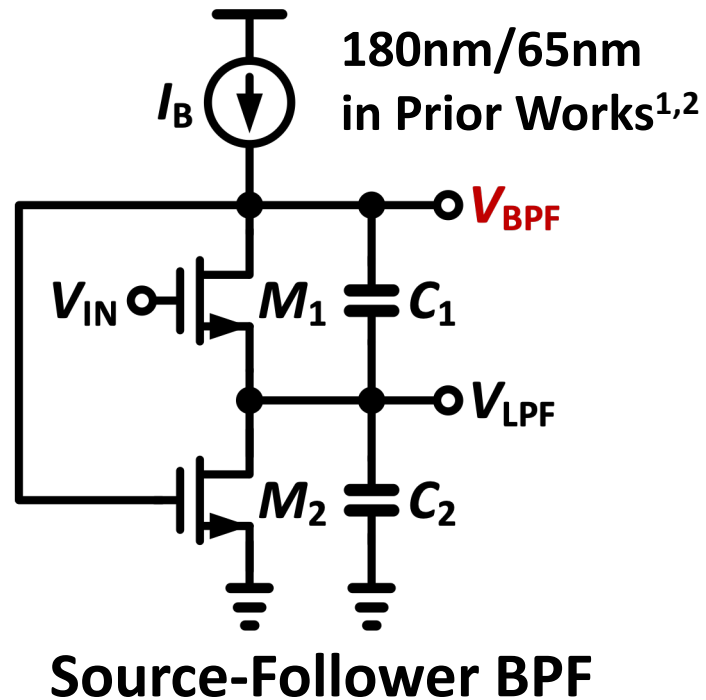
➔ Reduced SNR

✘ DC Gain ($g_m r_o$)

➔ Nonlinearity/Gain Error

➔ Large L (Area/Power)

Analog FEx (Voltage Domain)



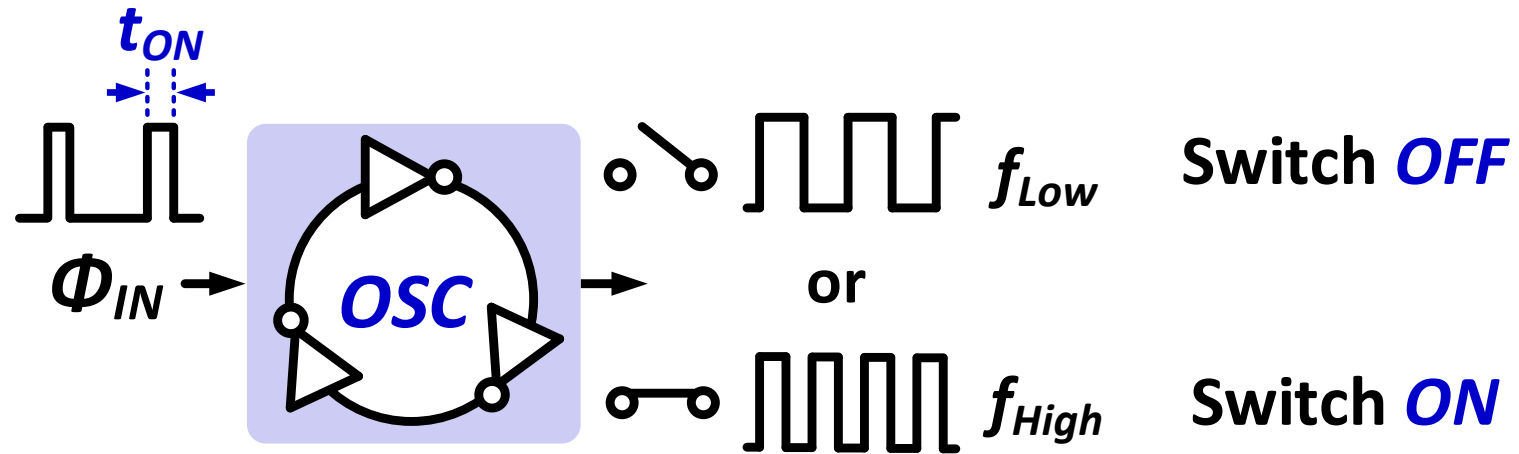
Can Analog FEx Benefit
from Technology Scaling?

¹M. Yang ISSCC 2018

²D. Wang, ISSCC 2021

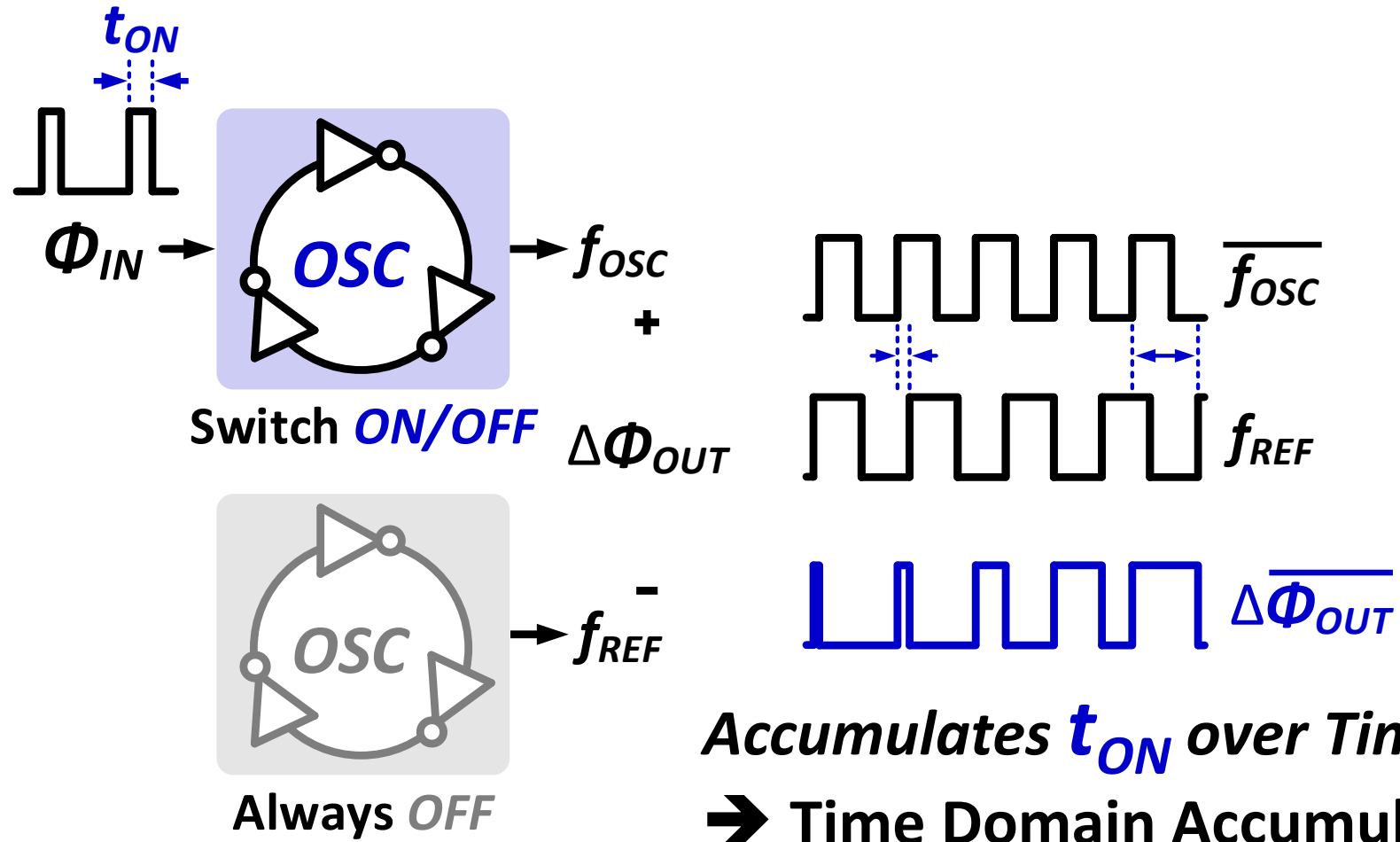
³K. Kim, CAS Magazine 2023

Analog FEx (Time Domain)

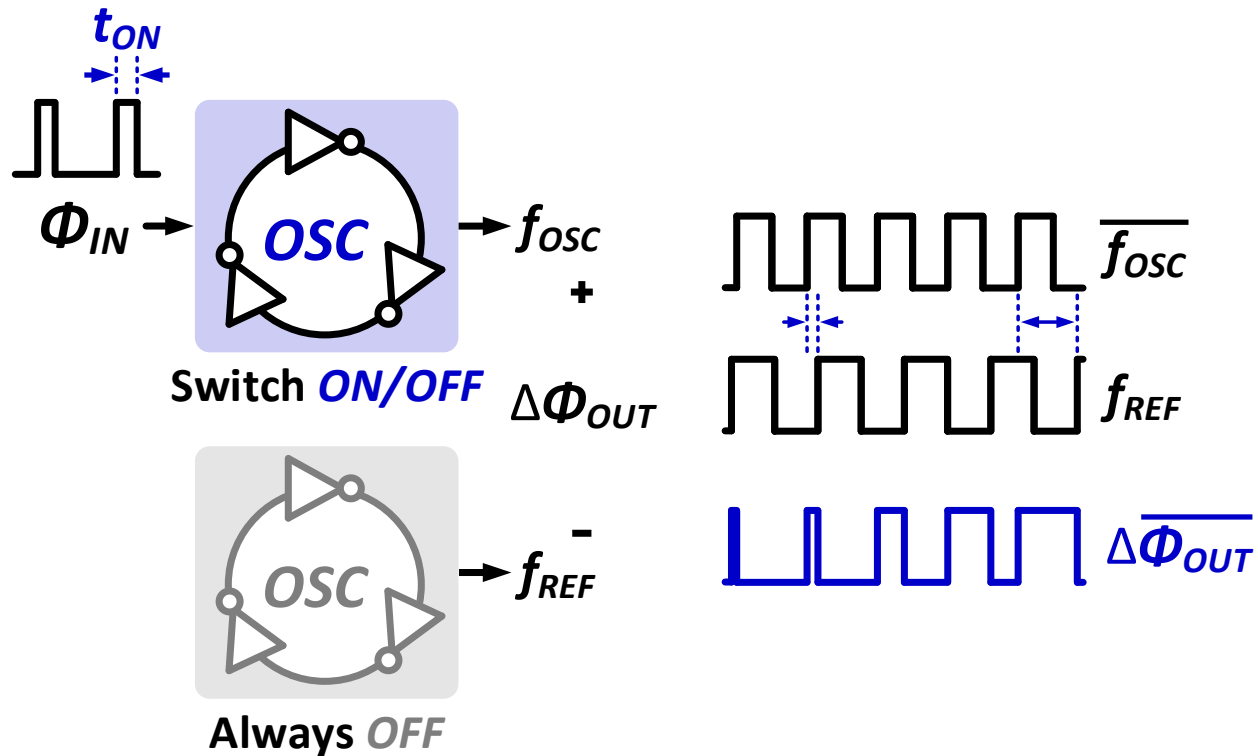


Average Frequency $\propto t_{ON}$

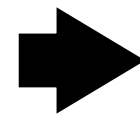
Analog FEx (Time Domain)



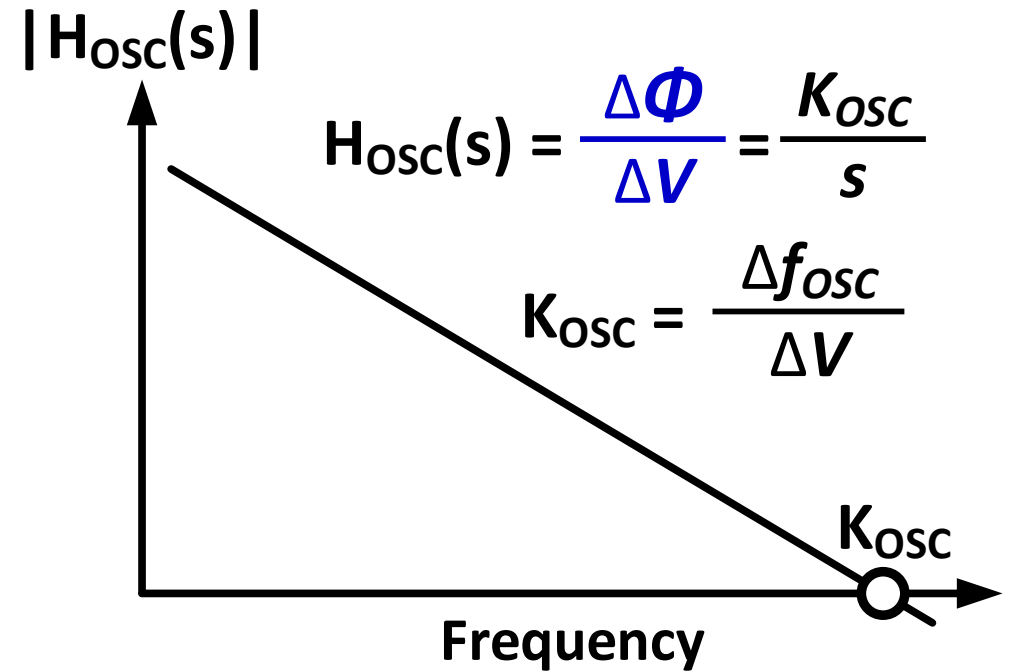
Analog FEx (Time Domain)



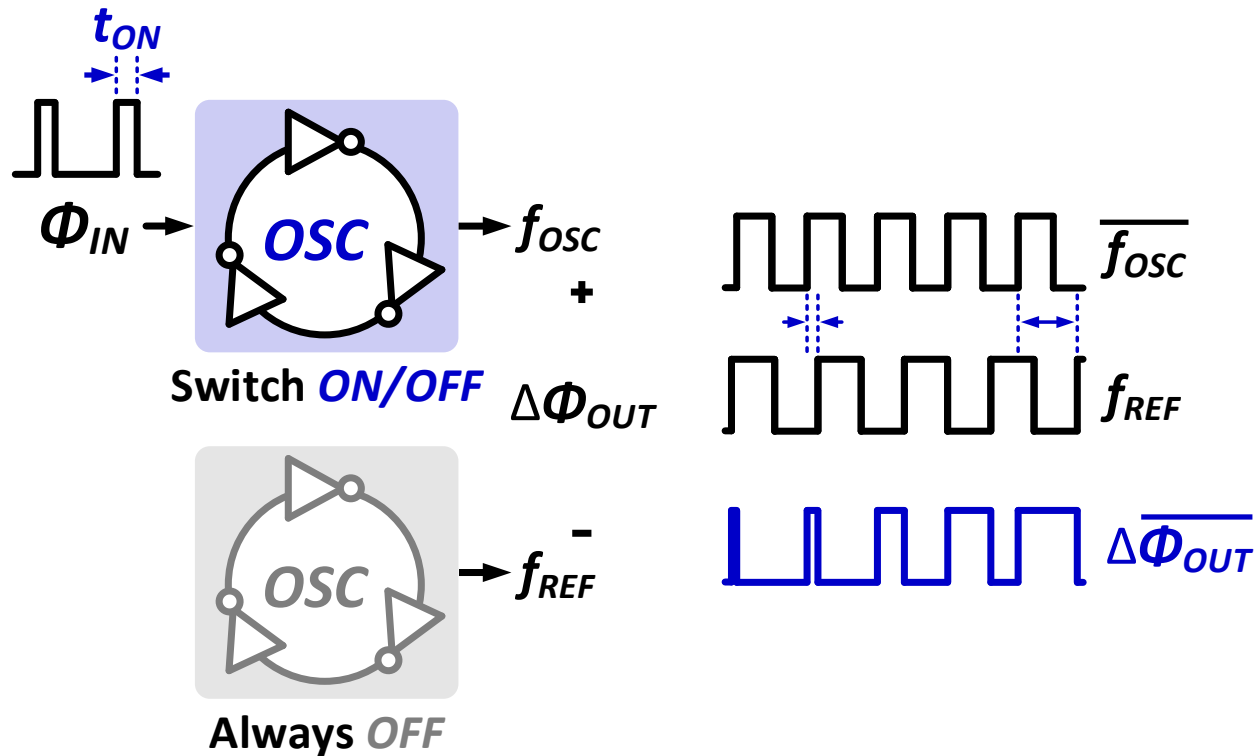
Accumulates t_{ON} over Time
 → Time Domain Accumulator!



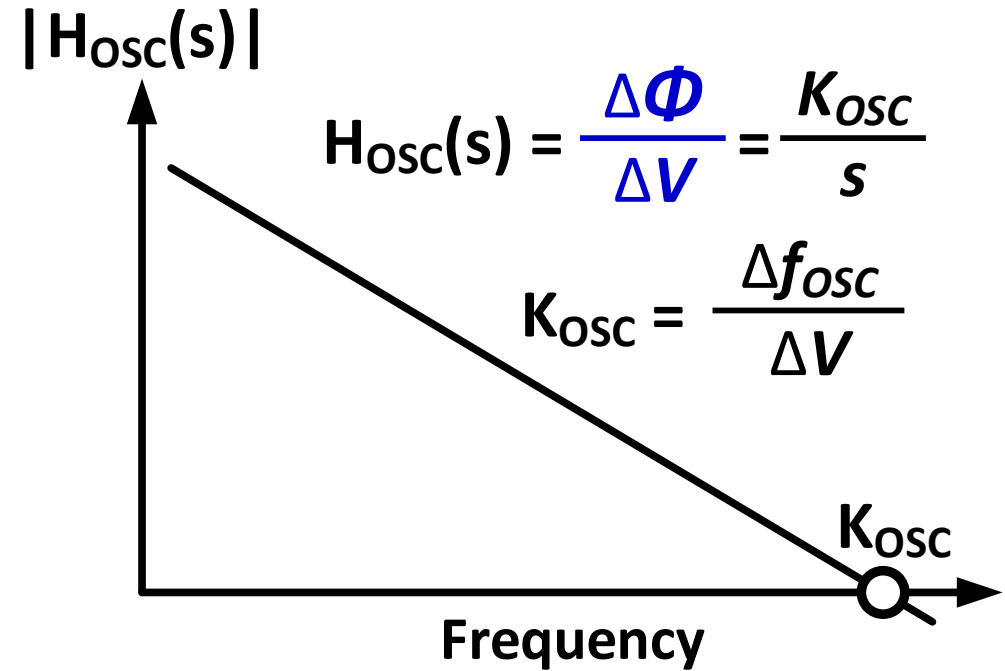
OSC is an **Ideal**
 Time-to-Phase **Integrator**



Analog FEx (Time Domain)

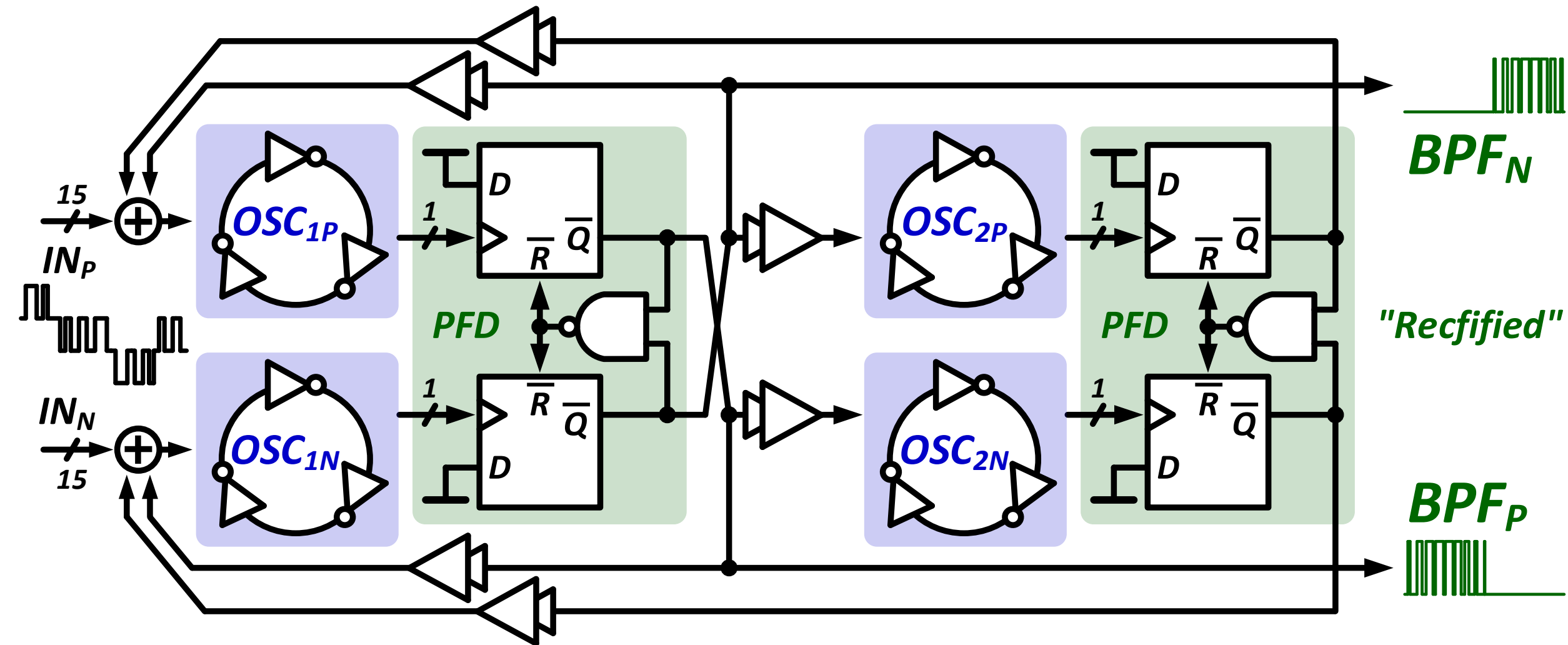


Accumulates t_{ON} over Time
 → Time Domain Accumulator!

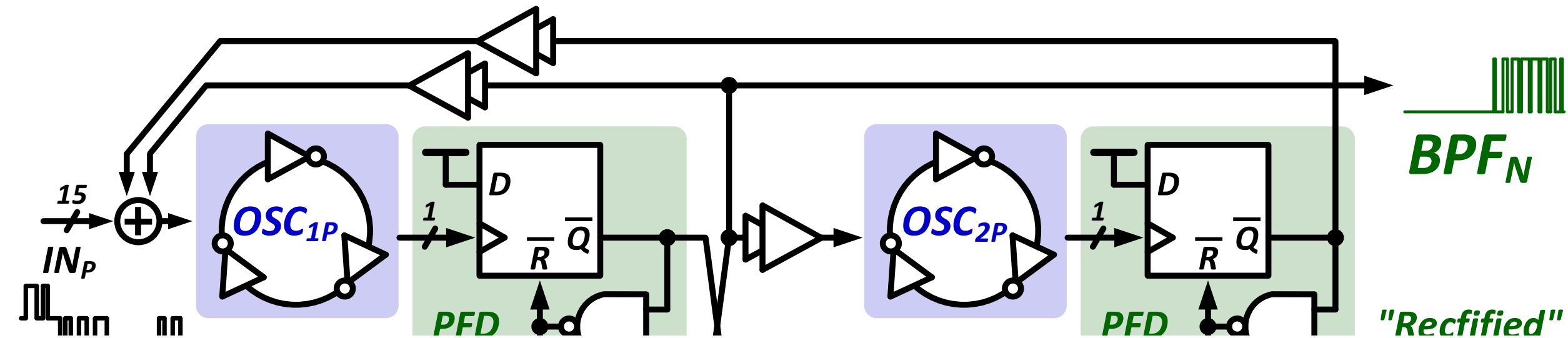


- ✓ Infinite DC Gain
- ✓ Free from Signal Swing Loss
- ✓ Uses Digital Cells

Time Domain 2nd-Order BPF w/ Inherent Rectification



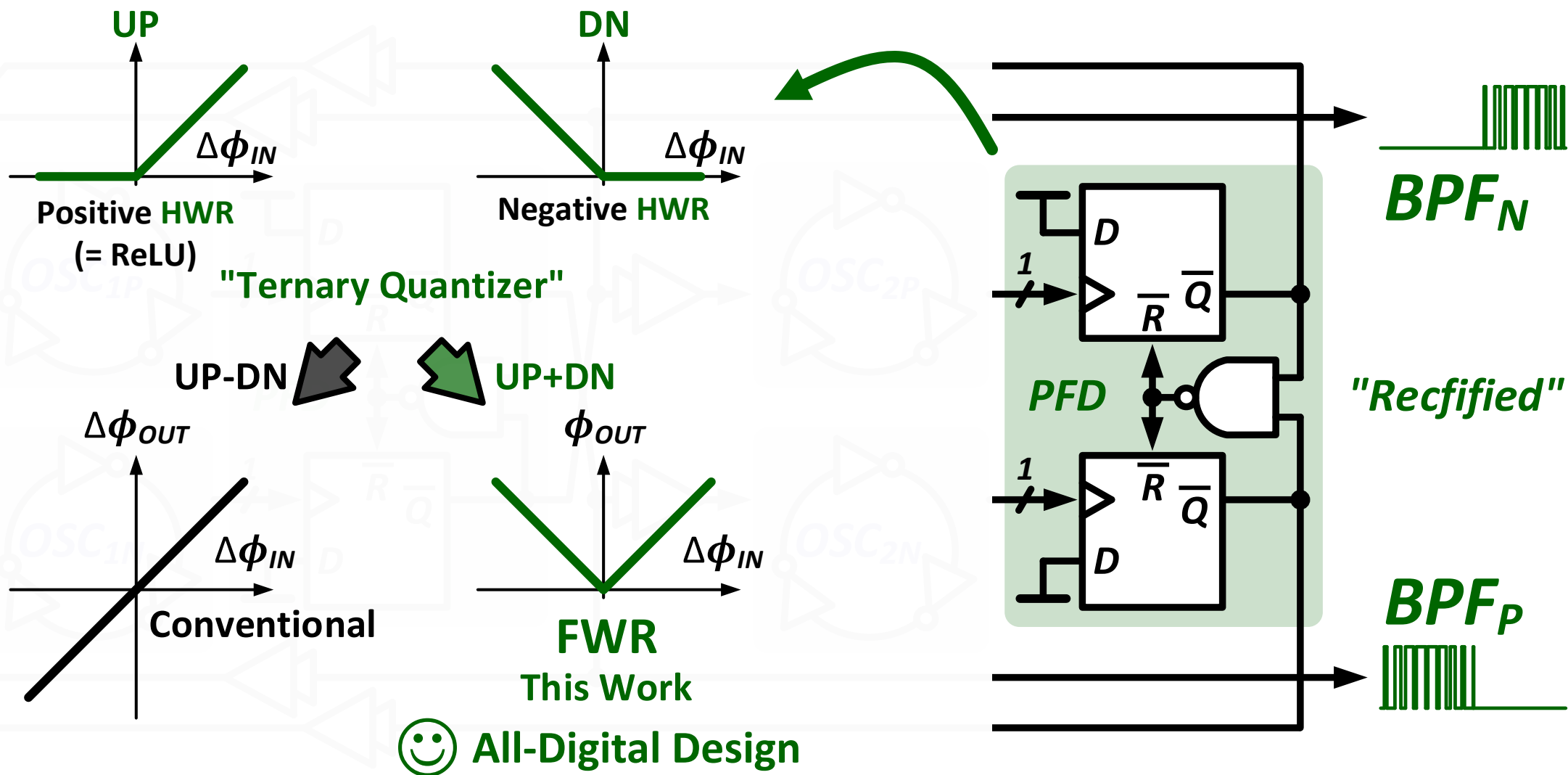
Time Domain 2nd-Order BPF w/ Inherent Rectification



Pseudo-Differential Circuit

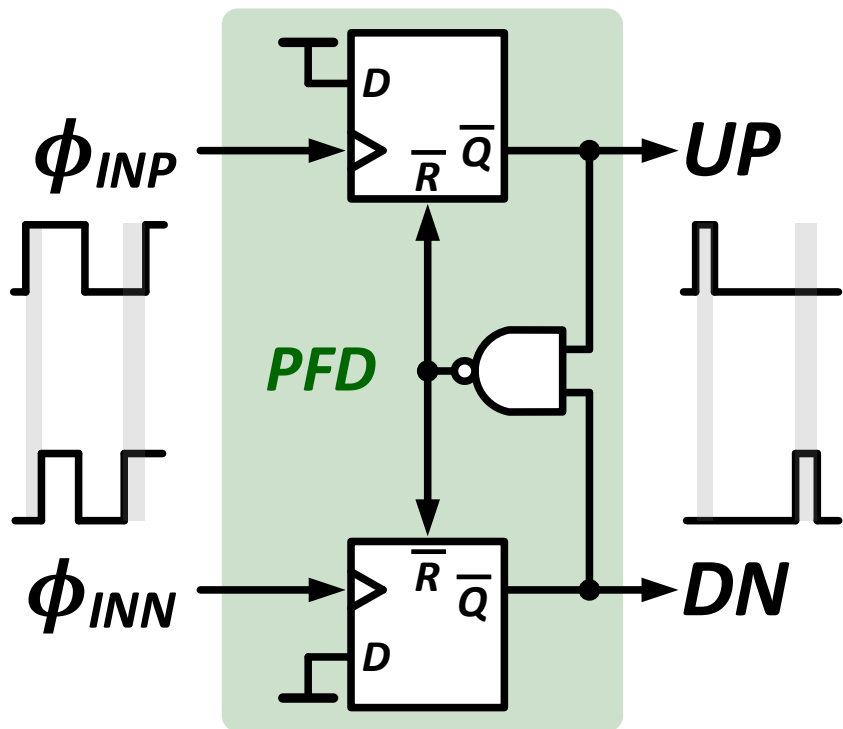
(2 Integrator + 2 Feedback → 2nd-Order Circuit)

Time Domain 2nd-Order BPF w/ Inherent Rectification



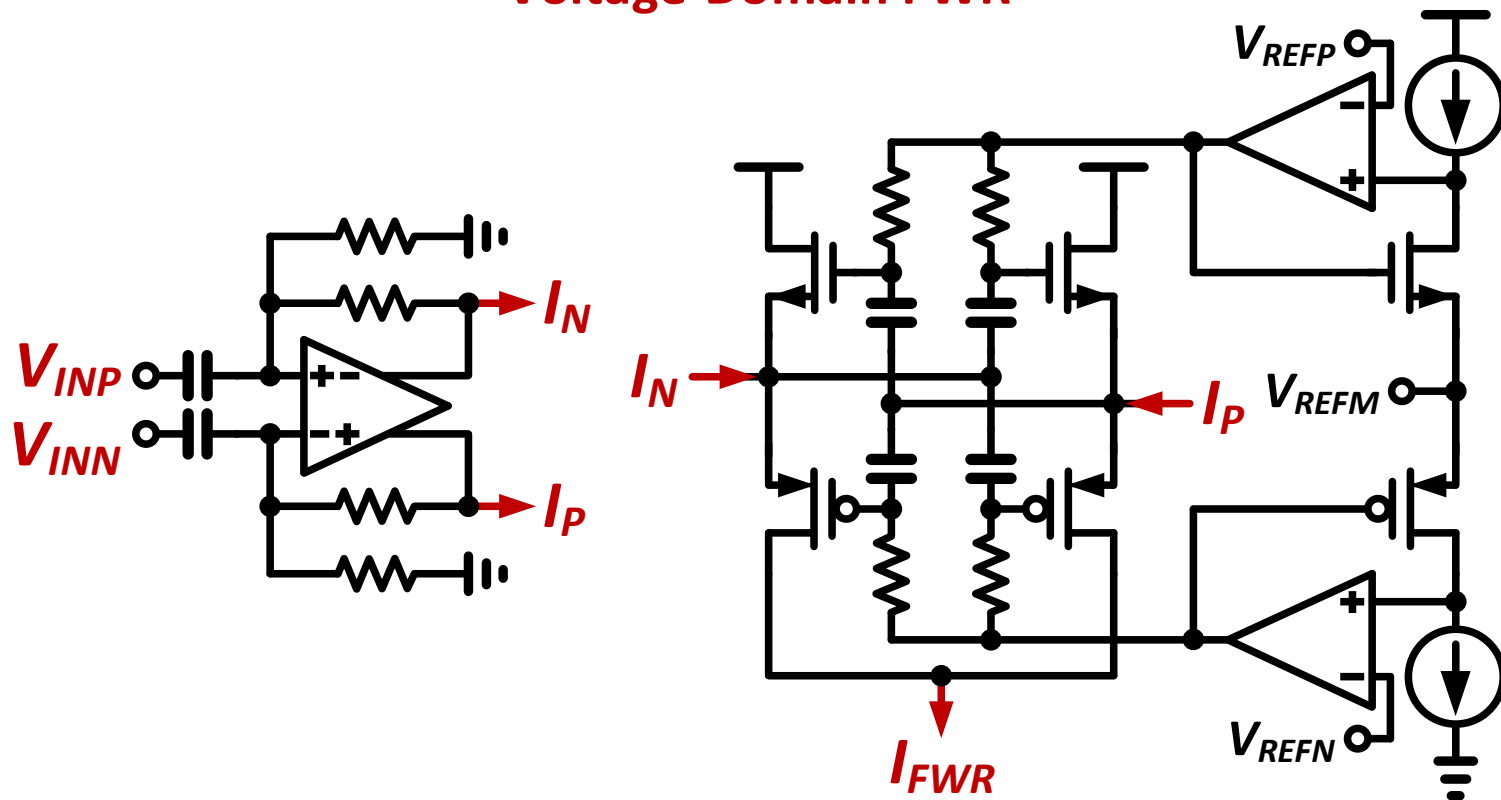
Time Domain¹ vs Voltage Domain² Rectifiers

Time-Domain FWR



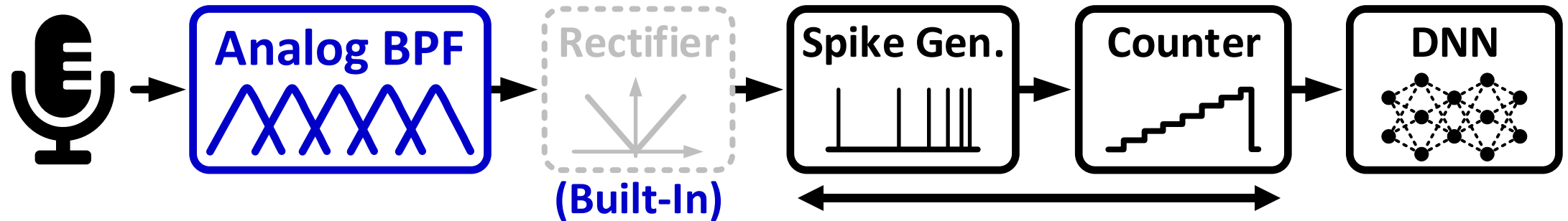
2 D-FF + 1 NAND

Voltage-Domain FWR



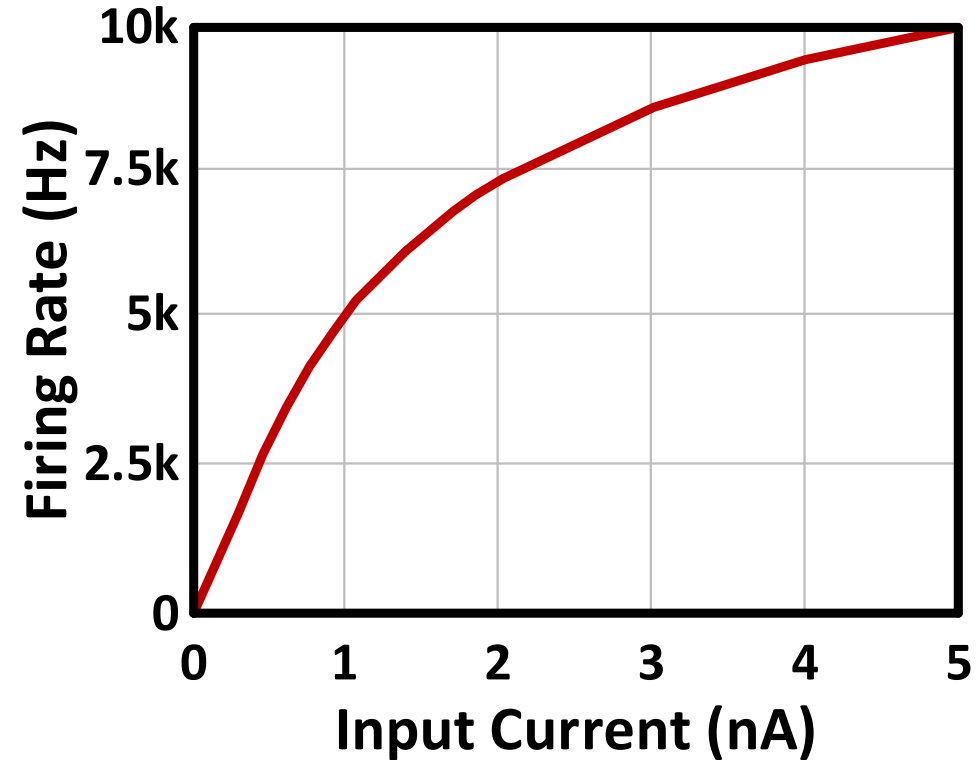
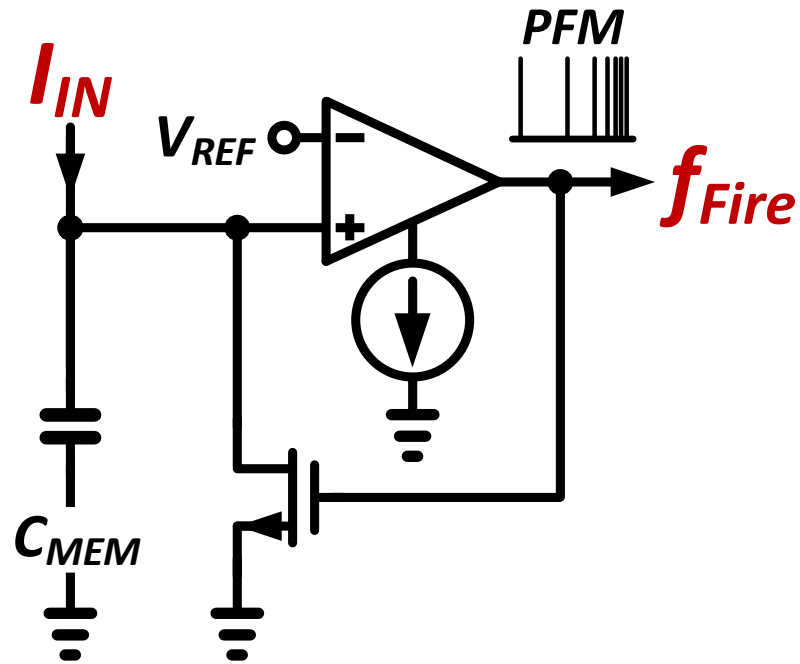
3 OTAs + 2 I_{BIAS} + 3 V_{REF} + 8 Resistors + 6 Capacitors

Time Domain Neuromorphic FEx



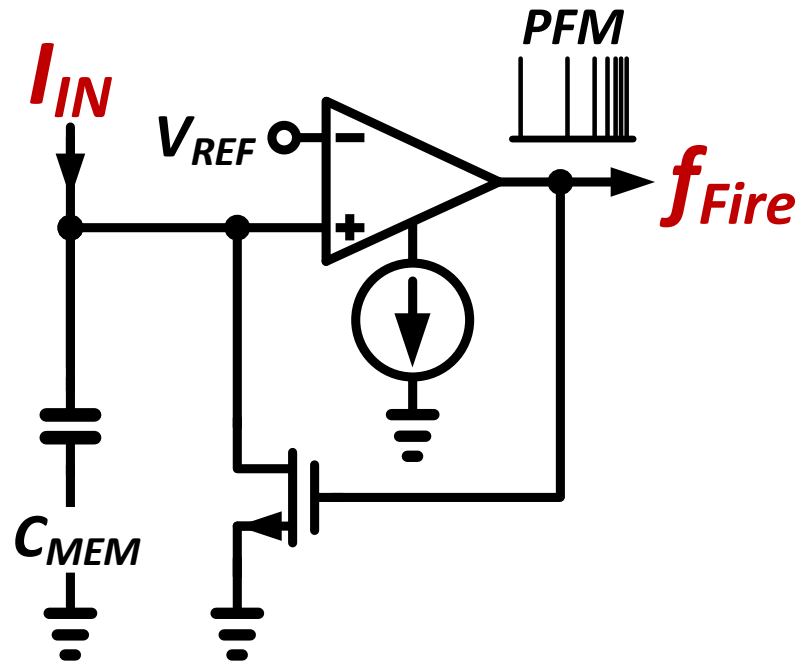
How about These Blocks?

Integrate-Fire (IF) Neuron (Voltage Domain)

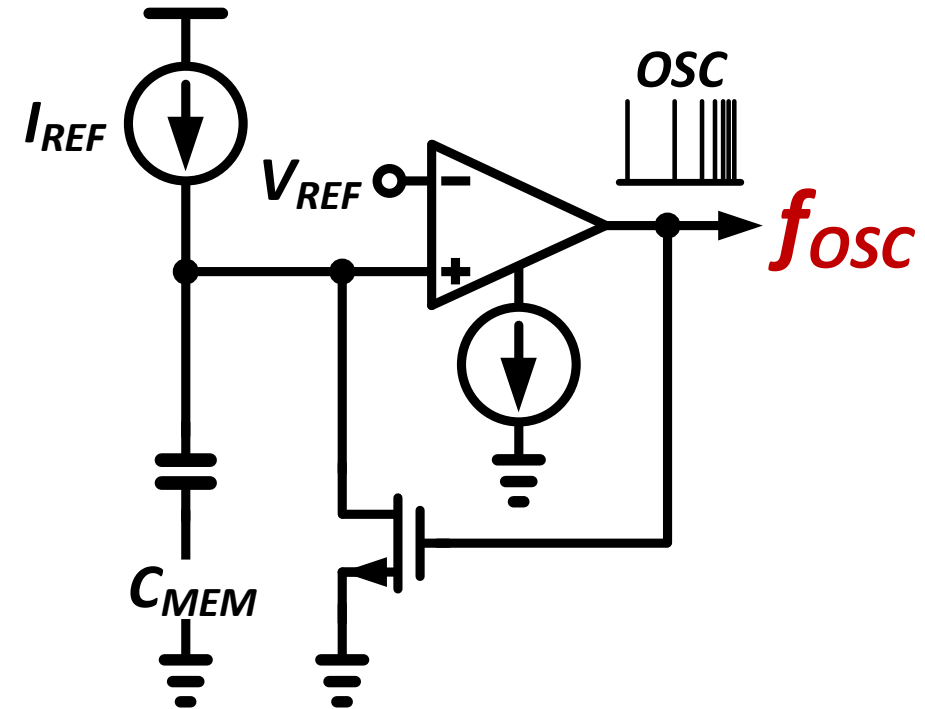


IF Neuron¹ Has Been Widely Adopted in Building Blocks of Analog Neuromorphic Circuits

Integrate-Fire (IF) Neuron (Voltage Domain)



IAF Neuron¹



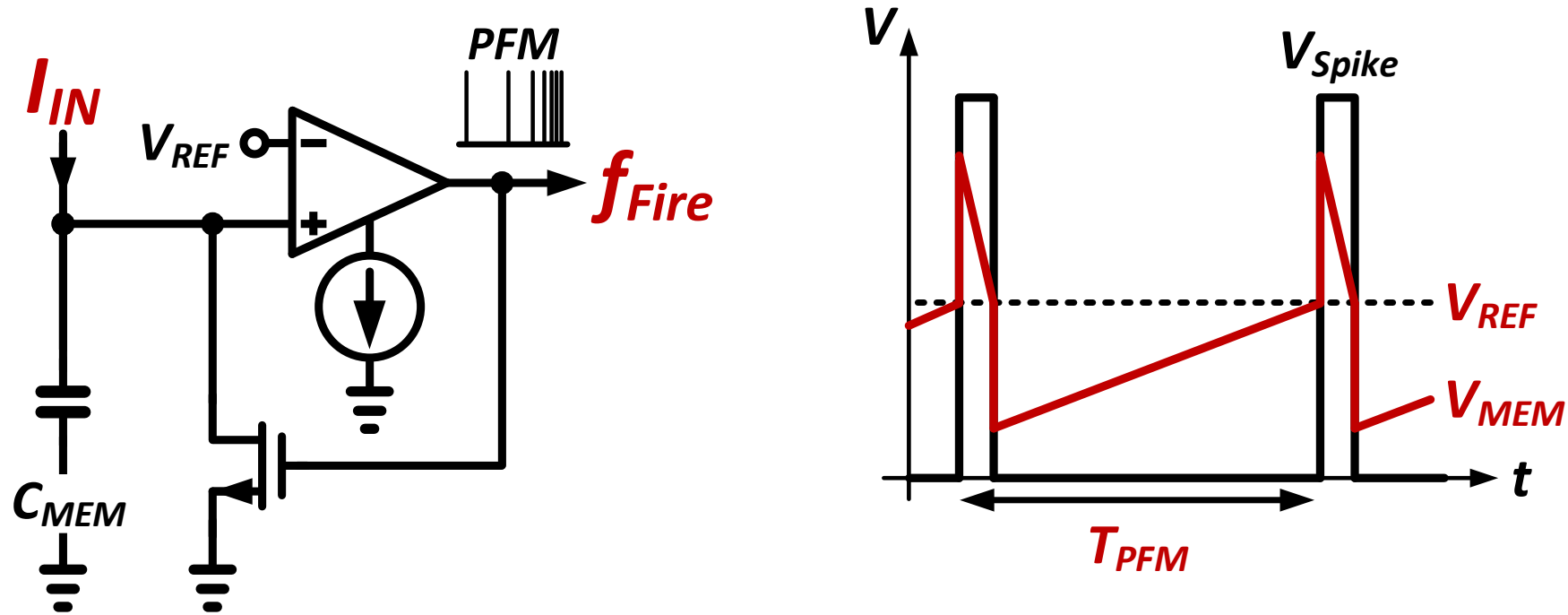
Relaxation Oscillator²

IF Neuron¹ Can be Interpreted as
a Current-Controlled **Relaxation Oscillator²**

¹C. Mead, 1989

²A. A. Abidi, JSSC 1983

Integrate-Fire (IF) Neuron (Voltage Domain)



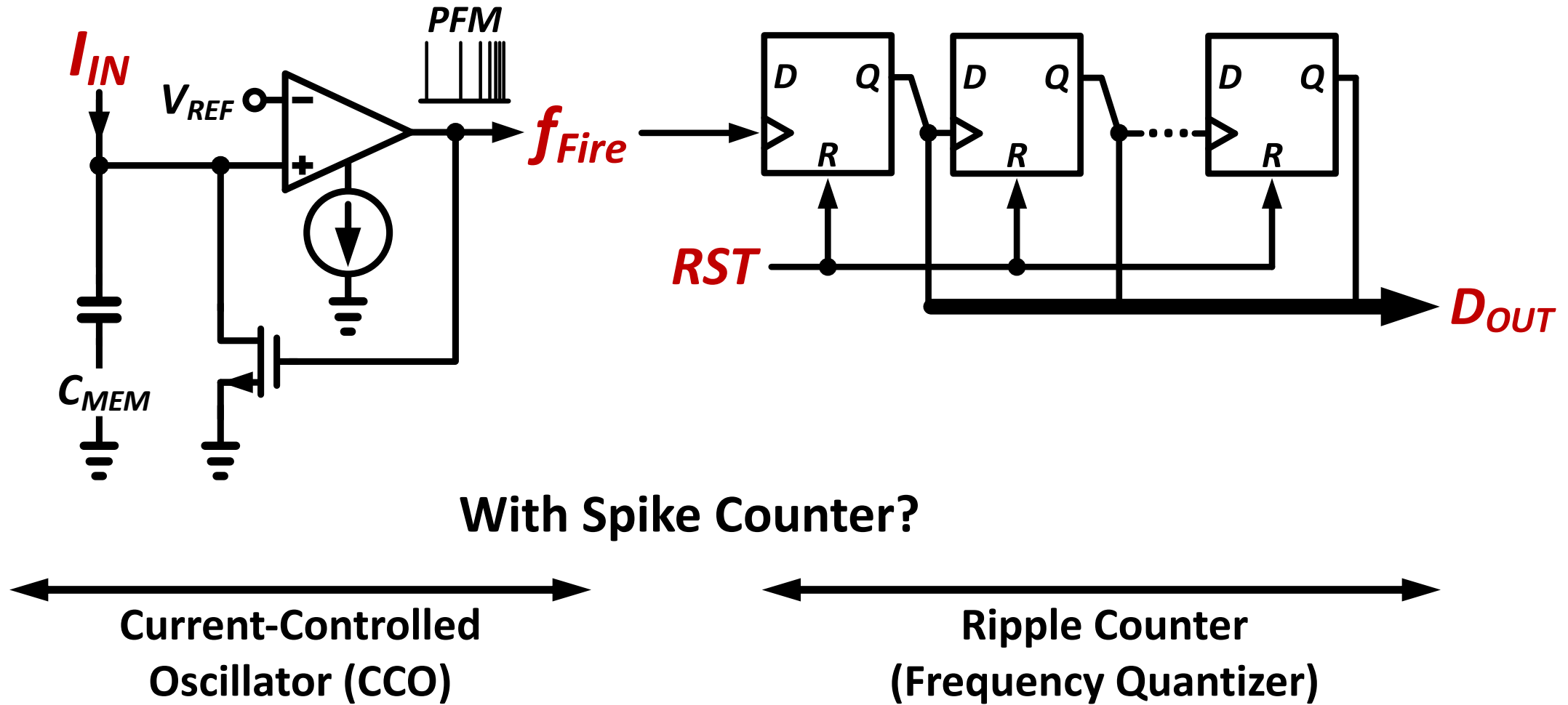
Regarding Technology Scaling?

Voltage Domain Integral (V_{MEM})

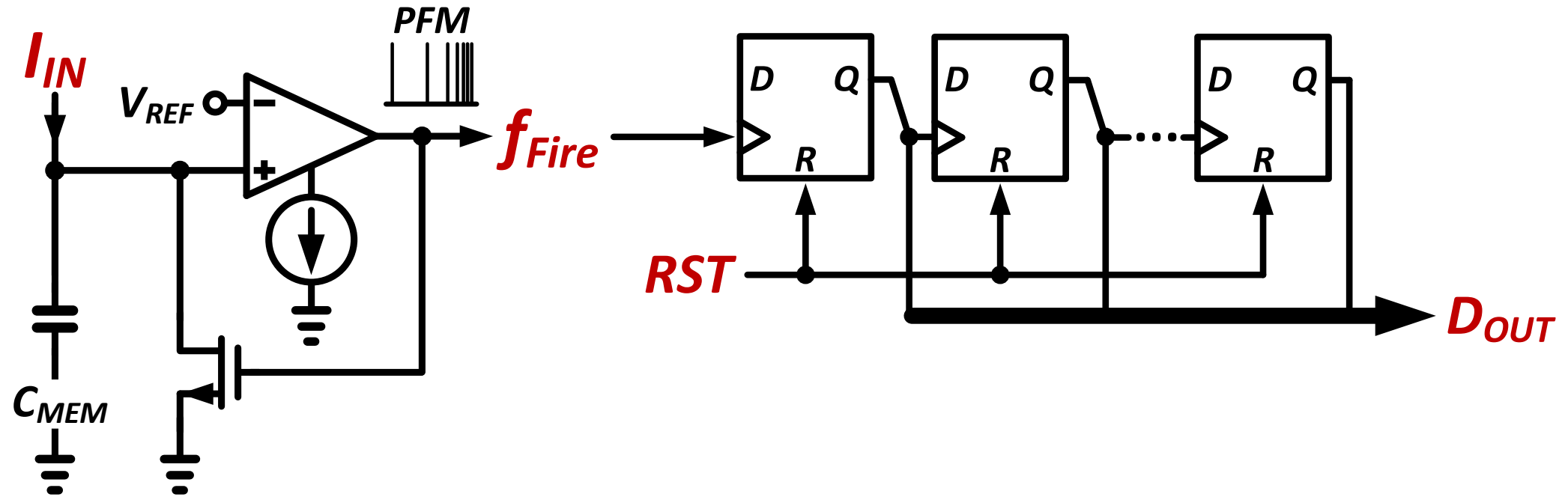
✗ Headroom

✗ Static Comparator

Integrate-Fire (IF) Neuron (Voltage Domain)

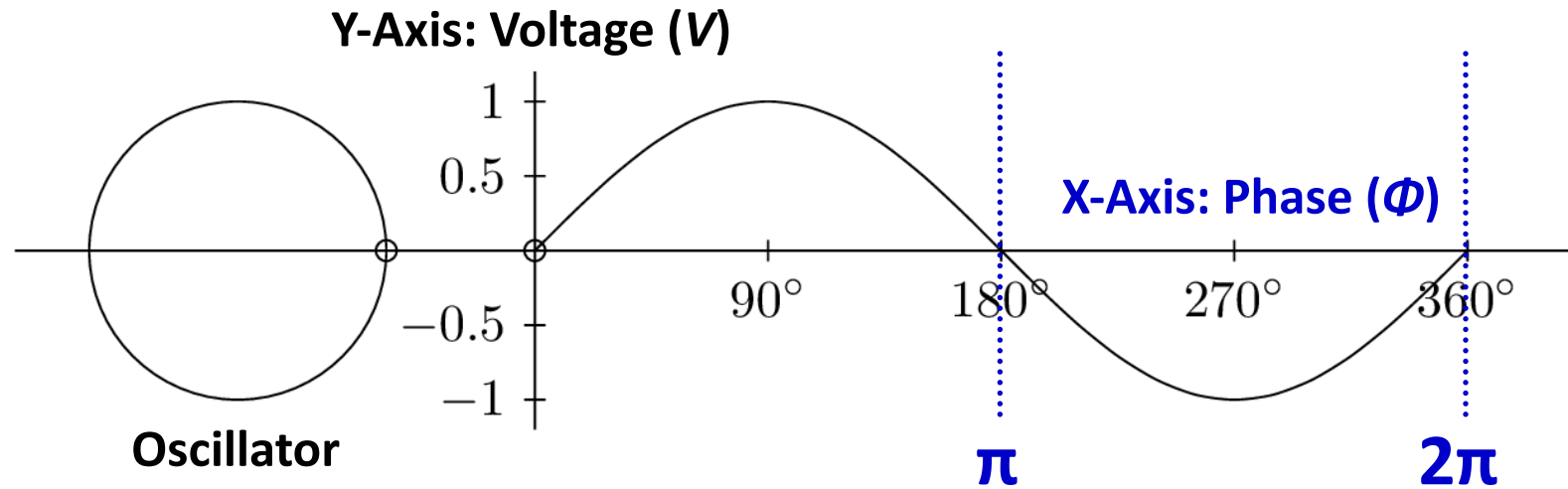
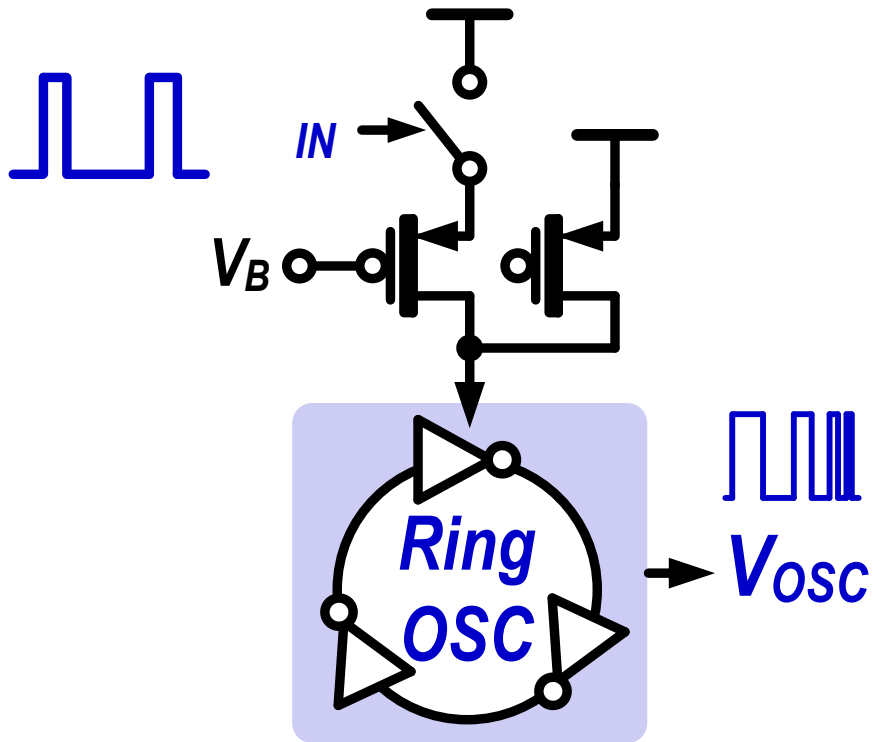


Integrate-Fire (IF) Neuron (Voltage Domain)



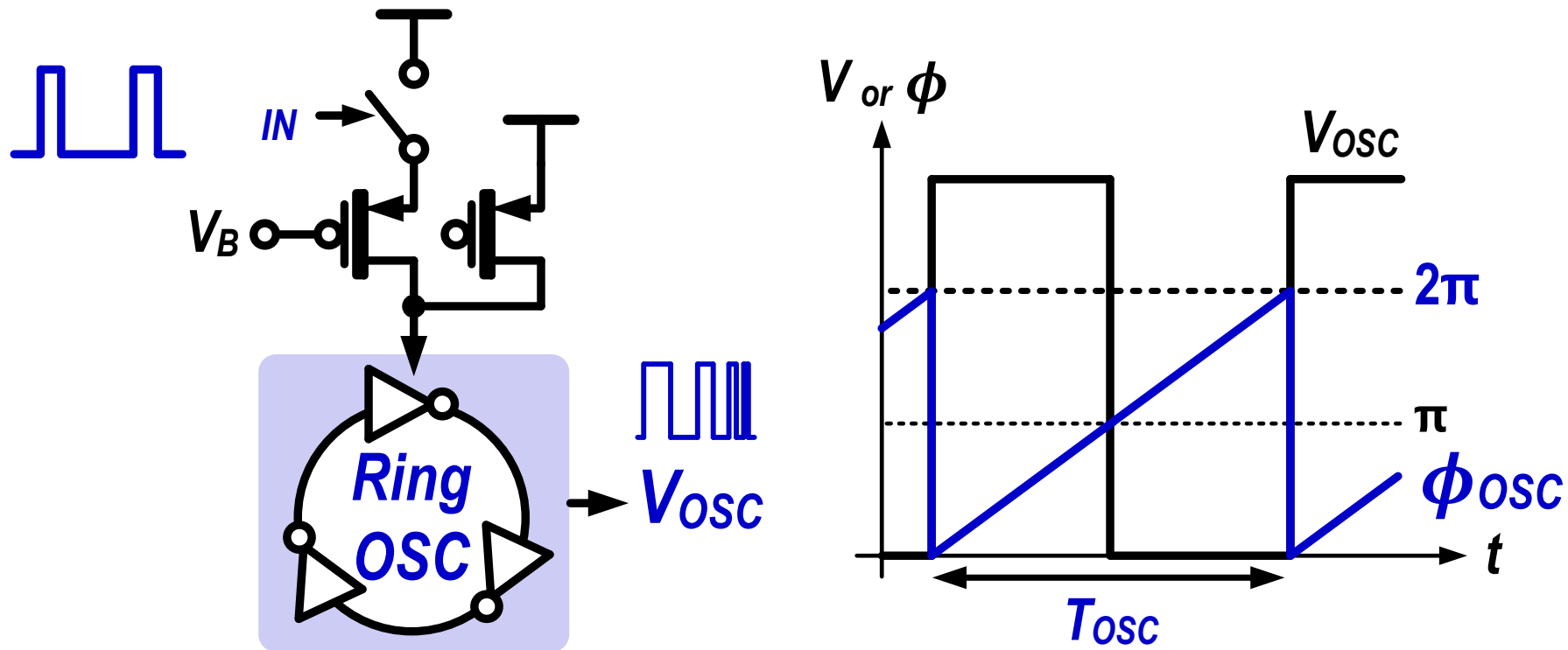
➔ CCO-Based $\Delta\Sigma$ Modulator !

Integrate-Fire (IF) Neuron (Time/Phase Domain)



Phase Domain IAF Neuron (Φ_{osc})

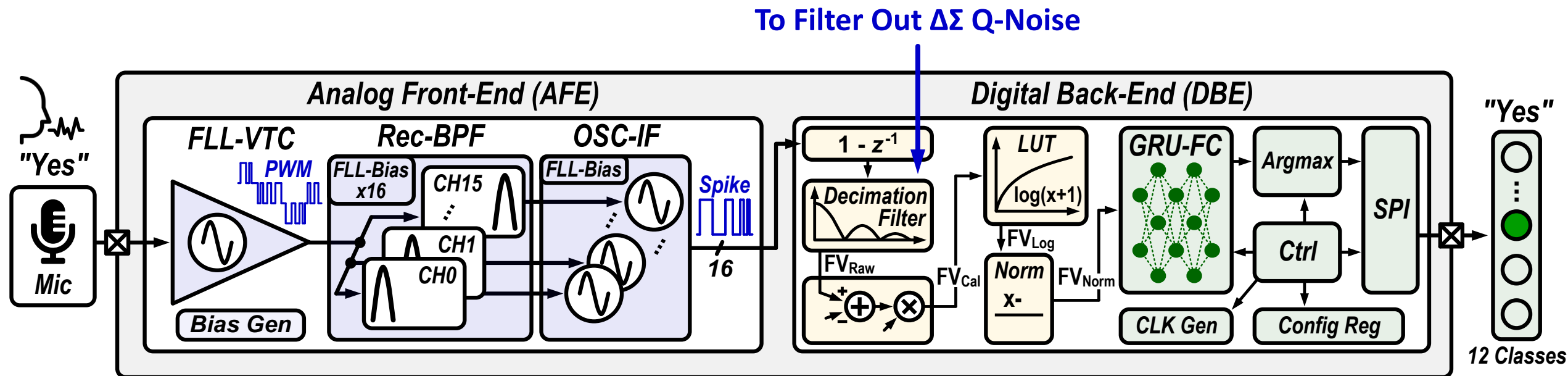
Integrate-Fire (IF) Neuron (Time/Phase Domain)



Phase Domain IAF Neuron (ϕ_{osc})

- ✓ Technology Scalable
- ✓ No Static Comparator (Inherent 2π Threshold)

Neuromorphic Keyword Spotting Chip



Ring-Oscillator-Based

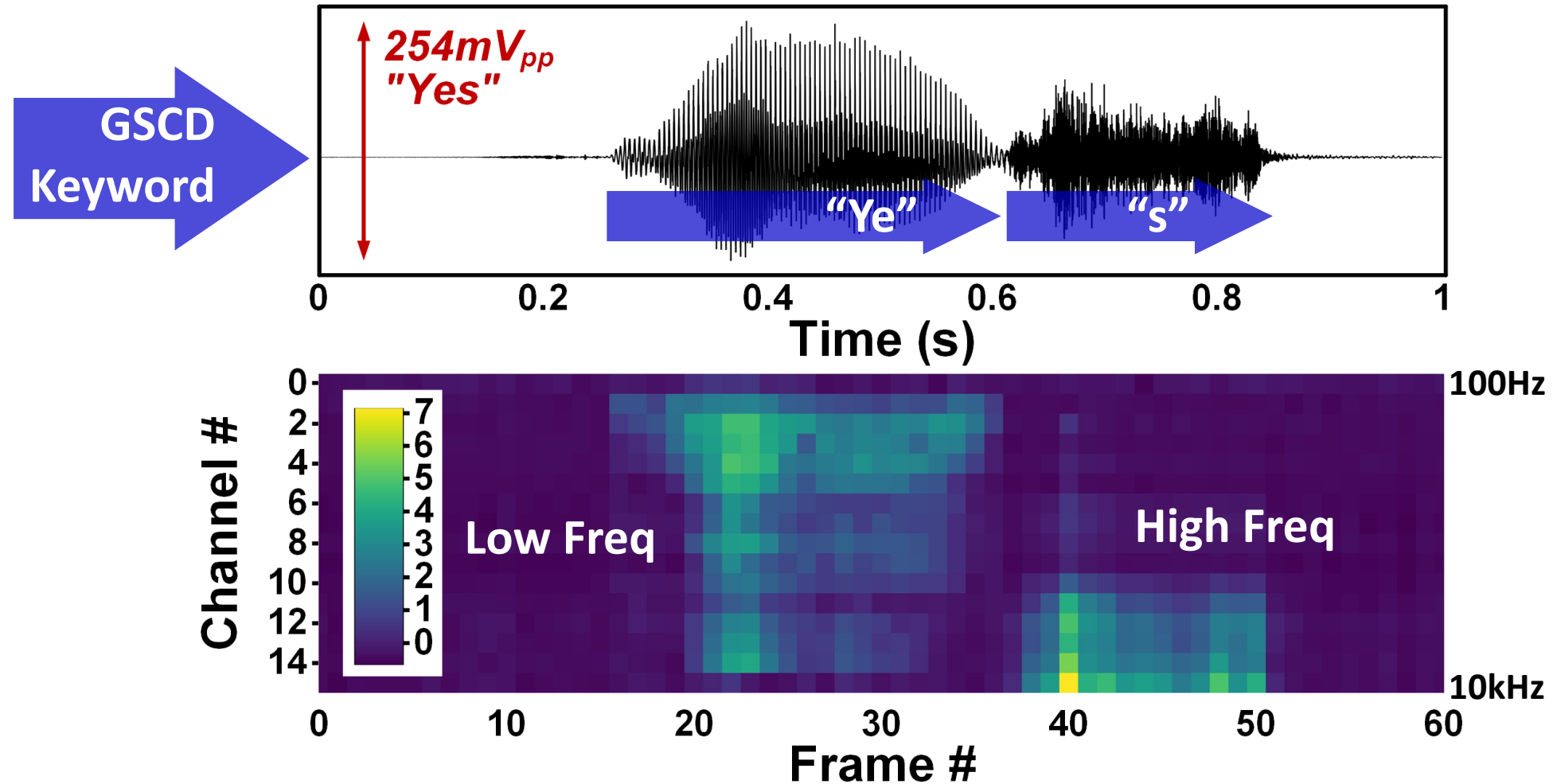
Time Domain Feature Extractor
(Technology Scalable)

GRU-FC RNN Classifier

The First Chip and *The Only Chip* that integrates
Time Domain Analog FEx + Digital Classifier all on-chip, even to date

Audio Response of FEx

GSCD: Google Speech Command Dataset



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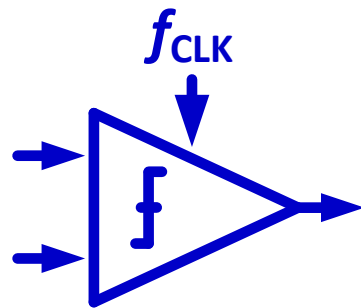
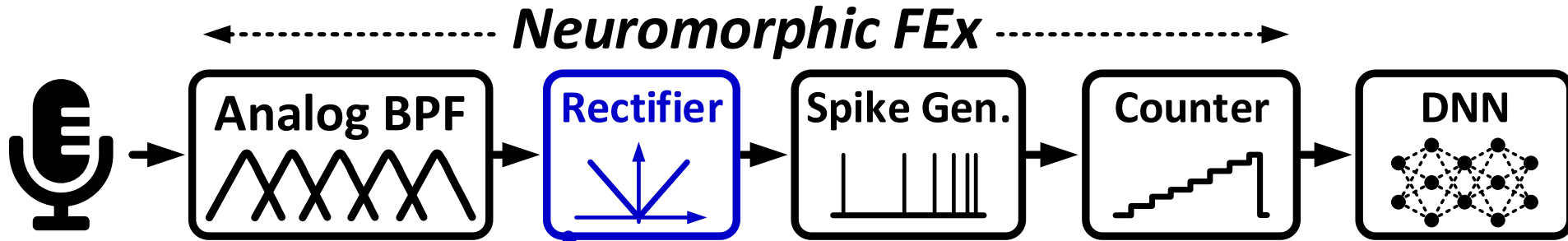
Sidharth Thomas; Sam Razavian; Wei Sun; Benyamin Fallahi Motlagh; Anthony D. Kim; Yu Wu; Benjamin S. Williams; Aydin Babakhani

A Calibration-Free Fractional-N Analog PLL With Negligible DSM Quantization Noise

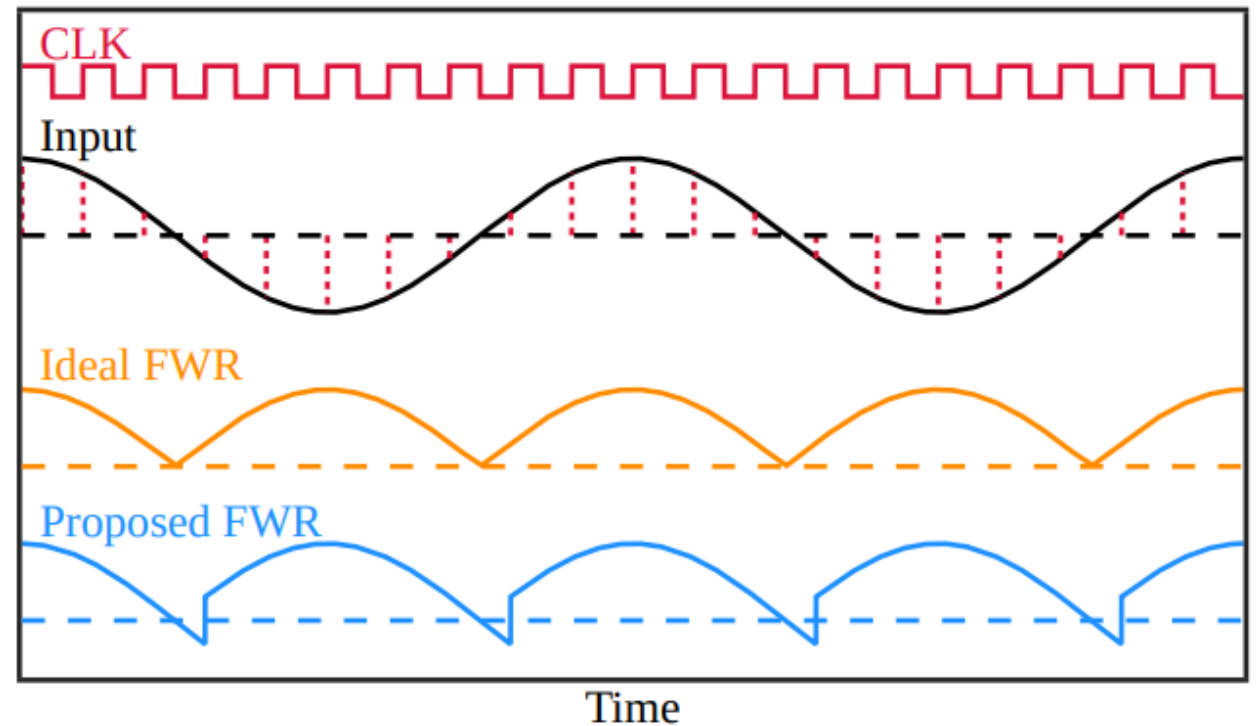
David Murphy; Dihang Yang; Hooman Darabi; Arya Behzad

Please Find Design Secrets in JSSC 2022 Paper!

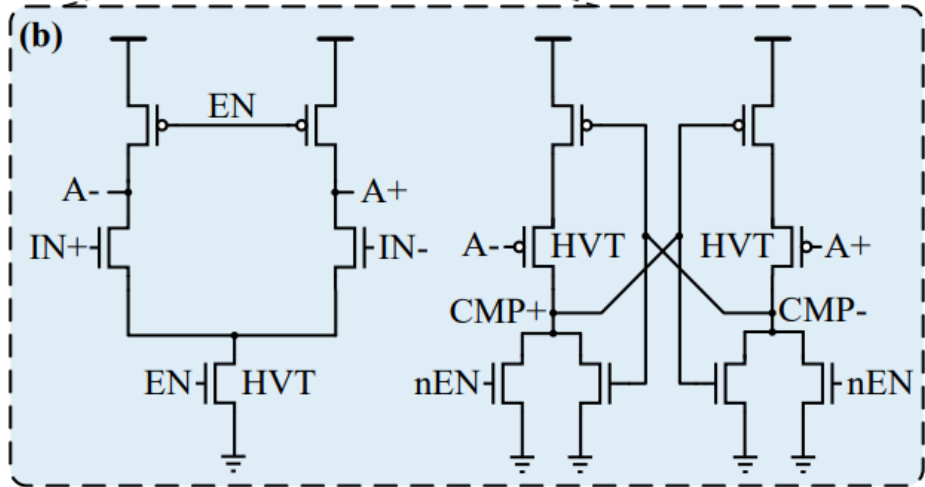
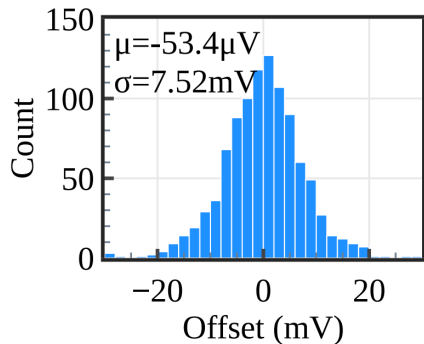
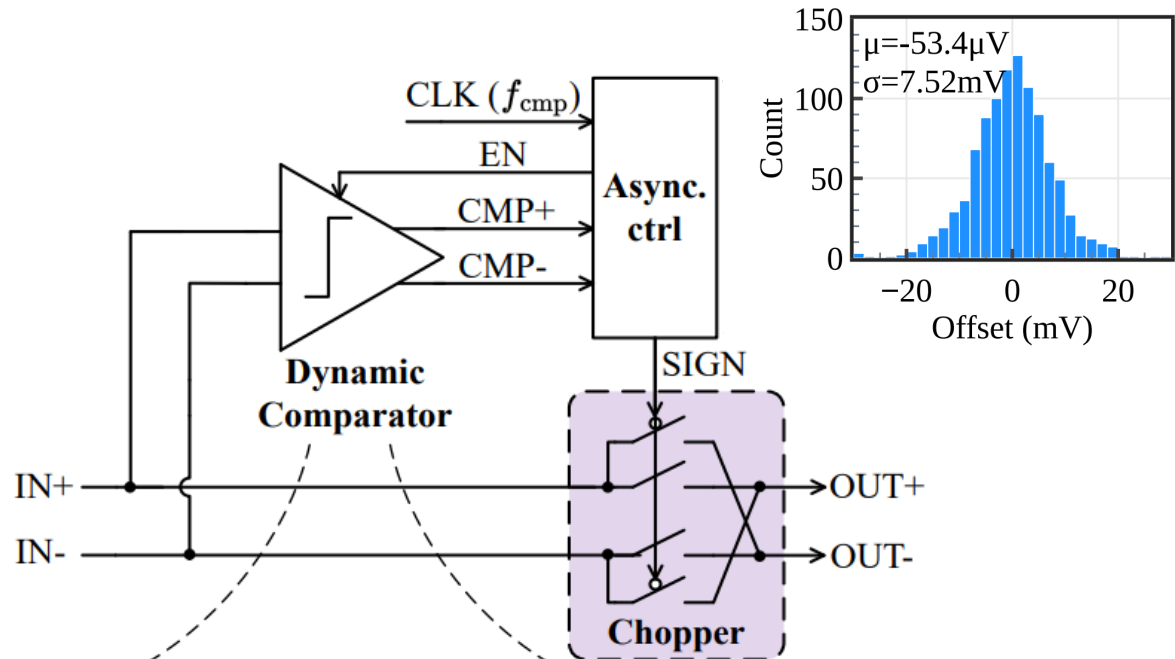
Hardware-Aware Algorithm Optimization



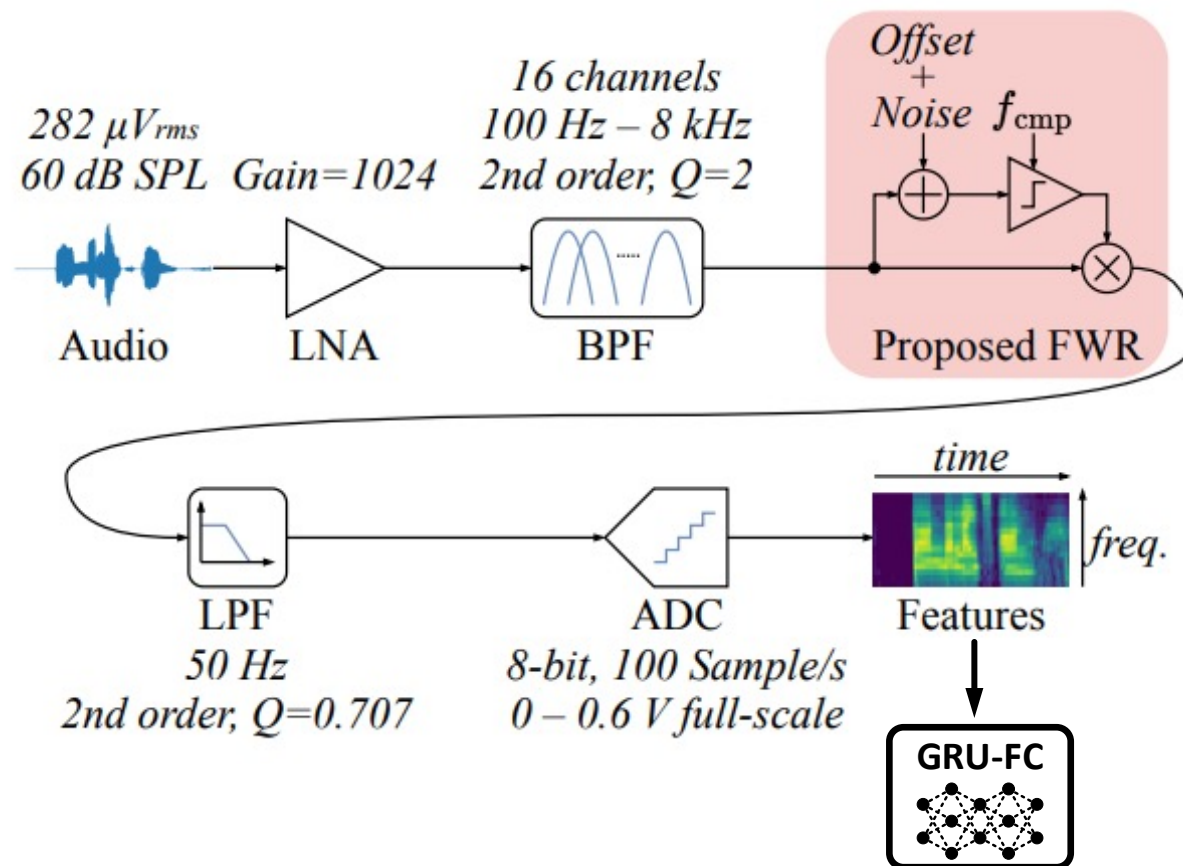
What If We Use
a **Dynamic Comparator**
To **Rectify** The Signal?



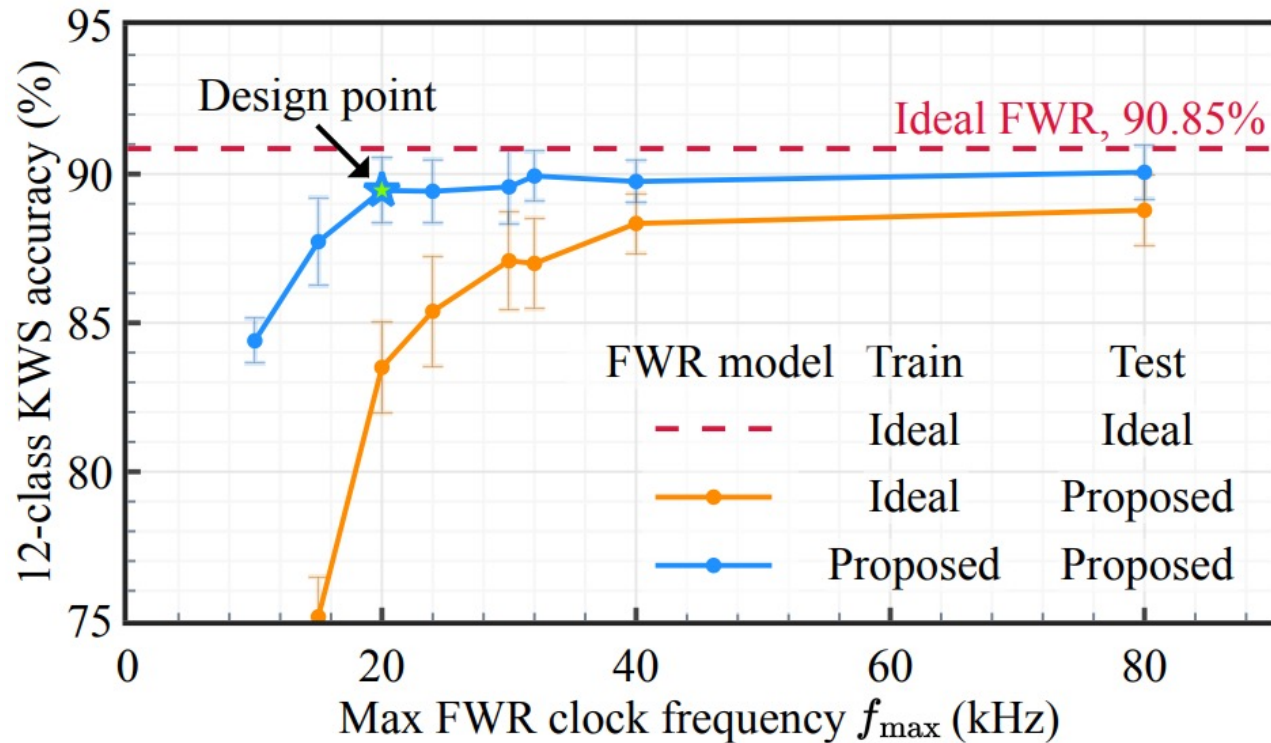
Hardware-Algorithm Co-Optimization



Analog Circuits are **Python-Modeled**,
Included into **RNN Training Loop**
Under **PyTorch Framework**



Hardware-Algorithm Co-Optimization



Hardware-Aware Training



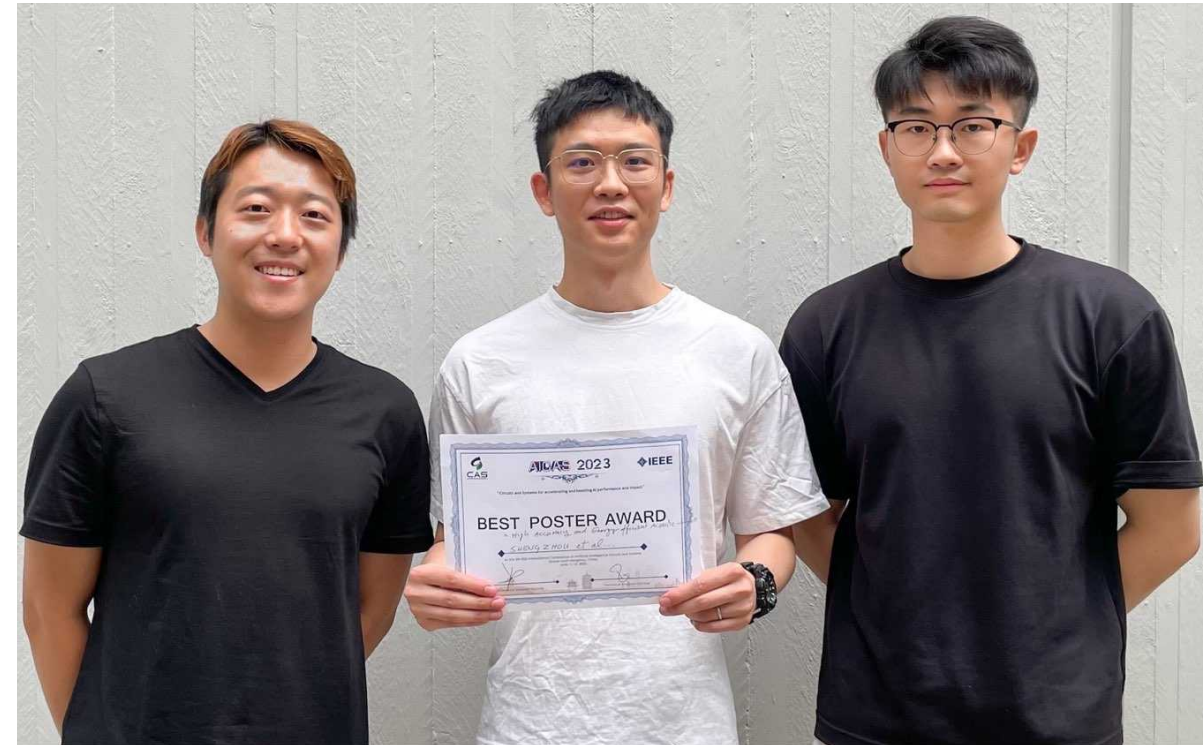
- ✓ **Negligible** Accuracy Loss
- ✓ **31.2x** Less Rectifier Power than State-of-the-Art
- ✓ **Without** Increase of Network Size

AICAS 2023 Best Poster Award

High-Accuracy and Energy-Efficient Acoustic Inference using Hardware-Aware Training and a 0.34 nW/Ch Full-Wave Rectifier

Sheng Zhou*, Xi Chen*, Kwantae Kim, Shih-Chii Liu

Institute of Neuroinformatics, University of Zurich and ETH Zurich, Zurich, Switzerland
{shengzhou, xichennn, kwantae, shih}@ini.uzh.ch



Outline

- Visions of IoT Sensors
- Biomedical Sensor
- Neuromorphic Sensor
- **Outlook**

Summary

- **Tiny, Sensory, Intelligent, Wireless IoT Platforms**
- **Bioimpedance Sensor**
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Summary

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 - Will Enrich Our Daily Lives in the Next 10 Years
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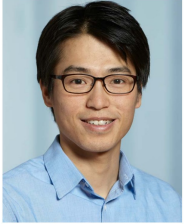
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- **Outlook**
 - Multiple Cross-Domain / Interdisciplinary Research Opportunities
 - Wireless TRx, Device-Circuit, Circuit Theory, Vocal Studies, ...

Acknowledgments

ETH zürich



Taekwang Jang

ETH zürich



Tobi Delbruck



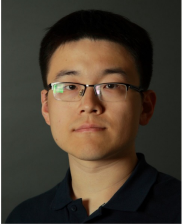
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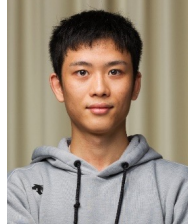


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Sangyeob Kim

Thank you!

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